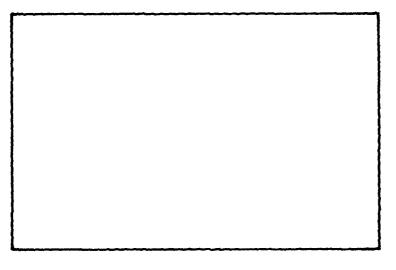
ILLINOIS UNIV AT CHICAGO CIRCLE ELECTROMAGNETIC IMAGI--ETC F/G 20/14
A HIGH FREQUENCY INVERSE SCATTERING MODEL TO RECOVER THE SPECUL--ETC(U)
MAY 82 B FOO N00014-80-C-0708 AD-A115 805 UNCLASSIFIED EMID-CL-1982-05-21-01 NL 1 or 2 40 A 115805

7	
10	
20	
10	
_	
A	
AD	

COMMUNICATIONS LABORATORY



Department of Information Engineering University of Illinois at Chicago Circle Box 4348, Chicago IL 60690, USA



TIE FILE COPY

SELECTE JUN 1 8 1982

DISTRIBUTION STATEMENT A

Approved for public release; Distribution Unlimited

Q 9

Acoustics, Electromagnetics, Optics - Circuits and Networks
Communication Theory and Systems - Electronic Devices



COMMUNICATIONS LABORATORY

Department of Information Engineering University of Illinois at Chicago Circle Box 434B, Chicago IL 606BD,USA

A HIGH FREQUENCY INVERSE SCATTERING

MODEL TO RECOVER THE SPECULAR POINT CURVATURE
FROM POLARIMETRIC SCATTERING DATA

by

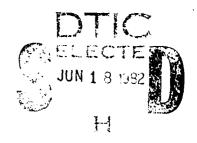
Bing-Yuen Foo

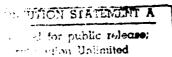
Communications Laboratory Report Number

EMID-CL-1982-05-21-01

M.Sc. Thesis

May 21, 1982





SECURITY CLASSIFICATION OF THIS PAGE (When Data Entered)

ECIPIENT'S CATALOG NUMBER
YPE OF REPORT & PERIOD COVERED M.S. Thesis ERFORMING ORG. REPORT NUMBER EMID-CL-82-05-18-01
ONTRACT OF GRANT NUMBER(*) ONE NOO014-80-C-0708 ARMY DAAG 29-80-K-0027 ONE NOO014-80-C-0773
PROGRAM ELEMENT, PROJECT, TASK AREA & WORK UNIT NUMBERS
REPORT DATE May 21, 1982 NUMBER OF PAGES 16 + 145 = 161
SECURITY CLASS. (of this report) UNCLASSIFIED DECLASSIFICATION/DOWNGRADING

16. DISTRIBUTION STATEMENT (of this Report)

UNLIMITED

17. DISTRIBUTION STATEMENT (of the abatract entered in Block 20, if different from Report)

UNLIMITED

18. SUPPLEMENTARY NOTES

The findings of this report are not to be construed as an official Department of the Navy position unless so designated by other authorized documents.

19. KEY WORDS (Continue on reverse side if necessary and identify by block number)

Specular point curvature, scattering matrix element, phase difference of like polarized terms, circular polarization.

20. ABSTRACT (Continue on reverse side if necessary and identify by block number)

Based on the time-domain first order correction to the physical optics current approximation, a relationship between the phase factors of the polarimetric scattering matrix elements and the principal curvatures at the specular point of a scatterer is established.

(Continued on next page)

DD 1 JAN 73 1473

EDITION OF 1 NOV 65 IS OBSOLETE S/N 0102-LF-014-6601

UNCLASSIFIED

SECURITY CLASSIFICATION OF THIS PAGE (When Data Entered)

The above phase-curvature relationship is tested by applying it to theoretical as well as experimental backscattering data obtained for a prolate spheroidal scatterer. The results of these tests not only determine the acceptability of the phase-curvature relationship, they also point out the range of kb values over which the first order correction to the physical optics currents is valid, and which serves as a compromise range between the high frequency condition required by the curvature recovery model and the drawback to lower frequencies required to prevent critical magnification of measurement errors.

Another curvature recovery equation is derived by transforming the linear polarization basis to the circular polarization basis. The curvature recovery model is proven to satisfy the image reconstruction identities of invariant transformation. A scattering ratio is defined and its behavior is investigated at high frequencies. Its plot on the complex plane provides a simple way to help judge the accuracy of polarimetric scattering measurement, regardless of whether a linear or a circular polarization basis is used.

Acce	ssion For	A SE
NTIS	GRA&I	
DTIC		- 1
Unan	aounded 🚟	
Just	ification	
	entrem and the second and	
By		
Dist	ribution/	
Ava	114 .lium Codes	
	A distribution	
Mst	p.oial	
Λ	1 4	ı
5	1	- 1
H	1	- 1

INSPECTED

UNCLASSIFIED

A HIGH FREQUENCY INVERSE SCATTERING MODEL TO RECOVER THE SPECULAR POINT CURVATURE FROM POLALLMETRIC SCATTERING DATA

BY

BING-YUEN FOO

B.Sc. University of Manitoba

THESIS

May 21, 1982

Submitted in partial fulfillment of the requirements for the degree of Master of Science in Information Engineering in the Graduate College of the University of Illinois at Chicago Circle, 1982

Chicago, Illinois

UNIVERSITY OF ILLINOIS AT CHICAGO CIRCLE THE GRADUATE COLLEGE

May	21,	1982	

I HEREBY RECOMMEND THAT THE THESIS PREPARED UNDER MY
SUPERVISION BY Bing-Yuen Foo
ENTITLED A High Frequency Inverse Scattering Model to Recover
the Specular Point Curvature from Polarimetric Scattering Data
BE ACCEPTED IN PARTIAL FULFILLMENT OF THE REQUIREMENTS FOR
THE DEGREE OF Master of Science (Electrical Engineering)
Dr. Wolfgang-M. Boerner Walloam M. Bourn

Recommendation concurred in

Dr. Wolfgang-M. Boerner &

Dr. Bruce H. McCormick

Dr. Wai-Kai Chen

Dr. Piergiorgio L. E. Uslenghi

Dr. J. Richard Huynen (ext.)

Dr. Jonathan D. Young (ext.) (ACCEPTED BY CORRESPONDENCE)

Dr. Sujeet K. Chaudhuri (ext.) (ACCEPTED BY CORRESPONDENCE)

Committee

on

Final Examination

ABSTRACT

Based on the time-domain first order correction to the physical optics current approximation, a relationship between the phase factors of the polarimetric scattering matrix elements and the principal curvatures at the specular point of a scatterer is established.

The above phase-curvature relationship is tested by applying it to theoretical as well as experimental backscattering data obtained for a prolate spheroidal scatterer. The results of these tests not only determine the acceptability of the phase-curvature relationship, they also point out the range of kb values over which the first order correction to the physical optics currents is valid, and which serves as a compromise range between the high frequency condition required by the curvature recovery model and the drawback to lower frequencies required to prevent critical magnification of measurement errors.

Another curvature recovery equation is derived by transforming the linear polarization basis to the circular polarization basis.

The curvature recovery model is proven to satisfy the image reconstruction identities of invariant transformation. A scattering ratio is defined and its behavior is investigated at high frequencies. Its plots on the complex plane provides a simple way to help judge the accuracy of polarimetric scattering measurement, regardless of whether a linear or a circular polarization basis is used.

ACKNOWLEDGMENTS

I am deeply indebted to my adviser, Dr. Wolfgang-Martin Boerner, who introduced to me the topic of Electromagnetic Inverse Scattering, continually exposed me to different research fields, and guided the research and the completion of this thesis. His enthusiasm, his dynamic attitude and dedication to research, which I admire, impress me profoundly.

I am also indebted and grateful to Dr. Sujeet Kumar Chaudhuri, who enlightened me with much patience during his stay in Chicago in 1981. Not only did he derive the phase-curvature relationship described in this thesis, but he also kindly provided the theoretical data for numerical verification. Without the full direction and help of Dr. Boerner and Dr. Chaudhuri, the completion of this thesis would not have been possible.

I would like to express my deep gratitude to Dr. Jonathan D. Young, of the Electro-Science Laboratory at the Ohio State University, Columbus, for providing the smoothened scattering data, without which many insights of the numerical investigation would not have been obtained. I also thank him for allowing me to view the complete procedure of obtaining the measurement data during a recent visit of the EMID-CL-UIC research staff to the Electro-Science Laboratory on March 25 and 26, and for his patient explanations, illuminating discussions and instructions for the appropriate use of the data.

I am extremely grateful to Mr. Benedict Chuk-Min Ho, who patiently helped me comprehend the theory of Electromagnetic Inverse Scattering.

The many invaluable discussions, suggestions and continual assistance must be acknowledged.

This work was supported, in parts, by the National Science and Engineering Research Council of Canada under Operating Grants A7240 and A3804, the U.S. Office of Naval Research under Grant N00014-80-C-0773, the Naval Air Systems Command Research Program under N00019-80-C-0620, and the U.S. Army Research Office under D-AAG-28-80K-0027. I am very thankful for all the financial support I received during my studies and the preparation of this thesis.

I would like to thank M.K who has been giving me constant encouragement and incentive.

Special thanks are due to Mr. Chau-Wing Yang for his fruitful discussions and great assistance in the preparation of this thesis.

Special thanks are also due to Miss Kathleen Sluis and Miss Sue Bourgart for arranging to let me use the facilities in their office during the typing of the manuscript.

Last but not least, I would like to acknowledge my colleagues in the Electromagnetic Imaging Division, namely, Mr. C.Y. Chan, Mr. S. Saatchi, Mr. M. Davidovitz, Mr. J. Nespor, Mr. B. Beker, Mr. A. Manson, Mr. V. Mirmira and Mr. D. Kruchten, for their encouragement and stimulating discussions.

TABLE OF CONTENTS

PAGE	_
ABSTRACT iii	
ACKNOWLEDGMENTSiv	,
TABLE OF CONTENTS vi	
LIST OF SYMBOLS viii	
LIST OF FIGURES xii	
CHAPTER I INTRODUCTION AND LITERATURE REVIEW 1	
1.1 Introduction 1	
1.2 Literature Review	:
CHAPTER II THE SPACE-TIME INTEGRAL EQUATION AND FIRST ORDER CORRECTION TO PHYSICAL OPTICS	,
2.1 Introduction 7	,
2.2 The Space-Time Integral Equation 8	;
2.3 A First Order Correction to Physical Optics Using the Space-Time Integral Equation	
CHAPTER III CURVATURE RECOVERY FROM HIGH FREQUENCY [S] MATRIX ELEMENTS	;
3.1 Derivation of the Phase-Curvature Relationship	;
3.2 Numerical Analysis 20)
3.2.1 Relative Phase Error between TE and TM Incidences 23	;
3.2.2 Rotation of Target with respect to Incidence Direction	ŀ
3.2.3 Canting 21	ŀ
3.3 Discussion of Second Order Corrections to the Physical Optics Approximation	+

		<u>.</u>	PAGE
CHAPTER	-	THE HF CURVATURE RECOVERY MODEL AND THE TRANSFORMATION INVARIANTS OF THE SCATTERING MATRIX	. 28
4.1	Rela	tion to the Transformation Invariants of [S]	. 28
4.2	The I	HF Scattering Ratio	. 29
4.3	Inte	rpretation of the HF Phase Sum $(\phi_{AA} + \phi_{BB})$. 31
4.4	Nume	rical Analysis	. 31
4.5	Trans	sformation to Circular Polarization Basis Vectors	. 32
4.6	Curva	ature Recovery from the Circular Polarization [C]	. 38
4.7	Trans	sformation Invariants of the Scattering Matrix	. 40
4.8	An Ir	nterpretation of the Scattering Ratio	. 43
CHAPTER	V 1	NUMERICAL VERIFICATION	. 47
5.1	Data	Description	. 47
5.2	Direc	ct Verification of the Phase-Curvature Relationship	. 48
5.2	2.1	Theoretical Data	. 48
5.2	2.2	Experimental Data	. 48
5.3	Veri	fication of the Scattering Ratio D	. 50
5.	3.1	Theoretical Data	. 50
5.	3.2	Experimental Data	. 51
CHAPTER	VI (CONCLUSION	. 91
APPENDIX	X 1 1	DERIVATIONS OF EXPRESSIONS FOR PRINCIPAL CURVATURES FOR PROLATE SPHEROIDS	. 94
APPENDI	x 11 (COMPUTER PROGRAMS	100
APPENDI	x 111	THEORETICAL DATA	109
APPENDI	x IV	EXPERIMENTAL DATA	114
REFEREN	CES .		139
RESUME			143
M.S. TH	ESIS	DEFENSE ANNOUNCEMENT	145

LIST OF SYMBOLS

```
polarization angle defined in Figure 3.1
\alpha
                         angle defined in the general form of [T]
ģ
\nabla
                         the del operator
                         wavelength
                         eigenvalue in (4.36)
                         an operator defined in section 2.2
                         relative phase of like polarized elements, \phi_{22} - \phi_{11}
                         in the linear polarization scattering matrix
                         aspect angle defined in Figure 3.2
                         phase angles defined in the general form of [T]
$7, $2, $3, $4
                         phase of S_{AA}
^{\dot{\phi}}\!AA
                         phase of S_{BB}
\phi_{BB}
                         phases of the elements of the scattering matrix
φ<sub>11</sub>, φ<sub>22</sub>, φ<sub>12</sub>
                         retarded time
θ
                         phase of A_{\epsilon}(k)
Ŕ
                         ratio of like polarized elements, S_{22}/S_{11} in the linear polarization scattering matrix
                         magnitude of R
\mathcal{H}
                          patch radius
                         optimum radar cross-section
omax
                          arbitrary orthonormai vectors
a<sub>1</sub>, a<sub>2</sub>
                          semi-minor axis of prolate spheroid
                          unit vector in the direction of the incident magnetic field
\mathtt{a}_{\mathtt{Hi}}
                          unit vector in the direction of receiver polarization
^{\mathrm{a}}Hr
                          normal vector to surface of scatterer
a<sub>n</sub>, n
```

•	
^a R	unit vector of R
$A(t), A_{F}(k)$	silhouette area of scatterer and its Fourier Transform
(A,B), (A,B)	linear polarization basis with orthonormal vectors A, B
Āp	vector potential due to surface current
âu	unit vector in the u-direction at specular point
â'u	unit vector in the u-direction at integration point
$\hat{\mathbf{a}}_{\mathbf{v}}$	unit vector in the v-direction at specular point
â',	unit vector in the v-direction at integration point
Ъ	semi-minor axis of prolate spheroid
С	semi-major axis of prolate spheroid
[C], [C(RL)]	circular polarization scattering matrix with basis (R,L)
c _{RR} , c _{LL} , c _{RL}	elements of [C(RL)]
c	free space velocity of electromagnetic wave
De	denominator defined in section 3.2.1
D	the scattering ratio defined in section 4.2
e -	relative phase error
<u>E</u> i	incident electric field polarization vector
E_{H}^{i}, E_{V}^{i}	components of $\underline{E}^{\hat{\mathbf{i}}}$ in the horizontal and vertical directions
E_L^i , E_R^i	components of $\underline{E}^{\hat{\mathbf{I}}}$ in the left circular and right circular directions
<u>E</u> ^S	scattered electric field polarization vector
E _L , E _R	components of $\underline{\mathtt{E}}^{\mathbf{S}}$ in the left circular and right circular directions
E, F, G	the coefficients of the first fundamental form of a surface

<u>h</u>	polarization vector
(H,V), (Ĥ,V)	linear polarization basis with unit vectors in the horizontal and vertical directions
Ħ	total magnetic field
Ħ _i	incident magnetic field
H _s	scattered magnetic field
$\vec{H}_{s(po)}$	physical optics scattered magnetic field impulse response
H _s (pol)	first order correction to physical optics far- field impulse response
Hui	component of \vec{H}_1 in the u-direction
H vi	component of \vec{H}_{i} in the v-direction
î, ĵ, k	unit Cartesian vectors
Im	imaginary part of
\vec{J}	surface current density
J po →	physical optics current
$\vec{J}_{ t pol}$	first order correction to physical optics current
\dot{f}_{ϵ}	correction current term
J _u	scalar current component in the u-direction
$J_{\mathbf{v}}$	scalar current component in the v-direction
Ř	vector k(frequency) space
k	wave number
К _u	principal curvature in the u-direction
$K_{\mathbf{v}}$	principal curvature in the v-direction
К	Gaussian curvature
L, M, N	coefficients of the second fundamental form of a surface
(R,L)	circular polarization basis in the right-eircular and left-eircular directions
[P]	power scattering matrix

ř	position vector to observation point
r'	position vector to integration point
r_{\circ}	radar range
र ि	$\vec{r} - \vec{r}'$
R	magnitude of \overrightarrow{R}
Re	real part of
$\vec{r}(u,v)$	curvilinear mapping of surface
S	surface of integration
S E	integration patch
[S],[S(A,B)],[S(AB)]	scattering matrix with general polarization basis (A,B)
s ₁₁ , s ₂₂	like polarized elements of scattering matrix [S]
s ₁₂ , s ₂₁	cross-polarized elements of scattering matrix [S]
S _{AA} , S _{BB} , S _{AB}	elements of scattering matrix with polarization basis (A,B)
[T], [T(RL;HV)]	unitary matrix which transforms linear polarization with basis (H,V) to circular polarization with basis (R,L)
Tr	trace of a square matrix
t	time
u, v	curvilinear coordinates of surface
x, y, z	Cartesian coordinates of surface

LIST OF FIGURES

FIGURE	PAGE
2.1	Geometry for the Derivation of the Space-Time Integral Equation
3.1	Specular Point Coordinate System 16
3.2	Incidence for Equatorial Specular Point 22
4.1	Incidence and Scattered Coordinate Systems 36
5.1(a)	Right-hand Side of Equation (3.12) versus k 52
5.1(Ь)	Convergence of k Times Re $\{(1 - \mathring{R})/(1 + \mathring{R})\}$ 53
5.1(c) 5.1(d)	Curvature Recovery from Im $\{(1 - \hat{R})/(1 + \hat{R})\}$ 5^h The Scattering Chart from Theoretical Data 55
5.2(a)(i)	Experimental Version of Figure 5.1(d) (Broadside Incidence, 2-4 GHz)
5.2(a)(ii)	Experimental Version of Figure 5.1(b) (Broadside Incidence, 2-4 GHz)
5.2(a)(iii)	Experimental Version of Figure 5.1(c) (Broadside Incidence, 2-4 GHz)
5.2(b)(i)	Experimental Version of Figure 5.1(d) (Broadside Incidence, 4-8 GHz)
5.2(b)(ii)	Experimental Version of Figure 5.1(b) (Broadside Incidence, 4-8 GHz)
5.2(b)(iii)	Experimental Version of Figure 5.1(c) (Broadside Incidence, 4-8 GHz)
5.3(a)(i)	Experimental Version of Figure 5.1(d) (Nose-on Incidence, 2-4 GHz)
5.3(a)(ii)	Experimental Version of Figure 5.1(b) (Nose-on Incidence, 2-4 GHz)
5.3(a)(iii)	Experimental Version of Figure 5.1(c) (Nose-on Incidence, 2-4 GHz)

FIGURE		PAGE
5.3(b)(i)	Experimental Version of Figure 5.1(d) (Nose-on Incidence, 4-8 GHz)	65
5.3(b)(ii)	Experimental Version of Figure 5.1(b)	
	(Nose-on Incidence, 4-8 GHz)	. 66
5.3(b)(iii)	Experimental Version of Figure 5.1(c) (Nose-on Incidence, 4-8 GHz)	. 67
5.4(a)(i)	Experimental Version of Figure 5.1(d) (45 degree Incidence, 2-4 GHz)	. 68
5.4(a)(ii)	Experimental Version of Figure 5.1(b) (45 degree Incidence, 2-4 GHz)	. 69
5.4(a)(iii)	Experimental Version of Figure 5.1(c) (45 degree Incidence, 2-4 GHz)	
5.4(b)(i)	Experimental Version of Figure 5.1(d) (45 degree Incidence, 4-8 GHz)	. 71
5.4(b)(ii)	Experimental Version of Figure 5.1(b) (45 degree Incidence, 4~8 GHz)	. 72
5.4(b)(iii)	Experimental Version of Figure 5.1(c) (45 degree Incidence, 4~8 GHz)	. 73
5.5	Effects of Relative Phase Error on	
(i)	the Scattering Chart from Theoretical Data	
(ii)	k times the Real Part, k Re $\{(1 - \overset{\circ}{R})/(1 + \overset{\circ}{R})\}$.	
(;;;)	k times the Imag. Part, k Im $\{(1 - \tilde{R})/(1 + \tilde{R})\}$	
5.6(a)	Complex Plot of D from Theoretical Data	. 77
5.6(b)	Amplitude Plot of D from Theoretical Data	. 78
5.7(a)	Experimental Version of Figure 5.6(a) (Broadside Incidence, 2-4 GHz)	. 79
5.7(b)	Experimental Version of Figure 5.6(b) (Broadside Incidence, 2-4 GHz)	. 80
5.8(a)	Experimental Version of Figure 5.6(a) (Nose-on Incidence, 2-4 GHz)	. 81

FIGURE		PAGE
5.8(b)	Experimental Version of Figure 5.6(b) (Nose-on Incidence, 2-4 GHz)	. 82
5.9(a)	Experimental Version of Figure 5.6(a) (45 degree Incidence, 2-4 GHz)	. 83
5.9(b)	Experimental Version of Figure 5.6(b) (45 degree Incidence, 2-4 GHz)	. 84
5.10(a)	Experimental Version of Figure 5.6(a) (Broadside Incidence, 4-8 GHz)	. 85
5.10(b)	Experimental Version of Figure 5.6(b) (Broadside Incidence, 4-8 GHz)	. 86
5.11(a)	Experimental Version of Figure 5.6(a) (Nose-on Incidence, 4-8 GHz)	. 87
5.11(ь)	Experimental Version of Figure 5.6(b) (Nose-on Incidence, 4-8 GHz)	. 88
5.12(a)	Experimental Version of Figure 5.6(a) (45 degree Incidence, 4-8 GHz)	. 89
5.12(b)	Experimental Version of Figure 5.6(b) (45 degree Incidence, 4-8 GHz)	. 90
A(1)	Geometry of Prolate Spheroid for Curvature Calculation	. 98

CHAPTER I

INTRODUCTION AND LITERATURE REVIEW

1.1 Introduction

When a conducting object is illuminated by electromagnetic radiation, in general, radiation is scattered in all directions by the object. The problem of direct scattering is that of determining the scattered field in all directions when the properties of the incident field as well as those of the object are known. The problem of far-field inverse scattering is defined as that of extracting the geometrical properties and/or reconstructing the shape of the scatterer under interrogation, given the incident field within the neighborhood of the scatterer and the scattered far field. This problem is fundamental to problems such as radar target classification, discrimination and identification in remote sensing and surveillance (Boerner, 1978, 1980).

It has been demonstrated that an electromagnetic scatterer acts as a sensitive polarization transformer, depending on its profile. Thus, depolarization effects must be taken into account in this problem of vector nature. Sinclair (1948) introduced the scattering matrix of a radar target for complete depolarization characterization. The question then arises as to what geometrical properties of the surface profile of the target may be extracted from the scattering matrix of the target. A study of differential geometry reveals that for a smooth, convex shape, there exists a pair of lines of principal curvatures orthogonal to and intersecting with each other at any specular point on the scatterer's surface. Since the principal

curvatures at a point totally determine any other normal curvatures at the same point, these two principal curvatures serve as a complete curvature characterization at the specular point of the scatterer. It will be shown in this thesis that if the polarization of the incident magnetic field is in one of the directions of the principal curvatures at the specular point of interest, then the cross-polarized backscattered returns vanish. This kind of null polarizations generates the idea that the principal curvatures are possibly related to the non-zero co-polarized backscattered returns, and it also generated many recent studies in radar polarimetry. This specific degeneracy of target depolarization phenomenology provides the initial motivation for writing this thesis.

1.2 Literature Review

Inverse techniques cover a vast amount of literature and they have been developed in many otherwise diverse fields of physical sciences where the characteristics of a medium are estimated from experimental data, obtained from measurements made usually at a distance from the medium, utilizing the laws that relate these characteristics to the experimental data in a given situation. The inverse problem of electromagnetic scattering has not been solved in the general case. Stringent requirements are often needed to be kept on the shapes of the scatterer to be recovered and on the operating frequency range. Moreover, the existing solutions generally demand an exhaustive amount of input information of as many aspects and frequencies as possible. There are many approaches for obtaining

approximate solution to the electromagnetic inverse scattering problem. The following only highlights the ramp response approach and other specific approaches which have been proven successful and potentially promising for future research, and which are directly relevant to this thesis. An excellent and quite complete review of electromagnetic inverse problems is given in Boerner [21].

Kennaugh, Cosgriff and Moffatt have introduced the use of the impulse response in electromagnetic scattering problems [10,12]. Using the physical optics approximation, it was shown that the impulse response of a smooth conducting object is directly proportional to the second derivative of the cross-sectional area of the scatterer. This remarkable high frequency inverse scattering identity is known as the Kennaugh-Cosqriff formula [10]. The area profile can thus be recovered from the ramp response, since the ramp response is then directly proportional to the cross-sectional area rather than its derivative (Kennaugh and Moffatt, 1965). Young [22] used the ramp response synthesized from complex scattering data at ten harmonically related frequencies in the target's low resonance regime to estimate the area function, from which a "likely image" of the target was generated at three orthogonal look angles using his approximate limiting surface technique. The images obtained are decent, but in general they are not uniquely specified since more than one shape satisfies any three look-angle area function set.

A more systematic approach was carried out by Das and Boerner, who showed that the reconstruction of a smooth conducting target, convex in shape, can be considered as a two-step process: (i) an electromagnetic step of obtaining suitable radar measurables from which one can make an estimation of a geometrical function of the conducting scatterer; and

(ii)a geometrical step of reconstructing the shape and size of the object from the knowledge of the geometrical function estimated in the first step. With the Radon transform concept, they showed that the inverse three-dimensional Radon transform of the cross-sectional area of a scatterer is simply the target's characteristic function, which is a complete specification of shape. The area function can, as in the previous approach, be estimated by the ramp response method (Boerner, 1980). Thus, the electromagnetic inverse scattering problem can be formulated as the classical Radon problem. Moreover, they also indicated that the classical Radon problem is intimately related to the problem of reconstruction from projections which has long been investigated and applied in diverse fields, particularly in Computer-Assisted Tomography. Thus, the many reconstruction techniques and algorithms well developed in other fields can be applied in radar target imaging as well.

A solution for the inverse scattering problem using the spacetime Integral equation was reported by Bennett et al [23]. The technique developed was iterative and restricted to the class of rotationally symmetric conducting targets. In this approach, the inverse problem is formulated as an inversion of the space-time integral equation. The shapes reconstructed are excellent for this simple class of shapes, but no extension to more general shapes has been made recently.

An important aspect of electromagnetic inverse scattering is to incorporate the problem with utilization of polarization. A

monostatic inverse scattering model based on polarization utilization was developed by Chaudhuri and Boerner [2,17,24]. In brief, approximation was made of high frequency scattered fields and an equivalent ellipsoidal model was developed. The utilization of the space-time integral equation and Minkowski's theory lead to a system of equations for the recovery of the surface profile. The numerical technique is iterative, and when the curvature difference at the specular point approaches to zero, the recovery of that particular point is not possible; the system of the recovery equations also becomes ill-conditioned even if the curvature difference is small.

Based on the first order polarization correction to physical optics [8], Ho graphically reconstructed the shape of a sphere-capped cylinder with polarization correction incorporated into the Radon transform approach [25,26]. The results show that the quality of images are significantly improved with polarization correction. It must be noted here that Ho took advantage of the plane symmetry of the sphere-capped cylinder, and reduced the Radon transform by one dimension. In this class of objects, the two-dimensional inverse transform of the area function normal to the equatorial plane gives the width perpendicular to the plane. The width over this plane of symmetry is actually a complete specification of the shape. Since in the two-dimensional case, the Radon transform becomes the projection, thus the filtered back projection algorithm was directly borrowed from the theory of reconstruction from projections (Shepp and Logan, 1974).

In this thesis, a high frequency inverse scattering model is

developed for the recovery of the specular point curvature from polarimetric scattering data. Not only does the model show that the specular geometry can be directly extracted from polarimetric data, but it also contributes in viewing the electromagnetic inverse scattering problem as one of reconstruction from curvatures in differential geometry. In the introduction of the next chapter, the main objective of this thesis will be clearly specified.

CHAPTER II

THE SPACE-TIME INTEGRAL EQUATION AND FIRST ORDER CORRECTION
TO PHYSICAL OPTICS

2.1 Introduction

In recent years, due to advances made in radar technology, it has become possible to measure the complete relative phase scattering matrix of an object reliably [1,2,3]. Thus the utilization of these polarimetric scattering data in radar target identification or discrimination and in other inverse scattering applications has become of considerable interest in current theoretical and experimental research efforts [1-7]. The main objective of this thesis is to investigate the information content of the scattering matrix [S], on the shape, size, curvature, etc. of a scatterer, when it is given for the monostatic case and in the high frequency region (i.e. the wavelength of the interrogating signal is small compared to the object characteristic dimension).

The scattering matrix [S], which manifests total polarization information for a fixed frequency and a given aspect, is comprised of four measurable complex elements (four magnitudes and four phases). It will be shown here that the difference in suitable phase terms in this matrix, under the high frequency interrogation conditions, can lead to the recovery of the difference in principal curvatures at the specular point, from a given general [S¹ matrix. This procedure avoids unitary transformations used in the pursuit of cross-polarization nulls required in certain radar target identification techniques [3-5]. The underlying concept used to achieve the above results is based on Bennett's first order far-field polarization correction [8,9]

to the Kennaugh-Cosgriff's physical optics formula for the electromagnetic backscattered field [10,11].

In this chapter, the space-time integral equation, and how it is used to obtain a first order correction to physical optics, is discussed so that later, in Chapter III, its possible extension to obtain higher order correction terms can be presented clearly. The relationship between the principal curvatures and the general [S] matrix is also developed in Chapter III. A discussion of numerical verification with theoretical as well as measured data is given in Chapter V.

2.2 The Space-Time Integral Equation

An electromagnetic wave incident on a perfectly conducting body induces currents on the surface of the scatterer, which in turn radiate and produce the scattered field. The current distribution produces a vector potential given by

$$\vec{A}_p(\vec{r},t) = \frac{1}{4\pi} \iint_S \frac{\vec{J}(\vec{r},\tau)}{R} dS$$
,

where $\vec{J}(\vec{r},t)$ is the induced surface current density at time t, \vec{r} is the position vector to the observation point, \vec{r}' is that to an integration point, $R = |\vec{r} - \vec{r}'|$, $\tau = t - R/c$. The geometry is illustrated in Figure 2.1.

The total magnetic field \vec{H} is equal to the sum of the incident field \vec{H}_i and the scattered field \vec{H}_s due to \vec{J} ,

$$\vec{H}(\vec{r},t) = \vec{H}_{i}(\vec{r},t) + \vec{H}_{s}(\vec{r},t)$$

where $\vec{H}_s(\vec{r},t) = \nabla \times \vec{A}_p$

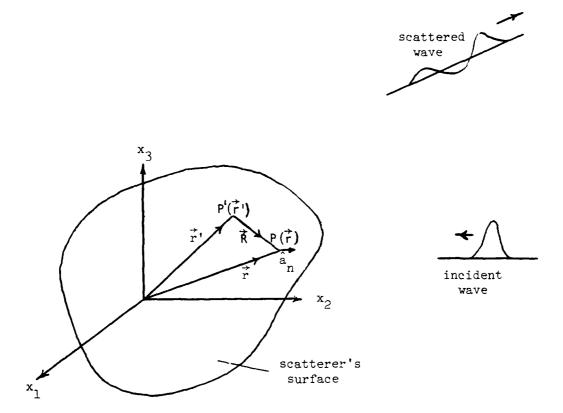


Figure 2.1 Geometry for the Derivation of the Space-Time Integral Equation

(Redrawn after Bennett, "Time Domain Solution Via Integral Equations--- Surfaces and Composite Bodies" July 1979)

$$= \frac{1}{4\pi} \mathcal{M}_{S} L(\vec{J}(\vec{r}',\tau)) \times \hat{a}_{R} dS' [16]$$

The operator L is defined as $1/R^2 + (1/Rc) \Im / \Im \tau$, and \hat{a}_R is the unit vector of \vec{R} . An expression for \vec{J} can be obtained by shrinking the observation point to a point on the scatterer's surface under a limiting procedure. For the case of a perfectly conducting scatterer, the H-field boundary condition $(\vec{J} = \hat{a}_n \times \vec{H})$, with \hat{a}_n being the unit outward normal vector to the surface, can then be invoked to yield \vec{a} vector integral equation as shown in Bennett [23]

$$\vec{J}(\vec{r},t) = 2\hat{a}_n \times \vec{H}_i(\vec{r},t) + \frac{1}{2\pi} II_s \hat{a}_n \times (L\vec{J}(\vec{r}',\tau) \times \hat{a}_R) dS' (2.1)$$

The first term on the right-hand side of (2.1) is identified a, the physical optics approximation and is also the source term, whereas the second integral represents the contribution of retarded currents at points on the scatterer's surface other than the observation point $P(\vec{r})$ i.e.

$$\vec{J}(\vec{r},t) = \vec{J}_{po}(\vec{r},t) + \vec{J}_{\varepsilon}(\vec{r},t),$$
with $\vec{J}_{\varepsilon}(\vec{r},t) = \frac{1}{2\pi} \iint_{\varepsilon} \hat{a}_{n} \times \{L\vec{J}(\vec{r}',\tau) \times \hat{a}_{R}\} dS'$ (2.2)

being the correction to the physical optics current \vec{J}_{po} . For the far scattered magnetic field (\vec{H}_s) , with R \rightarrow r, it can be shown that [8]

$$\mathbf{r}_{O}\overset{\rightarrow}{\mathbf{H}}_{S}(\overset{\rightarrow}{\mathbf{r}},t) = \frac{1}{\mathbf{H}_{T}} \iint_{S} \left[\frac{\partial}{\partial \tau} \overset{\rightarrow}{\mathbf{r}}(\overset{\rightarrow}{\mathbf{r}}',\tau) \right] \times \hat{\mathbf{a}}_{r} dS'$$
 (2.3)

where r_0 is the radar range. Using (2.3), the physical optics approximation for the far scattered impulse response field was derived [8,23] and is equivalent to the Kennaugh-Cosgriff formula [10-12] which can be written as

$$r_{o}H_{s(po)}(\vec{r},t) = \frac{1}{2\pi} \frac{d^{2}}{dt^{2}} A(t) \hat{a}_{H_{i}}$$
 (2.4)

where a_{H_i} is the unit vector in the direction of \vec{H}_i , and A(t) is the silhouette area of the scatterer as delineated by the incident impulsive plane wavefront moving at half the free space velocity, c. Discussion here is restricted to the illuminated side of a smooth, conducting, convex object.

Since the impulse response given in (2.4) depends solely on the area function, it manifests no depolarization effects. Depolarization effects were taken care of by Bennett et al [8,9] in their first order correction to physical optics approximation.

2.3 A First Order Correction to Physical Optics Using the Space-Time Integral Equation [8]

To obtain an analytic expression for the first order correction to the physical optics far field, the correction to the induced surface current needs to be considered first. The integral in (2.2) cannot be handled analytically without the knowledge of the geometry of the scatterer. Yet the integration can be carried out over a patch, \mathbf{S}_{ε} , around the specular point, assuming the patch being so small that it is virtually flat. The contributions of retarded currents outside the patch and effects such as creeping waves are ignored to avoid the total surface integration. Under this "leading edge" simplification, the first order correction obtained is valid and accurate towards the high frequency end of the phasor frequency response.

The expansion of the vector triple cross-product in (2.2) yields the first order correction current

$$\begin{split} \vec{J}_{po1}(\vec{r},t) &= \frac{1}{2\pi} \, \text{ff}_{S_{\epsilon}}[LJ_{u}(\vec{r}',\tau)][\,(\hat{a}_{n} \, . \, \, \hat{a}_{R})\hat{a}_{u}' \, - \, (\hat{a}_{n} \, . \, \, \hat{a}_{u}')\hat{a}_{R}] \, \, dS' \\ &+ \frac{1}{2\pi} \, \text{ff}_{S_{\epsilon}}[LJ_{v}(\vec{r}',\tau)][\,(\hat{a}_{n} \, . \, \, \hat{a}_{R})\hat{a}_{v}' \, - \, (\hat{a}_{n} \, . \, \, \hat{a}_{v}')\hat{a}_{R}] \, \, dS' \end{split}$$

where the primed quantities are associated with the integration point P'(\vec{r} ') and \hat{a}_u , \hat{a}_v are the unit vectors tangent to the principal curves at the point of interest. The geometry information embedded in \hat{a}_n . \hat{a}_R can be extracted by expressing $\hat{a}_R = (\vec{r} - \vec{r}')/R$ in a Taylor series expansion as follows:

$$\vec{r}' - \vec{r} = \vec{r}_{u} \Delta u + \vec{r}_{v} \Delta v$$

$$+ \frac{1}{2!} \left[\vec{r}_{uu} (\Delta u)^{2} + 2 \vec{r}_{uv} (\Delta u) (\Delta v) + \vec{r}_{vv} (\Delta v)^{2} \right]$$

$$+ \frac{1}{3!} \left[\vec{r}_{uuu} (\Delta u)^{3} + 3 \vec{r}_{uuv} (\Delta u)^{2} (\Delta v) + 3 \vec{r}_{uvv} (\Delta u) (\Delta v)^{2} + \vec{r}_{vvv} (\Delta v)^{3} \right]$$

+ ,..,

where $\vec{r}_u = \frac{\partial \vec{r}}{\partial u}$, etc. The above series describes any neighboring point $(\Delta u, \Delta v)$ in the vicinity of \vec{r} in terms of the derivatives of \vec{r} at the specular point. The geometry of the small patch is thus extrapolated in terms of the properties of the surface at the specular point.

The scalar product of \hat{a}_n and \hat{a}_R expressed by a series truncated at second order (i.e. terms $(\Delta u)^2$ and $(\Delta v)^2$, etc.) introduces (E, F, G) and (L, M, N), the coefficients of the first and second fundamental forms of the surface $\vec{r}(u,v)$ [13,14]. To simplify the algebra, the principal curves are chosen as the parametric curves to represent the curvilinear mapping $\vec{r}(u,v)$ for the surface of the patch, thus

forcing F and M to zero [8,13,14]. The principal curvatures along \hat{a}_u and \hat{a}_v at the specular point are obtained as [13]

$$K_u = \frac{L}{E}$$

$$K_{V} = \frac{N}{G}$$
,

respectively. Finally, $LJ_u(\vec{r}',\tau)$ and $LJ_v(\vec{r}',\tau)$ can be approximated as $J_u(\vec{r},t)/R^2$ and $J_v(\vec{r},t)/R^2$, respectively, by assuming that the currents are spatially constant on a small, flat, circular patch of radius ε_0 . With the procedure outlined above the analytic expression arrived at in [8] is

$$\vec{J}_{\varepsilon}(\vec{r},t) = [\hat{a}_{u}J_{u}(\vec{r},t) - \hat{a}_{v}J_{v}(\vec{r},t)] \frac{K_{u} - K_{v}}{4} \varepsilon_{o}$$
 (2.5)

The corresponding first order far-field impulse response correction was obtained by assuming physical optics currents for J_u and J_v in the above equation and then substituting it into (2.3). A crucial assumption made in [8] is that the patch radius ε_0 increases linearly with time t, spreading from the specular point at the leading edge. The final expression for the first order correction to the scattered far field is

$$r_{oH_{s(po1)}}(\vec{r},t) = \frac{K_{u} - K_{v}}{4 \pi} \cdot [\hat{a}_{uH_{ui}} - \hat{a}_{vH_{vi}}] \frac{dA}{dt},$$
 (2.6)

where H_{ui} and H_{vi} are the components of \hat{H}_{i} in the directions of \hat{a}_{u} and \hat{a}_{v} , respectively.

It is clear that the first order correction exhibits depolarization effects, which are proportional to the difference in the principal curvatures at the specular point. Moreover, the

first order correction takes the functional form of the first derivative of the silhouette area function A(t), whereas the physical optics far field, which exhibits no depolarization effects, takes the functional form of the second derivative of A(t). The practical aspects of using this first order correction in geometry extraction (curvature difference at the specular point in this case) are analyzed in the next chapter.

CHAPTER III

CURVATURE RECOVERY FROM HIGH FREQUENCY [S] MATRIX ELEMENTS

3.1 Derivation of the Phase-Curvature Relationship

The polarimetric scattering data measured with a monostatic radar system are given by

$$[S] = \begin{bmatrix} s_{11} & s_{12} \\ s_{21} & s_{22} \end{bmatrix} = \begin{bmatrix} |s_{11}|e^{j\phi}11 & |s_{12}|e^{j\phi}12 \\ |s_{21}|e^{j\phi}21 & |s_{22}|e^{j\phi}22 \end{bmatrix}$$

The general polarization geometry with respect to the principal directions at the specular point is shown in Figure 3.1. The elements S_{11} and S_{22} represent the backscattered signals when the transmitted and the received polarizations are identical, i.e. \hat{a}_1 and \hat{a}_2 , respectively. On the other hand, S_{12} and S_{21} represent the cases when the transmitted and the received polarizations are orthogonal to each other (transmit- \hat{a}_1 , receive- $\hat{a}_2 \rightarrow S_{21}$, etc.). In order to relate the measurable [S] matrix to the theory presented in Chapter II, the total physical optics scattered far field (i.e. the physical optics and first order correction) is transformed from the time (t) domain to the frequency (k, the wave number) domain by using Fourier transformation. In the time domain, combining (2.4) and (2.6), the total high frequency scattered far field is

$$r_{o}\vec{H}_{s}(\vec{r},t) = \frac{1}{2\pi} \frac{d^{2}}{dt^{2}} A(t) \hat{a}_{H_{i}} + \frac{K_{u} - K_{v}}{4\pi} \cdot \frac{d}{dt} A(t)$$

$$\cdot [(\hat{a}_{H_{i}} \cdot \hat{a}_{u}) \hat{a}_{u} - (\hat{a}_{H_{i}} \cdot \hat{a}_{v}) \hat{a}_{v}] \qquad (3.1)$$

The Fourier transform of (3.1), with the initial condition

$$A_F(0) = \{F.T.[A(t)]\}_{k=0} = 0,$$

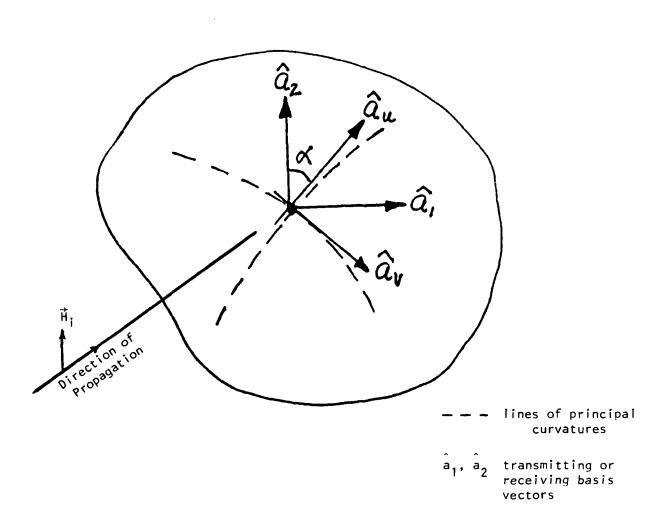


Figure 3.1 Specular Point Coordinate Systems

yields

$$r_0 \hat{H}_s(\vec{r}, k) \approx \frac{1}{2 \pi} (jk)^2 A_F(k) \hat{a}_{H_i}$$

+ $(jk) A_F(k) \frac{K_u - K_v}{4 \pi} [(\hat{a}_{H_i} \cdot \hat{a}_u) \hat{a}_u - (\hat{a}_{H_i} \cdot \hat{a}_v) \hat{a}_v]$ (3.2)

where $A_F(k) = F.T.[A(t)].$

From the geometry in Figure 3.1, it is seen that

$$\hat{a}_{1} = \sin \alpha \, \hat{a}_{u} + \cos \alpha \, \hat{a}_{v} \quad \text{and}$$

$$\hat{a}_{2} = \cos \alpha \, \hat{a}_{u} - \sin \alpha \, \hat{a}_{v}$$
(3.3)

For measuring \hat{s}_{11} the transmitter polarization becomes $\hat{a}_{H_1} = \hat{a}_1$, and the receiver polarization becomes $\hat{a}_{H_1} = \hat{a}_1$. Thus, using (3.2), one has

$$S_{11} = \hat{a}_{Hr}. (r_0 \vec{h}_s)$$

$$= \frac{1}{2\pi} (jk)^2 A_F(k) (\hat{a}_1 \cdot \hat{a}_1)$$

$$+ (jk) A_F(k) \frac{K_U - K_V}{4\pi} [(\hat{a}_1 \cdot \hat{a}_U) (\hat{a}_1 \cdot \hat{a}_U)$$

$$- (\hat{a}_1 \cdot \hat{a}_V) (\hat{a}_1 \cdot \hat{a}_V)]$$
(3.4)

Similarly, for S_{22} , $\hat{a}_{H_1} = \hat{a}_2$, $\hat{a}_{Hr} = \hat{a}_2$; for S_{21} , $\hat{a}_{H_1} = \hat{a}_1$, $\hat{a}_{Hr} = \hat{a}_2$; and for S_{12} , $\hat{a}_{H_1} = \hat{a}_2$, $\hat{a}_{Hr} = \hat{a}_1$. Now using (3.2), (3.3) and (3.4), one gets (ignoring scale factors)

$$S_{11} = \frac{1}{2\pi} (jk)^2 A_F(k) - (jk) A_F(k) \frac{K_u - K_v}{4\pi} \cos 2\alpha$$
 (3.5)

$$S_{22} = \frac{1}{2\pi} (jk)^2 A_F(k) + (jk) A_F(k) \frac{K_U - K_V}{4\pi} \cos 2\alpha$$
 (3.6)

$$S_{21} = (jk)A_{F}(k) \frac{K_{U} - K_{V}}{4\pi} \sin 2\alpha$$
 (3.7)

$$S_{21} = S_{12}$$
 (reciprocity satisfied)

Using (3.5) and (3.6), it can be shown that

$$\frac{K_{u} - K_{v}}{2} = -j \frac{k}{\cos 2\alpha} \frac{S_{11} - S_{22}}{S_{11} + S_{22}}$$

$$= -j \frac{k}{\cos 2\alpha} \frac{1 - R}{1 + R}$$
(3.8)

where $R = Re^{j\phi}d$

$$= \frac{s_{22}}{s_{11}}$$

$$= \frac{|s_{22}|}{|s_{11}|} e^{j(\phi_{22} - \phi_{11})}$$

It is clear that in order for the equality in (3.8) to hold true and therefore represent a physical case in which $(K_U^- - K_V^-)$ is a real number, one needs

Re
$$\frac{1 - \frac{\hat{R}}{R}}{1 + \frac{\hat{R}}{R}} = 0$$
, implying $1 - R^2 = 0$ (3.9)

where Re stands for real part.

Thus for (3.8) to represent a physical situation, the condition required is

$$R = \frac{|s_{22}|}{|s_{11}|} = 1$$
, implying $|s_{22}| = |s_{11}|$

From electromagnetic scattering theory, it is known that the above condition is attained at relatively high frequencies (i.e. physical optics to geometrical optics region). It is interesting to note that algebraic manipulation of (3.8) independently points out that

it is a high frequency formula which is, of course, true since the physical optics approximations for the scattered fields are being used.

With the condition (3.9), now (3.8) can be written as

$$\frac{K_u - K_v}{2} = -\frac{k}{\cos 2\alpha} \tan(\phi_d/2) \tag{3.10}$$

Where $\phi_d = \phi_{22} - \phi_{11}$.

In the rest of the text, the above expression will be referred to as the "phase-curvature" relationship.

For the inverse scattering applications α is an unknown quantity (Figure 3.1). Thus, before (3.10) can be used to recover the curvature difference at the specular point, α needs to be determined. For this purpose, consider

$$\frac{S_{21}}{S_{11}} = \frac{\frac{K_{u} - K_{v}}{2} \sin 2\alpha}{(jk) - \frac{K_{u} - K_{v}}{2} \cos 2\alpha}$$

from (3.5) and (3.7). Using (3.10), one obtains

$$\frac{S_{21}}{S_{11}} = -\frac{(\tan 2\alpha) (\tan (\phi_d/2))}{\tan (\phi_d/2) + j}$$

$$\alpha = \frac{1}{2} \tan^{-1} \left\{ \frac{-S_{21}}{S_{11}} (1 + j \cot (\phi_d/2)) \right\}$$
(3.11)

Once again for α to be a real angle (and therefore representing the given physical situation) one needs

$$\frac{S_{21}}{S_{11}} = D_0 (1 - j \cot (\phi_d/2))$$

where $\mathbf{D}_{\mathbf{O}}$ is a real number.

Using (3.11), without applying any unitary transformation, the cross-polarized nulls of a given scattering matrix are known.

Two special cases of (3.10) are when α = 0, or $\pi/2$ (i.e. the incident linear polarization coincides with one of the directions of the principal curvatures at the specular point). For these special cases, there is no depolarization of the energy in the backscattered direction ($S_{21} = S_{12} = 0$). The corresponding [S] matrix is referred to as the cross-pol. null scattering matrix. The phase-curvature relationship becomes

$$\frac{K_u - K_v}{2} = \frac{1}{k} \tan \frac{\phi_{22} - \phi_{11}}{2}$$
 (3.12)

In chapter V, the validity of (3.12) will be numerically tested with both theoretical and experimental scattering matrix data.

3.2 Numerical Analysis

In this section, a numerical analysis of applying the phase-curvature relationship on the polarimetric scattering matrix of a prolate spheroidal scatterer is presented. The 2:1 prolate spheroid was used as a test case because of its well-defined finite curvature difference at any point on the surface and also because the theoretical as well as experimental data over a large range of frequencies were readily available for this object. For the time being, since both theoretical and experimental data are available only for the special cases in which the incident polarization coincides with one of the directions of the principal curvatures at the specular points (i.e. points on the equator of the prolate

spheroid (Figure 3.2)), equation (3.12) rather than the more general phase-curvature relationship (3.10) is directly tested. In these cases, the incident polarizations are along the vertical and horizontal directions. To verify the theory of curvature recovery, it is essential to check

- (i) whether the right-hand side of (3.12) will approach the actual value of $(K_u K_v)/2$ for a given equatorial specular point as k increases;
- (ii) whether k Re { $(1 \tilde{R})/(1 + \tilde{R})$ } will tend to zero despite that k increases;
- (iii) whether the imaginary part of $k \{ (1 \hat{R})/(1 + \hat{R}) \}$ will settle to the constant value of $(K_{_{\mathbf{U}}} K_{_{\mathbf{V}}})/2;$

A more interesting and compact presentation of results is to plot the right-hand side of (3.8) on a complex plane (i.e. the imaginary part versus the real part of $\{k(1-\tilde{R})/(1+\tilde{R})\}$. It is predicted that this scattering chart will be a spiral which, as frequency increases, will converge to (or hover around) a point on the imaginary axis. The distance of this point on the imaginary axis from the origin will be equal to the required value of $(K_u - K_v)/2$ for the specular point of interest. It is expected that this complex plane plot will be particularly useful when the input data are not very accurate and are noisy (measurement data). In the above tests, the exact value of the curvature difference $K_u - K_v$ has been calculated by using Minkowski's support function for ellipsoidal surfaces [17]. Another way is through the use of differential geometry (Appendix I). The value of k is normalized with respect

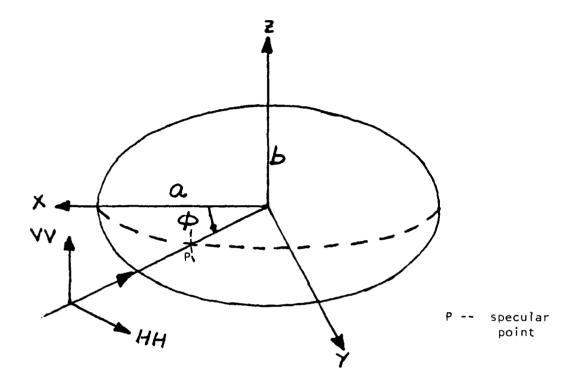


Figure 3.2 Incidence for Equatorial Specular Point

to the length of the semi-minor axis of the prolate spheroid.

It is useful to consider the following types of measurement errors for the interpretation of the numerical results and the explanation of the deviation from theoretical prediction given in chapter V.

3.2.1 Relative Phase Error between TE and TM Incidences

If the target is not illuminated simultaneously with TE (vertical) and TM (horizontal) polarizations, then even an offset of one millimeter between the separate positions of the same antenna along the direction of incidence will cause phase distortion of about ten degrees to the phase difference ($\phi_{vv} - \phi_{hh}$) at a frequency of 4 GHz. If, however, the target is slightly displaced in the direction of propagation with the TE and TM measurements simultaneously made, then both ϕ_{vv} and ϕ_{hh} are distorted to the same extent. Thus the relative phase difference and the subsequent calculations are not affected at all.

To investigate how the scattering chart is affected assuming that a relative phase error of e radians does occur, the complex ratio $(1 - \tilde{R})/(1 + \tilde{R})$ can be broken down into its real and imaginary parts, with ϕ_d replaced by $(\phi_d + e)$.

Re
$$\frac{1 - \frac{\gamma}{R}}{1 + \frac{\gamma}{R}} = \frac{1 - R^2}{De}$$

Im $\frac{1 - \frac{\gamma}{R}}{1 + \frac{\gamma}{R}} = \frac{-2R \sin(\phi_d + e)}{De}$

where $De = 1 + R^2 + 2R \cos (\phi_d + e)$. Note that a spheroidal wave function expansion could be used for error estimation.

It is obvious that the real part is much more resistant to

changes caused by the error e than the imaginary part. Thus the spiral will become a helix elongating mainly along the imaginary axis and away from the exact value of $(K_u - K_v)/2$. In Chapter V, the relative phase error will be simulated within the set of theoretical data.

3.2.2 Rotation of Target with respect to Incidence Direction

If the target is rotated with respect to the direction of incidence (i.e. the incident polarizations are not along the directions of principal curvatures), then it is the general phase-curvature relationship (3.10) which should be tested instead. Hence, a multiplicative factor of cos 2α (Figure 3.1) will account for the sole effect of rotation error. The scattering chart will retain its spiral shape however.

3.2.3 Canting

If the target is slightly canted with the z-axis still being horizontal, then obviously the effect will be that which resulted from changing the aspect angle ϕ (Figure 3.2).

3.3 Discussion of Second Order Corrections to the Physical Optics Approximation

The first order correction to the physical optics approximation due to Bennett et al [8] has been shown to be proportional to the difference in principal curvatures and to have the functional form of the first derivative of the silhouette area of the target. It might be expected that more geometry of the specular point can be identified by extending the first order correction. For a general case in which $(K_U - K_V)$ is not zero, geometrical parameters, such as

derivatives or mixed derivatives of principal curvatures, torsion of the principal curves, etc., can appear in the second order correction term. Another motivation in the pursuit of the second order correction is the possible relaxing of the high frequency restriction required for the validity of the first order correction term. In order to elaborate on this conjecture, note that, since $\frac{d^2A}{dt^2}$ of the physical optics term corresponds to $(jk)^2A_F(k)$ in the frequency domain, and $\frac{dA}{dt}$ of the first order correction term corresponds to $(jk)A_F(k)$, there might be a trend that higher order corrections become more dominant (important) for lower frequencies within the high frequency realm of physical optics.

One approach to the problem of extension is to take into account the higher order derivatives of $\vec{r}(u,v)$ that were truncated in the first order correction. The first order term has been obtained by retaining the second order derivatives of \vec{r} with respect to u and v. To obtain the second order correction term, the integrand in (2.2) for \vec{J}_{ϵ} has been expanded with inclusion of the higher order surface derivatives of \vec{r} . After some algebraic manipulations, the integrand was written as a sum of several terms like \hat{a}_u , \hat{a}_v , \hat{r}_{uu} , \hat{r}_{vv} , ... (up to the third order derivatives) multiplied by the powers of Δu and/or Δv . This is expected since these vectors introduce the higher order depolarization concepts. In contrast, the first order correction term is in the directions of \hat{a}_u and \hat{a}_v explicitly.

The vectors such as \vec{r}_{uu} , \vec{r}_{uv} , etc. can be resolved in \hat{a}_u , \hat{a}_v and \hat{a}_n directions by introducing the Christoffel symbols through the use of the Gauss and Weingarten equations in differential geometry

[12,13]. These Christoffel symbols depend on the basic geometric parameters of the surface, namely, (E, F, G) and (L, M, N). All terms in the \hat{a}_n direction can be neglected on the basis of physical considerations that the induced surface current cannot have a component in the normal direction to the surface. The appearance of the \hat{a}_n term is due to the truncation of the Taylor series. The totality of all terms in the series should, in theory, nullify the current component in the \hat{a}_n direction. Since the depolarization derivative vectors, and hence the Christoffel symbols, are all evaluated at the specular point, they can be taken outside the surface integral in (2.2), which then is written as a sum of integrals of the type

$$\iint_{S_{\varepsilon}} \frac{\left(\Delta u\right)^{2}}{R^{3}} dS' , \qquad \iint_{S_{\varepsilon}} \frac{\left(\Delta v\right)^{4}}{R^{4}} dS' , \dots etc.$$

For direct scattering problems the integration may yield complicated analytic solutions despite the fact that S_{ϵ} is known. For inverse scattering problems, where S_{ϵ} is unknown, the integration may be approximated by assuming a flat patch for S_{ϵ} . Since the Taylor series dictated a small patch, leading edge effects and thus the high frequency restrictions must be adhered to. An approximate expression for \vec{J}_{ϵ} can thus be obtained by choosing a circular patch of very small radius ϵ_{o} as the most simple case. However, the expression thus obtained has been found to have a factor of ϵ_{o}^{3} in contrast to ϵ_{o} in the first order correction. The relatively small value of ϵ_{o}^{3} compared with ϵ_{o} renders the expression for the second order correction current thus obtained insignificant. The

far-field correction term corresponding to this second order current term then has very little significance in comparision to the physical optics or the first order term, and therefore is of little consequence to most practical situations.

Nevertheless, it is suggested that one may assume some simple known curved patch (instead of a flat patch), and take into account the whole vicinity of the specular point rather than only the point itself, in further pursuit of specular geometry through the backscattered leading edge returns.

CHAPTER IV

THE HF CURVATURE RECOVERY MODEL AND THE TRANSFORMATION INVARIANTS OF THE SCATTERING MATRIX

4.1 Relation to the Transformation Invariants of [S]

The scattering matrix [S(A,B)] with respect to an orthogonal basis (A,B) can be transformed to [S'(A',B')] with respect to another orthogonal basis (A',B') through a unitary transformation [18,19,20]. The transformation is invariant and satisfies [3,38]

Span
$$\{[S(A,B)]\}\ = Span \{[S'(A',B')]\}\ = invariant$$
 (4.1)

and
$$Det \{[S(A,B)]\} = Det \{[S'(A',B')]\} = invariant$$
 (4.2)

where Det stands for the determinant of the scattering matrix and the span is defined in the following.

Applying equations (3.5 - 3.7) for a basis (A,B) with polarization angle α (Figure 3.1),

Span {[S(A,B)]} =
$$|S_{AA}|^2 + |S_{BB}|^2 + 2|S_{AB}|^2$$

= $(\frac{1}{2\pi})^2 k^2 |A_F(k)|^2 |jk - \frac{K_u - K_v}{2} \cos 2\alpha|^2$
+ $(\frac{1}{2\pi})^2 k^2 |A_F(k)|^2 |jk + \frac{K_u - K_v}{2} \cos 2\alpha|^2$
+ $2(\frac{1}{2\pi})^2 k^2 |A_F(k)|^2 [\frac{K_u - K_v}{2} \sin 2\alpha]^2$
= $2(\frac{1}{2\pi})^2 k^2 |A_F(k)|^2 \{k^2 + [\frac{K_u - K_v}{2}]^2\}$, (4.3)

which is independent of α .

For a given frequency (high enough so that the theory is valid) and a given aspect, the right-hand side of equation (4.3) is indeed invariant.

Similarly,

Det
$$\{ [S(A,B)] \} = S_{AA}S_{BB} - S_{AB}^{2}$$

$$= (\frac{1}{2\pi})^{2} (jk)^{2} A_{F}^{2}(k) \{ (jk)^{2} - [\frac{K_{u} - K_{v}}{2}]^{2} (\cos^{2}\alpha)$$

$$- \sin^{2}\alpha)^{2} \} - (\frac{1}{2\pi})^{2} (jk)^{2} A_{F}^{2}(k) [\frac{K_{u} - K_{v}}{2}]^{2} \sin^{2}2\alpha$$

$$= (\frac{1}{2\pi})^{2} k^{2} A_{F}^{2}(k) \{ k^{2} + [\frac{K_{u} - K_{v}}{2}]^{2} \}$$
(4.4)

which is also invariant and independent of α .

Hence equations (4.3) and (4.4) are the high frequency versions of the invariance equations (4.1) and (4.2).

4.2 The HF Scattering Ratio

It is interesting to define D (referred to as "the HF scattering ratio" in this thesis) as the ratio of Det $\{[S(A,B)]\}$ to Span $\{[S(A,B)]\}$,

$$D = \frac{s_{AA}s_{BB} - s_{AB}^{2}}{|s_{AA}|^{2} + |s_{BB}|^{2} + 2|s_{AB}|^{2}}$$

$$= \frac{A_{F}^{2}(k)}{2|A_{F}(k)|^{2}}$$

$$= 0.5 e^{j2\theta}$$
(4.5)

where
$$\theta = \text{Arg } A_F(k) = \text{the phase of } A_F(k)$$
 (4.6)

For the case in which incidence polarization is along one of the directions of principal curvatures, $|S_{AB}|$ equals zero. Let S_{AA} be $|S_{AA}|e^{j\,\phi}AA$ and so on, thus

$$D = \frac{|S_{AA}| |S_{BB}| e^{j(\phi_{AA} + \phi_{BB})}}{|S_{AA}|^2 + |S_{BB}|^2}$$
Thus $\theta = \text{Arg } A_F(k) = \frac{\phi_{AA} + \phi_{BB}}{2}$ (4.7)

and
$$\frac{|S_{AA}| |S_{BB}|}{|S_{AA}|^2 + |S_{BB}|^2} = \frac{1}{2}$$
, which requires that $|S_{AA}| = |S_{BB}|$. This

condition is consistent with high frequency electromagnetic scattering.

Returning again to the general case in which $|S_{\Delta R}| \neq 0$,

$$\begin{split} s_{AA}s_{BB} - s_{AB}^{\ 2} &= |s_{AA}||s_{BB}|e^{j(\phi_{AA} + \phi_{BB})} - |s_{AB}|^2 e^{j2\phi_{AB}} \\ &= (|s_{AA}||s_{BB}|\cos(\phi_{AA} + \phi_{BB}) - |s_{AB}|^2\cos2\phi_{AB}) \\ &+ j(|s_{AA}||s_{BB}|\sin(\phi_{AA} + \phi_{BB}) - |s_{AB}|^2\sin2\phi_{AB}) \end{split}$$

Considering amplitude only,

$$\frac{\{|s_{AA}|^{2}|s_{BB}|^{2}-2|s_{AA}||s_{BB}||s_{AB}|^{2}\cos(\phi_{AA}+\phi_{BB}-2\phi_{AB})+|s_{AB}|^{4}\}^{\frac{1}{2}}}{|s_{AA}|^{2}+|s_{BB}|^{2}+2|s_{AB}|^{2}}$$

$$=\frac{1}{2}$$

The above equation is an identity, if

$$|S_{\Delta\Delta}| = |S_{RR}| \tag{4.8}$$

(ii)
$$\phi_{AA} + \phi_{BB} - 2\phi_{AB} = \pi$$
 (4.9)

simultaneously hold.

Again, (4.8) is consistent with high frequency scattering. It should be noted that when (jk) is neglected compared to

 $(jk)^2$ in (3.5) and (3.6) for high frequencies, both (4.8) and particularly (4.9) result. Hence, in general,

$$|D| = 0.5$$
 (4.10)

is not a trivial result merely from (4.8) as in the previous special case, but rather a consequence of the first order correction to physical optics.

4.3 Interpretation of the HF Phase Sum $(\phi_{AA} + \phi_{BB})$

From the above analysis, with only the phase being considered, it is found that

$$\tan 2\theta = \tan 2(\text{Arg A}_{F}(k))$$

$$= \frac{|S_{AA}||S_{BB}|\sin (\phi_{AA} + \phi_{BB}) - |S_{AB}|^{2}\sin 2\phi_{AB}}{|S_{AA}||S_{BB}|\cos (\phi_{AA} + \phi_{BB}) - |S_{AB}|^{2}\cos 2\phi_{AB}}$$

Assuming (4.8) and neglecting $|S_{AB}|$ compared to $|S_{AA}|$, the above equation becomes (4.7), which is now also valid for the general case. Its validity enables the phase sum $(\phi_{AA} + \phi_{BB})$ to be interpreted as twice the argument of the Fourier transform of the silhouette_l area of the target within the high frequency range.

4.4 Numerical Analysis

ζ,,

If the real and imaginary parts of D, the scattering ratio, are plotted on a complex plane, a circle of radius 0.5 will be expected for high frequency polarimetric data input. Since phase changes rapidly with frequency, a circle rather than a cluster of points of phases about 2 Arg $A_{\sf F}(k)$ will appear. This, and a direct

plot of |D| versus k will be shown in Chapter V. One significance of these plots is to check the accuracy of experimental polarimetric data of high frequencies by observing the deviation from the perimeter of the circle of radius 0.5.

4.5 Transformation to Circular Polarization Basis Vectors

The orthonormal vectors along the horizontal and vertical directions are usually chosen as the polarization bases (denoted by (\hat{H},\hat{V})) for both the transmitting and the receiving systems. However, a circular polarization basis pair may also be used [36,40], particularly in radar meteorology, in which circular polarization has a particular appropriateness on account of the direct correspondence between the mean orientation angle and the relative phase of received circular polarization components[29]. Circular polarization has also been utilized in the backscatter measurements of dielectric spheroids [32,33]. One way to investigate the form which the phase-curvature relationship may take in circular polarization basis is to transform [S(HV)] with respect to the linear basis (\hat{H},\hat{V}) to [C(RL)] with respect to a circular polarization basis (\hat{R},\hat{L}) [31,6,27]. Such transformation of [S] depends on the specification of the transformation of (\hat{H},\hat{V}) to (\hat{R},\hat{L}) by a matrix [T(RL;HV)]:

$$\begin{bmatrix} \hat{R} \\ \hat{L} \end{bmatrix} = [T(RL; HV)] \begin{bmatrix} \hat{H} \\ \hat{V} \end{bmatrix}$$
 (4.11)

where \hat{L} and \hat{R} denote the left-circular and right-circular polarization vectors, respectively. The left-circular and right-circular senses are defined in Figure 4.1. In general, the transformation of the linear basis (\hat{H},\hat{V}) to any other orthonormal basis (\hat{A},\hat{B}) (not necessarily circular polarization basis) through [T] must satisfy the normalization requirements

$$\hat{H} \cdot \hat{H} = 1$$

$$\hat{V} \cdot \hat{V} = 1$$

$$\hat{A} \cdot \hat{A} = 1$$

and the orthogonality requirement

$$\hat{A} \cdot \hat{B} = 0$$

These requirements can be shown to be mathematically equivalent to

$$[T]^{*T} = [T]^{-1}$$

which satisfies the definition of a unitary matrix. Hence [T] is a unitary matrix, and its most general form can be written as [30]

$$[T] = \begin{bmatrix} e^{j\phi_1} \cos \beta & e^{j\phi_2} \sin \beta \\ -e^{j\phi_3} \sin \beta & e^{j\phi_4} \cos \beta \end{bmatrix}$$

with ϕ_2 - ϕ_1 = ϕ_4 - ϕ_3 . The most general basis (\hat{A}, \hat{B}) is an elliptic one. When all phases are set to zero, [T] is just an ordinary rotational matrix which rotates (\hat{H}, \hat{V}) to another linear basis by an angle β . An example is given by the invariant transformation described by equations (4.3) and (4.4), which shows from the curvature recovery model that rotational transformation alone renders the invariants independent of the polarization angle (Figure 3.1). A more general case can be given

by, for instance, setting

$$\beta = \pi/4$$

$$\phi_1 = \phi_4 = 0$$

$$\phi_2 = -\phi_3 = \pi/2$$

The corresponding [T] then becomes

$$[T] = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & j \\ j & 1 \end{bmatrix} , \qquad (4.12)$$

and the corresponding (A,B) reduces to a circular polarization basis

$$\begin{bmatrix} \hat{R} \\ \hat{L} \end{bmatrix} = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & j \\ j & 1 \end{bmatrix} \begin{bmatrix} \hat{H} \\ \hat{V} \end{bmatrix}$$

By limiting the transformation from (\hat{H},\hat{V}) to circular polarization basis only, any polarization vector can then be expressed in terms of either basis. For instance, the incident electric field polarization vector \underline{E}^i can be written as

$$\underline{\mathbf{E}^{i}} = \begin{bmatrix} \mathbf{E}_{H}^{i} & \mathbf{E}_{V}^{i} \end{bmatrix} \begin{bmatrix} \hat{\mathbf{H}} \\ \hat{\mathbf{v}} \end{bmatrix}$$
$$= \begin{bmatrix} \mathbf{E}_{R}^{i} & \mathbf{E}_{L}^{i} \end{bmatrix} \begin{bmatrix} \hat{\mathbf{R}} \\ \hat{\mathbf{L}} \end{bmatrix}$$

where the linear phasor components \mathbf{E}_{H}^{i} , \mathbf{E}_{V}^{i} can be related to the

circular phasor components E_{R}^{i} , E_{L}^{i} as follows

$$\begin{bmatrix} \vec{E}_{R} \\ \vec{E}_{L} \end{bmatrix} = [T(RL; HV)]^{*} \begin{bmatrix} \vec{E}_{H} \\ \vec{E}_{V} \end{bmatrix}$$

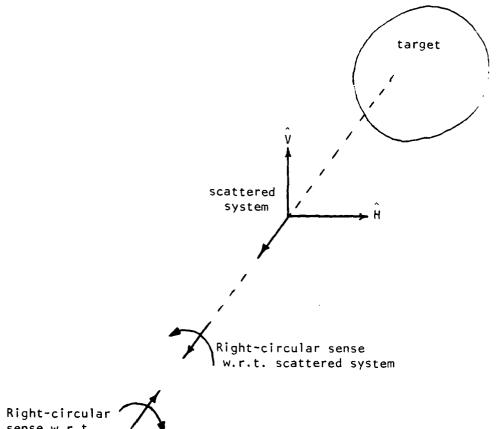
$$(4.13)$$

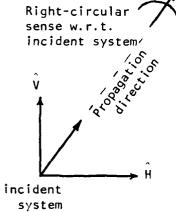
where the symbols * and [] $^{\mathsf{T}}$ denote conjugation and transposition, respectively. In (4.13), the incident polarization vector can be regarded as being transformed in changing the polarization basis as specified by the unitary matrix [T] given by (4.11). The scattered polarization vector $\underline{\mathbf{E}}^{\mathsf{S}}$ can similarly be transformed. $\underline{\mathbf{E}}^{\mathsf{S}}$ and $\underline{\mathbf{E}}^{\mathsf{I}}$ can both be specified in terms of the circular polarization basis. Yet, it is preferable to use distinct systems of unit vectors to specify the incident and scattered fields, so that right-hand elliptical polarization may have the same sense with regard to the coordinate system for incident radiation as it does with regard to the coordinate system for scattered radiation [30,7] (Figure 4.1). If (4.11) is prescribed for the incident system, the desired similarity of sense for the two coordinate systems can be accomplished by writing

$$\begin{bmatrix} \hat{R} \\ \hat{L} \end{bmatrix} = [T(RL; HV)]^{*} \begin{bmatrix} \hat{H} \\ \hat{V} \end{bmatrix}$$
(4.14)

for the scattered system. Because of the conjugation now introduced, the relative phases in [T] are negated, and thus the sense of rotation is reversed. Also, it follows that

$$\begin{bmatrix} \hat{R} \\ \hat{L} \end{bmatrix} = \begin{bmatrix} \hat{R} \\ \hat{L} \end{bmatrix}^*$$





w.r.t.--- with respect to

Figure 4.1 Incident and Scattered Coordinate Systems

On adopting this choice for the circular polarization basis of the scattered field, (4.15) follows from (4.14), with the double conjugation being ignored,

$$\begin{bmatrix} E_{R}^{s} \\ E_{L}^{s} \end{bmatrix} = [T(RL; HV)] \begin{bmatrix} E_{H}^{s} \\ E_{V}^{s} \end{bmatrix}$$

$$(4.15)$$

in the same way (4.13) follows from (4.11).

In terms of the linear basis (\hat{H},\hat{V}) , the scattering matrix [S(HV)] fully describes the scattered depolarized field, with the incident field given, by

$$\begin{bmatrix} E_{H}^{s} \\ E_{V}^{s} \end{bmatrix} = [s(HV)] \begin{bmatrix} E_{H}^{i} \\ E_{V}^{i} \end{bmatrix} . \tag{4.16}$$

Similarly, in terms of the circular basis (\hat{R},\hat{L}) , it follows that

$$\begin{bmatrix}
E_{R}^{s} \\
E_{L}^{s}
\end{bmatrix} = [C(RL)] \begin{bmatrix}
E_{R}^{i} \\
E_{L}^{i}
\end{bmatrix} .$$
(4.17)

Using the definition of the unitary matrix, it follows from (4.16), (4.17), (4.13) and (4.15) that

$$[C(RL)] = [T(RL;HV)] [S(HV)] [T(RL:HV)]^T$$
 (4.18)

which transforms [S] to [C] through [T] defined in (4.11), and which is of the form of congruence transformation.

4.6 Curvature Recovery from the Circular Polarization [C]

If the unitary matrix in (4.12) is adopted, then (4.18) becomes

$$[C(RL)] = \begin{bmatrix} \frac{S_{HH} - S_{VV}}{2} + j S_{HV} & j \frac{S_{HH} + S_{VV}}{2} \\ j \frac{S_{HH} + S_{VV}}{2} & \frac{S_{VV} - S_{HH}}{2} + j S_{HV} \end{bmatrix}$$
(4.19)

provided that $S_{HV} = S_{VH}$, which is true for the monostatic reciprocal case (also $C_{RL} = C_{LR}$). If conjugation in (4.14) is used to preserve the similarity of sense in both the scattered and incident systems, then an examination of (4.18) reveals that reciprocity is satisfied, i.e. [C] is symmetric(if [S] is symmetric). A different choice of [T] may result in a different [C]. For instance,

if
$$[T] = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & +j \\ & \\ 1 & -j \end{bmatrix}$$
 (4.12')

then [C] =
$$\frac{S_{HH} - S_{VV}}{2} + j S_{HV} \qquad \frac{S_{HH} + S_{VV}}{2}$$

$$\frac{S_{HH} + S_{VV}}{2} \qquad \frac{S_{HH} - S_{VV}}{2} - j S_{HV}$$
(4.19')

Returning to the [C] in (4.19), in view of (3.5) to (3.7), the matrix elements can be written as

$$C_{RR} = -k A_F(k) \frac{K_u - K_v}{4\pi} [\sin 2\alpha + j \cos 2\alpha]$$
 (4.20)

$$C_{LL} = -k A_F(k) \frac{K_u - K_v}{4\pi} [\sin 2\alpha - j \cos 2\alpha] \qquad (4.21)$$

$$C_{LR} = C_{RL} = -j \frac{1}{2\pi} k^2 A_F(k)$$
 (4.22)

Hence,

$$\left\{\frac{K_{U} - K_{V}}{2}\right\}^{2} = -k^{2} \frac{C_{LL} C_{RR}}{C_{LR}^{2}}$$
 (4.23)

which is an equation of curvature recovery from the scattering matrix in circular polarization.

Expressing the ratio C $_{LL}$ C $_{RR}$ / C $_{LR}^{2}$ in terms of the linear polarization [S] elements,

$$\frac{c_{LL} c_{RR}}{c_{LR}^2} = \frac{(1 - \Re)^2 + 4\left(\frac{s_{HV}}{s_{HH}}\right)^2}{(1 + \Re)^2}$$
(4.24)

Combining (4.23) and (4.24),

$$\frac{K_{u} - K_{v}}{2} = \pm jk \left\{ \frac{C_{LL} C_{RR}}{C_{LR}^{2}} \right\}^{\frac{1}{2}}$$
 (4.25)

$$= \pm jk \sqrt{\frac{(1-\hat{R})^2 + 4(\frac{S_{HV}}{S_{HH}})^2}{(1+\hat{R})^2}}$$
 (4.26)

Comparing (4.26) to (3.8), it can be observed that the unknown

polarization angle α in (3.8) is being disguised in (4.26) and appears in the form of S_{HV}/S_{HH} which is incorporated into the square root. Thus one advantage of utilizing circular polarization over utilizing linear polarization for curvature recovery is that the polarization angle does not have to be determined, but still the entire scattering matrix has to be measured.

It can be seen that for the case in which $S_{HV}=0$ ($\alpha=0$ or 90 degrees), (4.26) reduces to (3.8) and then (3.12). As in the derivation of the phase-curvature relationship, the imaginary part of the square root of C_{RR} C_{LL} / C_{RL} should give better curvature difference and the real part should vanish, as frequency is increasing.

It is to be noted that under suitable conditions, quantities such as the radar cross sections σ_{RL} , σ_{RR} , σ_{LL} and quantities derived from them (e.g. $(1-\tilde{R})/(1+\tilde{R})$) yield meaningful measurables in measurements of the backscatter of dielectric spheroids and hydrometeors [32,33,34]; in radar target discrimination techniques, the quantity $|C_{RR}|/|C_{LL}| - |C_{RL}|^2$ in [36], in view of (4.20) to (4.22), can be interpreted as $[-(1/2\pi)^2 k^4 |A_F(k)|^2]$, which reveals area information for a smooth, convex, conducting target at high frequencies.

4.7 Transformation Invariants of the Scattering Matrix

The transformation of [S] to [C] due to transforming the linear basis (\hat{H},\hat{V}) to the circular basis (\hat{R},\hat{L}) can be achieved by introducing the appropriate relative phase between the two orthonormal vectors of the (H,V) basis in addition to rotation. To investigate if the transformation is still invariant in changing linear to circular bases, the Det of both sides of (4.18) is taken :

Det
$$\{[C]\}\ = \ Det \ \{[T]\}\$$
. Det $\{[S]\}\$. Det $\{[T]^T\}$

$$= \ Det \ \{[T]\}\$$
. Det $\{[S]\}\$. Det $\{[T]^{*}-1\}$

$$= \ \{\frac{\ Det \ \{[T]\}}{\ |\ Det \ \{[T]\}\}}\}^2 \quad$$
. Det $\{[S]\}$

$$= \ e^{j2Arg(Det\{[T]\})} \quad$$
. Det $\{[S]\}$ (4.27)

From the general form of [T] given on page 33,

Arg(Det
$$\{[T]\}\$$
) = $\phi_1 + \phi_4 = \phi_2 + \phi_3$

Hence,

Det
$$\{[C]\} = e^{j2(\phi_1 + \phi_4)}$$
. Det $\{[S]\}$ (4.28)

It can be easily proven that $|\text{Det }\{[T]\}| = 1$ for the unitary matrix [T], but it is generally not true that the determinant of a unitary matrix is purely real. Thus, the determinant of [C] is strictly invariant iff $[Det \{[T]\}]$ is purely real, i.e.

Det
$$\{[T]\} = \pm 1$$
, (4.29)

otherwise the invariant differs only by a phase shift of $2(\phi_1 + \phi_4)$. For examples, the [T] given by (4.12) renders

$$c_{RR} c_{LL} - c_{RL}^2 = s_{HH} s_{VV} - s_{HV}^2$$
 (4.30)

as is also evident from (4.19). The invariant value is the same as that in (4.4) derived from the curvature recovery model, since the rotational matrix possesses a real, unity determinant. On the other hand, the [T] in (4.12) will not preserve strict transformation invariance, as is also

evident from (4.19') (i.e. the determinants of [C] and [S] are equal in magnitude but of opposite sign).

To show from the curvature recovery model the invariance of Span $\{[C]\}$, equations (4.20)-(4.22) and (3.5)-(3.7) can be used to give

Span {[C]} =
$$|c_{RR}|^2 + |c_{LL}|^2 + 2|c_{RL}|^2$$

= $|s_{AA}|^2 + |s_{BB}|^2 + 2|s_{AB}|^2$
= Span {[S]}

which equals the invariant value in (4.3).

It can be shown that the span is in general invariant, regardless of whether (4.29) holds or not. Denoting the trace of a square matrix (i.e. the sum of the diagonal elements) by Tr, it follows that

Span {[C]} =
$$Tr \{[C]^*[C]\}$$

= $Tr \{[[T][S][T]^T]^*[[T][S][T]^T]\}$ (from (4.18))
= $Tr \{[[T]^*[S]^*[S][T]^T]\}$

 $[S]^*[S]$ is identified as the power scattering matrix [P] of [S] [39,38], i.e.

$$[P] = [S]^{*}[S]$$
 (4.31)

Let [P'] be the power scattering matrix of [C], i.e.

$$[P'] = [C]^{*}[C] \tag{4.32}$$

$$= [T]^{*}[S]^{*}[S][T]^{T}$$
 (4.33)

Hence,

$$Span \{[C]\} = Tr \{[P']\}$$

$$= \sum_{i} P_{ii}^{\dagger}$$

$$= \sum_{k} \sum_{l} \{P_{kl} [\sum_{i} T_{ik}^{*} T_{il}]\}$$

Since the set of conditions

$$|T_{11}|^2 + |T_{21}|^2 = 1$$

 $|T_{12}|^2 + |T_{22}|^2 = 1$
 $T_{11}T_{12}^* + T_{21}T_{22}^* = 0$

is equivalent to the definition of a unitary matrix, hence (after some algebraic manipulation) it follows that

Span
$$\{[C]\} = P_{11} + P_{22}$$

Thus,

Span
$$\{[C]\}\ =\ Tr\ \{[P']\}\ =\ Tr\ \{[P]\}\ =\ Span\ \{[S]\}\ (4.34)$$

4.8 An Interpretation of the Scattering Ratio

The radar cross section $\sigma_{\mbox{\scriptsize rt}}$ has been defined in [41] as follows :

$$\sigma_{rt} = \left| \underline{h}^{r} \cdot [S] \underline{h}^{t} \right|^{2} \tag{4.35}$$

where \underline{h}^t is the transmitting polarization vector and \underline{h}^r is the antenna height [20]. In this definition, \underline{h}^t and \underline{h}^r are normalized to unity. It has been shown in [20] that for radar systems that use identical transmitting and receiving antennas, the radar cross section is maximum if

$$[S]\underline{h} = \lambda_S \underline{h}^* \tag{4.36}$$

where \underline{h} denotes the antenna polarization which yields maximum power reception, and λ_s denotes the complex eigenvalue of (4.36). Moreover, the scattering matrix can be diagonalized by a change-of-basis unitary transformation, using a unitary matrix which consists of the eigenvectors of (4.36) 20,42. The diagonalized form of [S] is

$$[s_d] = \begin{bmatrix} \lambda_{s1} & 0 \\ 0 & \lambda_{s2} \end{bmatrix}$$

where λ_{s1} and λ_{s2} are the eigenvalues, with $|\lambda_{s1}| > |\lambda_{s2}|$. The maximum radar cross section is given by [20]

$$\sigma_{\text{max}} = \left| \lambda_{s1} \right|^2 \tag{4.37}$$

The corresponding monostatic power scattering matrix is thus diagonalized too:

$$[P_d] = [S_d]^* [S_d] \quad \text{(from equation (4.31))}$$

$$= \begin{bmatrix} |\lambda_{s1}|^2 & 0 \\ 0 & |\lambda_{s2}|^2 \end{bmatrix}$$

By the invariance properties of [S],

Span {[S]} = Span {[S_d]}
=
$$|\lambda_{S1}|^2 + |\lambda_{S2}|^2$$
 (4.38)
= Tr {[P_d]}
= Tr {[P]}

Assuming (4.29) for strict invariance,

Det
$$\{[S]\}\ = Det \{[S_d]\}\$$

$$= \lambda_{S1} \lambda_{S2} \qquad (4.40)$$

Therefore,

$$D = \frac{\text{Det } \{[S]\}}{\text{Span } \{[S]\}}$$

$$= \frac{\lambda_{S1} \lambda_{S2}}{|\lambda_{S1}|^2 + |\lambda_{S2}|^2}$$
(4.41)

For high frequencies, it has been shown in this thesis that

$$D = 0.5 e^{j2Arg A_F(k)}$$
 (4.42)

Comparing (4.41) and (4.42) and assuming that frequency is increasingly high,

$$|\lambda_{s1}| \rightarrow |\lambda_{s2}| \tag{4.43}$$

and Arg
$$\lambda_{s1}$$
 + Arg λ_{s2} + 2 Arg A_F(k) (4.44)

Combining (4.37), (4.39), (4.41), (4.42) and (4.43),

Tr
$$\{[P]\} \simeq 2 \sigma_{\text{max}}$$
 (4.45)

for high frequencies.

Accordingly, |D| may be interpreted as the ratio of the maximum radar cross section to the trace of the power scattering matrix.

The power scattering matrix has been defined by Graves in [39] and

represents the total power backscattered from the target for any transmitting polarization.

To conclude this chapter, it is to be noted that the invariant transformation can of course be extended to the more general elliptic case. The scattering ratio has tacitly been extended in its definition to the general elliptic polarization in Section 4.8. Equations (4.7) and (4.10) which describe the behavior of the scattering ratio are thus generalized.

CHAPTER V

NUMERICAL VERIFICATION

5.1 Data Description

Both theoretical and experimental data are available to verify the special case of the phase-curvature relationship (3.12), i.e. the case in which the incident polarization coincides with one of the directions of the principal curvatures at specular points on the equator of the prolate spheroid. The theoretical data was obtained by a time-domain synthesis of the impulse response technique [15]. The solutions generated by this technique were checked against other theoretical solutions [8,16] with excellent agreement [15]. The theoretical data were converted in the form of amplitudes and phases of the elements of the scattering matrix. The experimental data were measured at the Electro-Science Laboratory of the Ohio State University (ESL-OSU). The experiments were conducted [28] on a frequency-domain range yielding the backscattered returns $\mathbf{S}_{\mathbf{V}\mathbf{V}}$ and $\mathbf{S}_{\mathbf{H}\mathbf{H}}$ (S $_{\mathbf{H}\mathbf{V}}$ and S $_{\mathbf{V}\mathbf{H}}$ being zero in this case). The 'size' of the prolate spheroid used in the experiment was 6 inch : 12 inch, and the data were measured for two principal polarization cases in which $\alpha = 0$ (VV or TE) and $\alpha = \pi/2$ (HH or TM) for aspect angles from 0 (nose-on) to SO degrees (broad-side on) in steps of 15 degrees (Figure 3.2). After measurement, the data were smoothed. Two of the smoothed frequency-domain data blocks, namely, the 2-4 GHz block and the 4-8 GHz block, were used in this thesis, covering a range of 3.19 to 12.76 in terms of the values of kb $(2\pi/V)$.b). The error bounds

on the experimental data were specified to be \pm 2 dB in magnitude and \pm 10 degrees in phase data.

5.2 Direct Verification of the Phase-Curvature Relationship

5.2.1 Theoretical Data

A typical result obtained with the theoretical test data is presented in Figure 5.1. In Figure 5.1(a), the right-hand side of (3.12) is plotted against kb (in steps of 0.1 from 0.1 to 18). From this graph it is clear that as frequency increases, the phasecurvature relationship becomes more accurate. The aspect angle is 90 degrees (broadside incidence), and the corresponding curvature difference divided by two is 0.375, to which the right-hand side of (3.12) converges. In Figure 5.1(b), the real part of $(1 - \frac{2}{K})/(1 + \frac{2}{K})$ multiplied with k approaches zero. Although not presented in this thesis, the real part itself (without the factor k) converges to zero at a much faster rate, particularly at large values of k. In Figure 5.1(c), the imaginary part multiplied with k tends to the value 0.375. The scattering chart (imaginary part versus real part) is shown in Figure 5.1(d). As predicted, the plot is indeed a spiral which, as k is increased, converges to the point (0, -0.375)on the imaginary axis.

5.2.2 Experimental Data

An extensive amount of testing of the phase-curvature relationship has been conducted with the experimental input data. It was realized that because of the nature of the tangent function, a direct test of (3.12) with input data which have a \pm 10 degree error in phase measurement, is not very useful in a graphical presentation. Thus the complex plane plots of some typical measurement data are presented in Figures 5.2 to 5.4.

The plots in Figure 5.2(a)(i) to (iii) are the experimental versions of the theoretical plots in Figures 5.1(d), (b) and (c). Figures 5.2(a)(i) does indeed hover around the predicted point on the imaginary axis. This behavior is not so clearly visible in Figure 5.2(b)(i), where according to the theoretical predictions this plot should have given better results for the higher frequency range. This discrepancy is mainly attributed to the following factor: the phase error magnified through the tangent function gets even more magnified through the multiplication with large values of k. The plots shown in Figures 5.3(a) and (b) are for the nose-on incidence case $(\phi = 0)$ for which there is no polarization dependence and $(K_U - K_V)/2$ vanishes. Figure 5.3(b) is of the 4-8 GHz block. The case presented in Figures 5.4(a) and (b) are for ϕ = 45 degrees, which is representative of a typical aspect. Once again, the results for the 4-8 GHz block are not as good as the 2-4 GHz block.

It is to be noted that the relative phase error mentioned in Chapter IV becomes significant at high frequencies. Figure 5.5 shows the effects of an error of 0.5 millimeters between the TE and the TM antenna positions along the direction of propagation. The error is simulated within the theoretical data. The scattering chart of

Figure 5.1(d) now changes from a spiral to a helix elongating mainly along the imaginary axis and away from the exact value of $(K_u - K_v)/2$. If the error is negative, the helix elongates in the opposite direction. An observation of the scattering charts of the 4-8 GHz block reveals that the plots somewhat look like a helix rather than a spiral. Thus, the relative phase error partially explains that, for the 4-8 GHz data, the scattering chart deviates more from theoretical prediction than for the 2-4 GHz data. It is to be noted that in the 4-8 GHz block, there were many data points for which accurate results were found. In general, all the experimental tests and the theoretical data supported the approximate phase-curvature relationship well.

5.3 Verification of the Scattering Ratio D

5.3.1 Theoretical Data

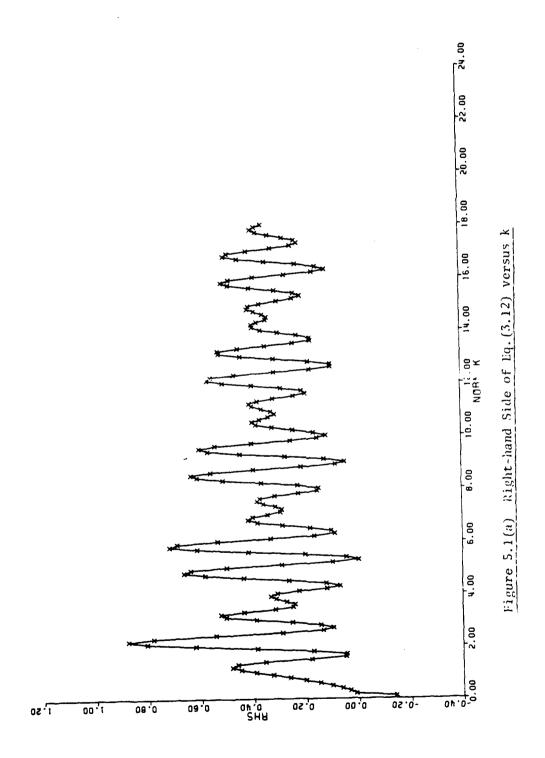
A complex plot of D (imaginary part versus real part) with the theoretical data as input is depicted in Figure 5.6(a). For high frequencies, the phase of D, in theory, converges to a constant value of 2 Arg $A_F(k)$. However, a circle rather than a cluster of points of phases around 2 Arg $A_F(k)$ results. This is due to the fact that the phase fluctuates rapidly at frequencies not high enough. Even the range of frequencies covered by the set of theoretical data (kb up to 18) is not sufficiently high to show the convergence. As predicted, the radius of the circle is indeed 0.5. In Figure 5.6(b), only the amplitude of D is plotted versus kb. Clearly, the amplitude converges to 0.5 even at low values of k, in contrast to the behavior

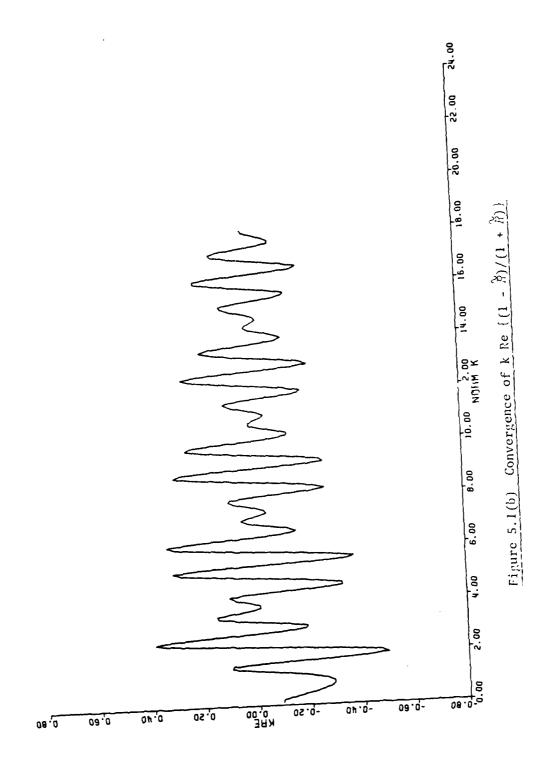
of the phase of D.

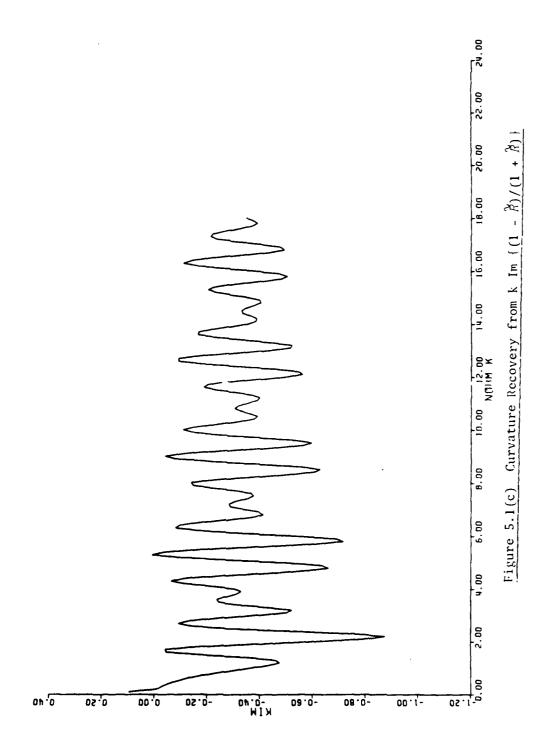
5.3.2 Experimental Data

Figures 5.7 to 5.12 are the experimental versions of Figure 5.6. Figures 5.7 to 5.9 cover the frequency range from 2 to 4 GHz, with aspect angles being 90, 0 and 45 degrees respectively, whereas the rest cover the frequency range from 4 to 8 GHz. All the complex plots of D depict an envelope circle of radius 0.5. The nearer the data points to this envelope, the more accurate they are (for high values of kb).

It is interesting to observe that both Figure 5.12 and 5.4(b) show that the data block of 4-8 GHz of 45 degree aspect is least accurate among the experimental data blocks.







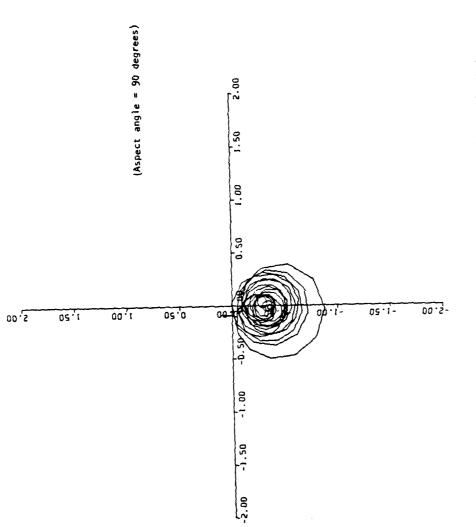


Figure 5.1(d) The Scattering Chart from Theoretical Data

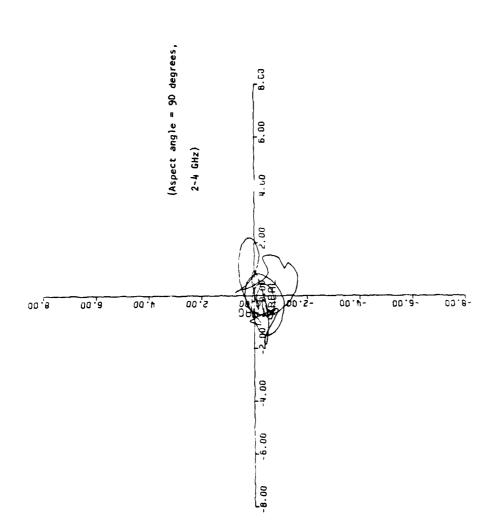
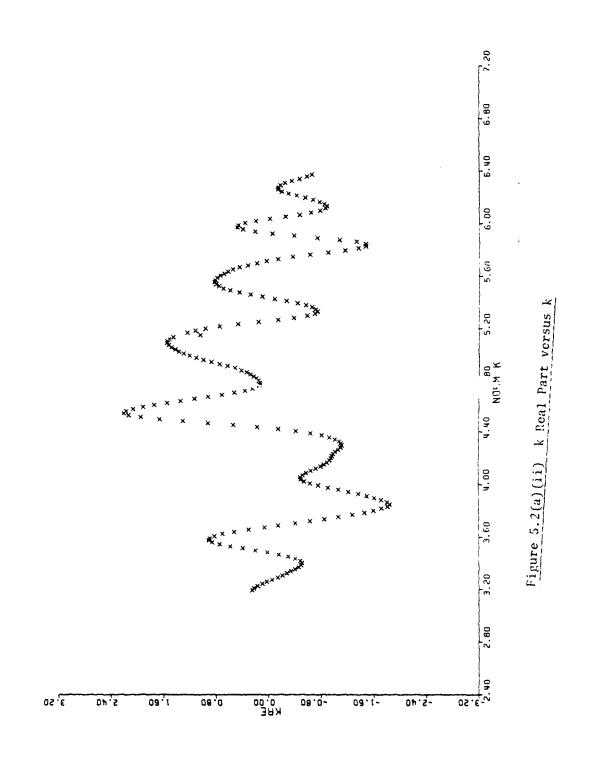
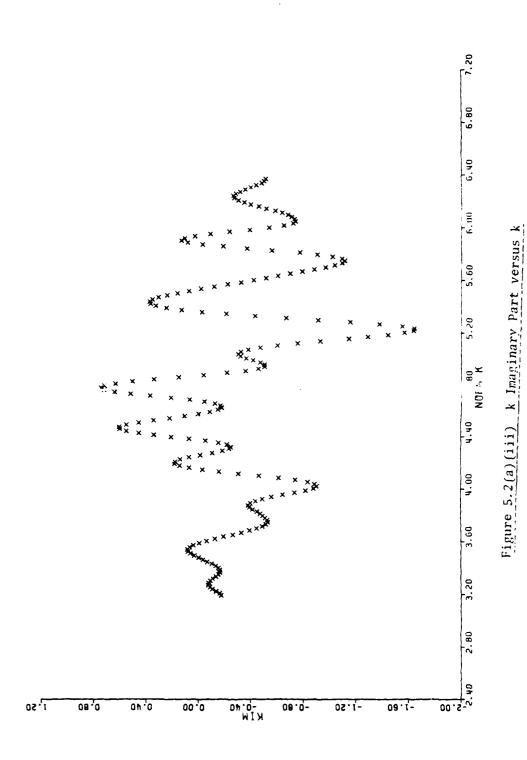
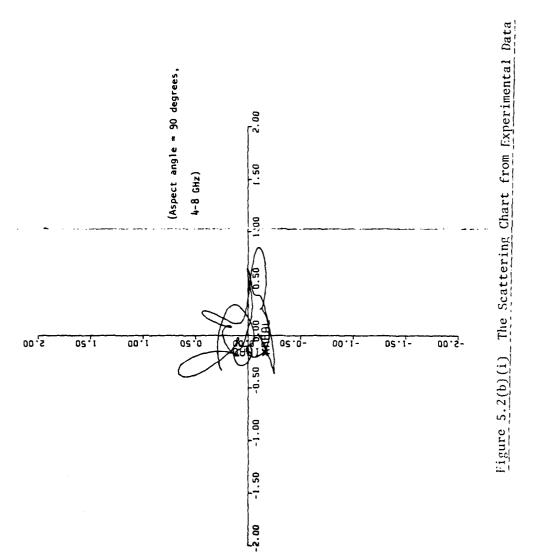
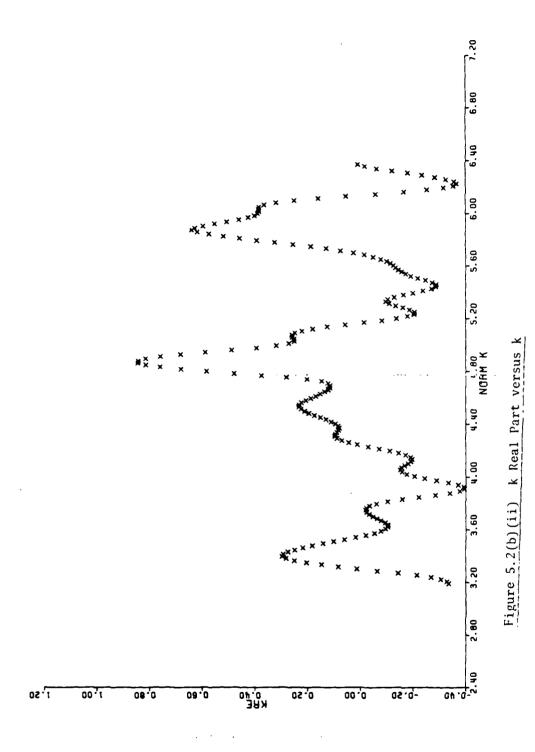


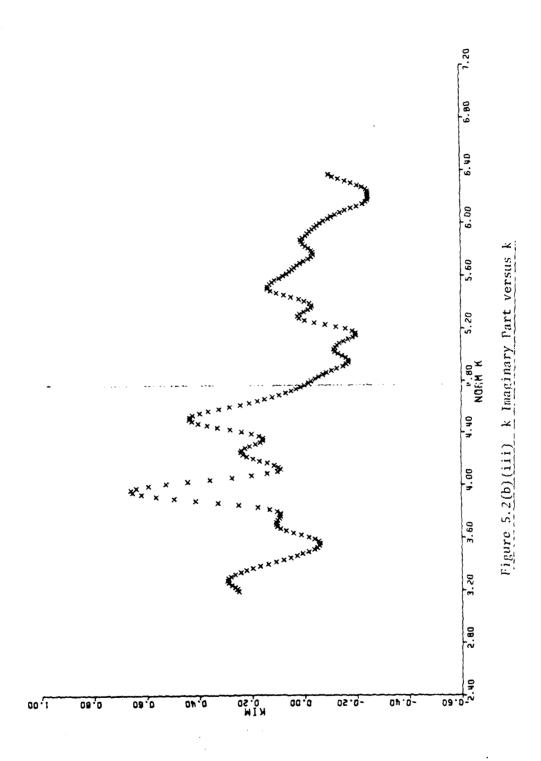
Figure 5.2(a)(i) The Scattering Chart from Experimental Nata

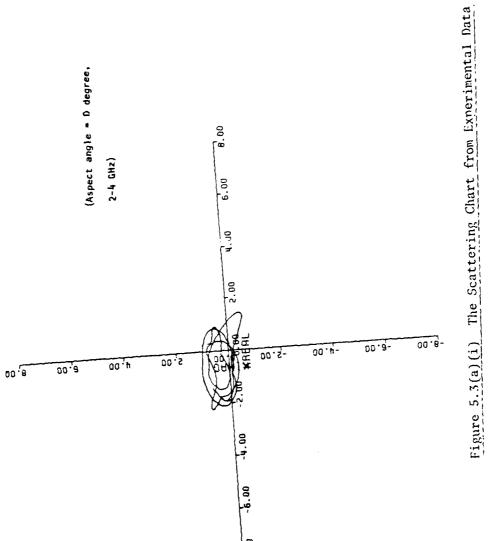


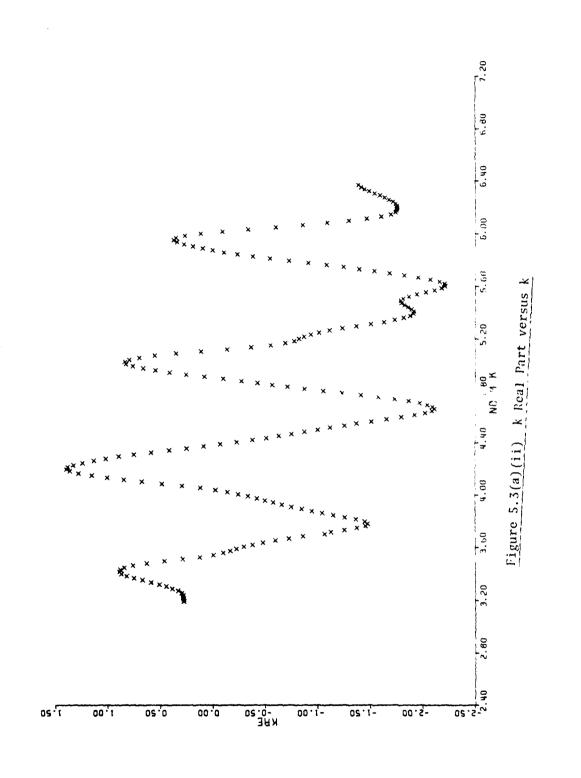


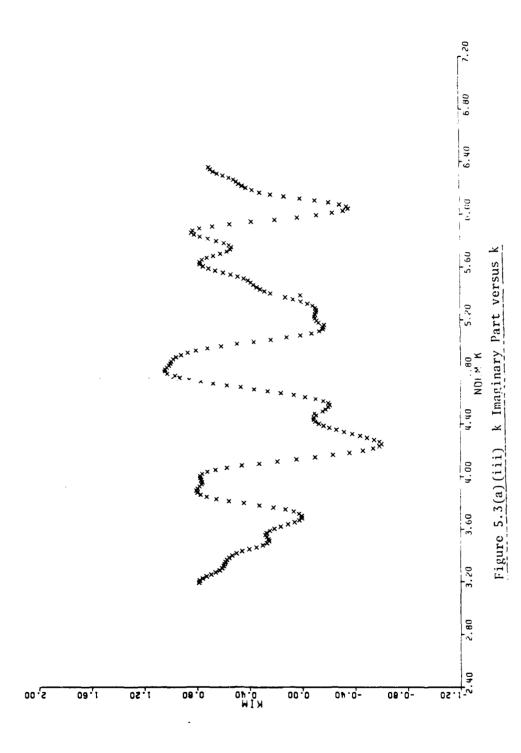












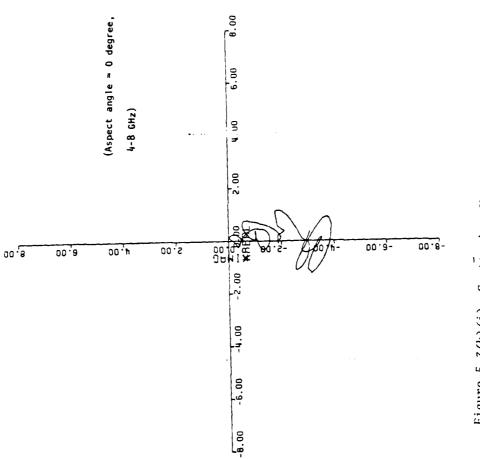


Figure 5.3(b)(i) Scattering Chart from Experimental Data

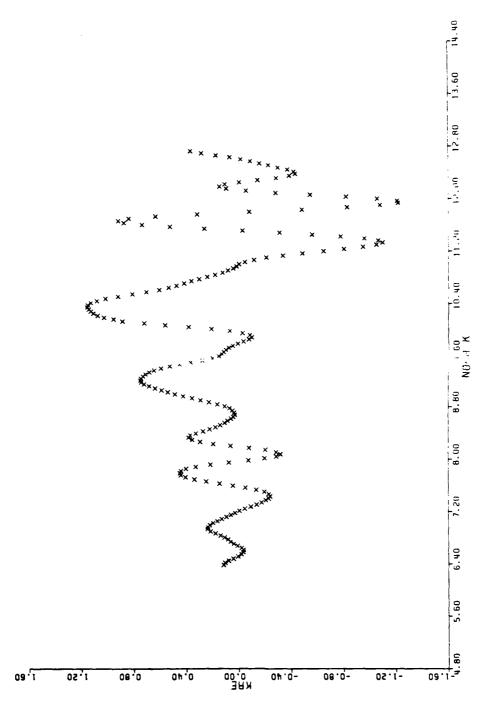
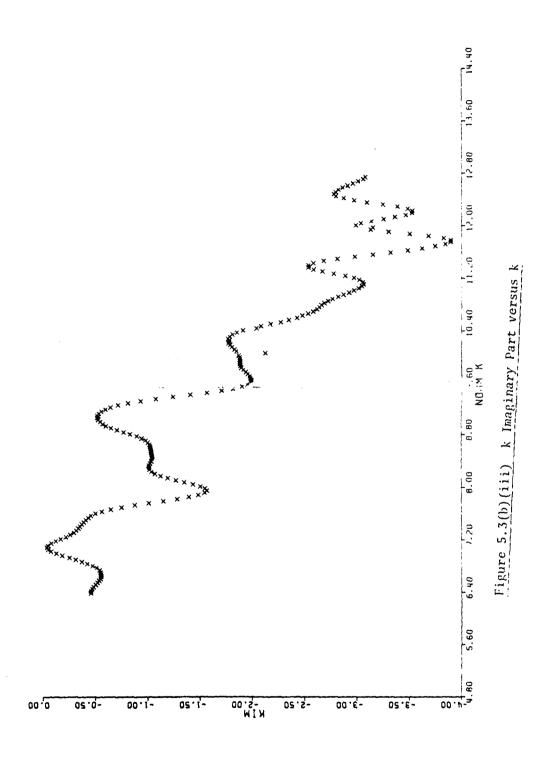


Figure 5.3(b)(ii) k Real Part versus k



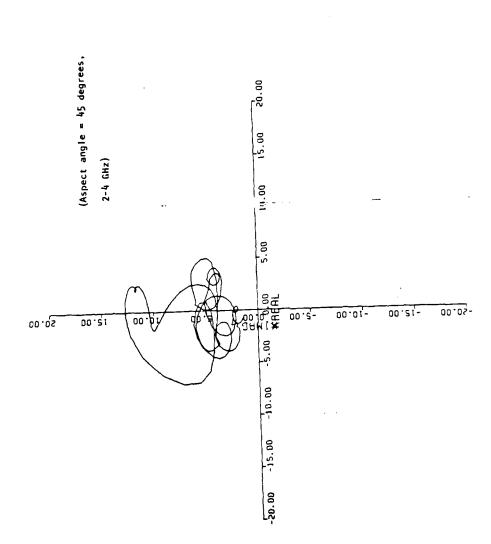
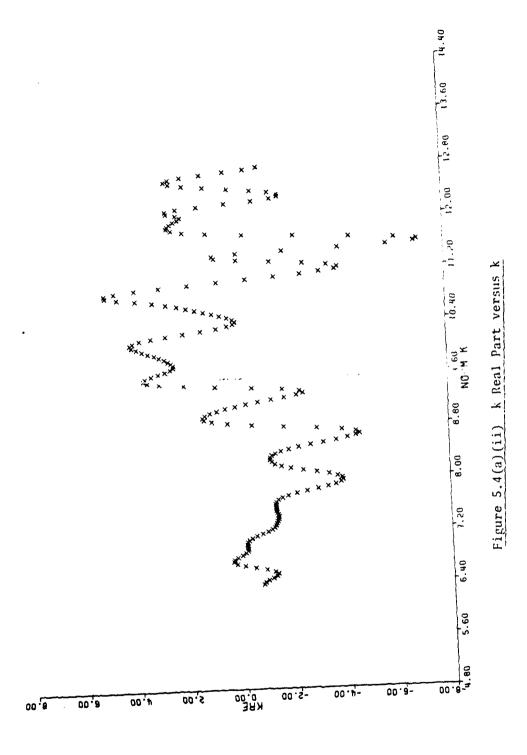
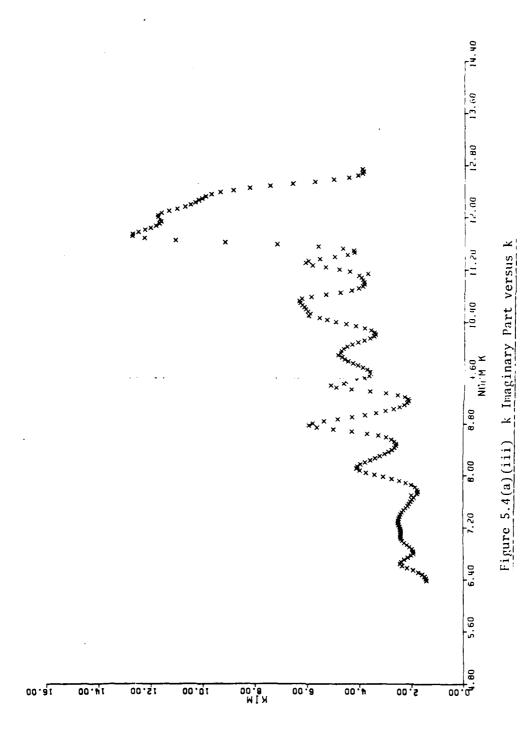


Figure 5.4(a)(i) The Scattering Chart from Experimental Data





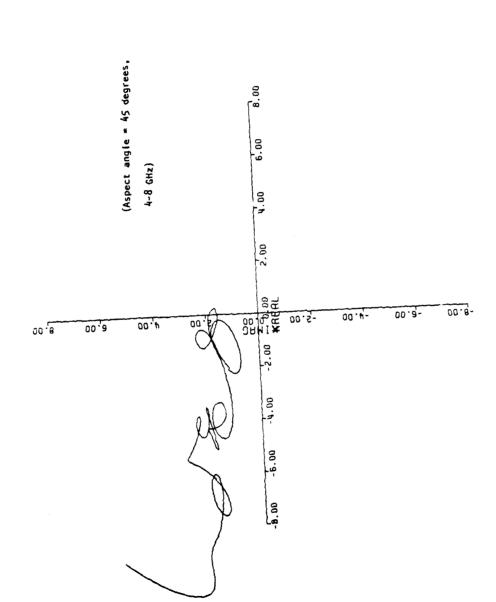
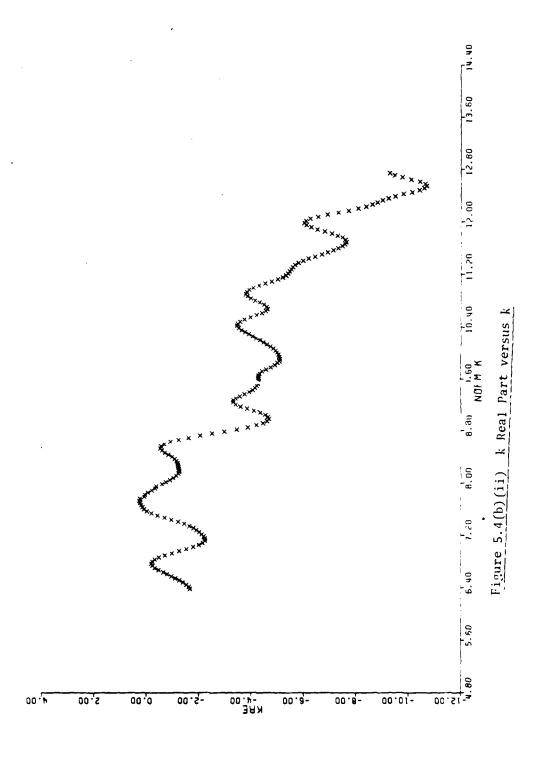
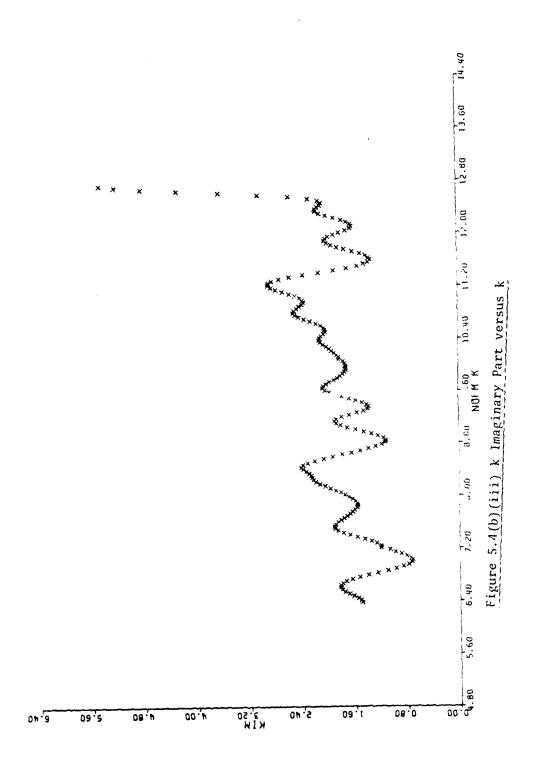
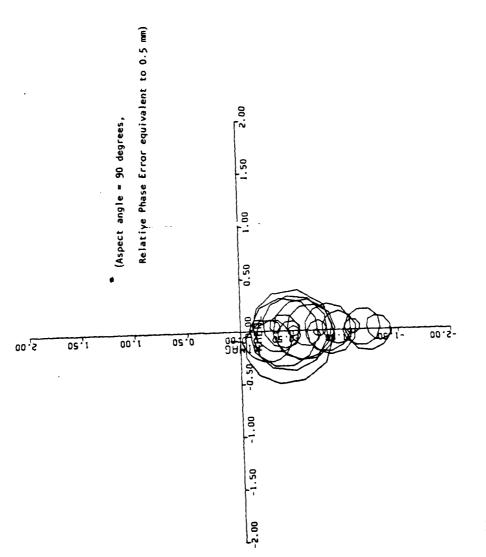


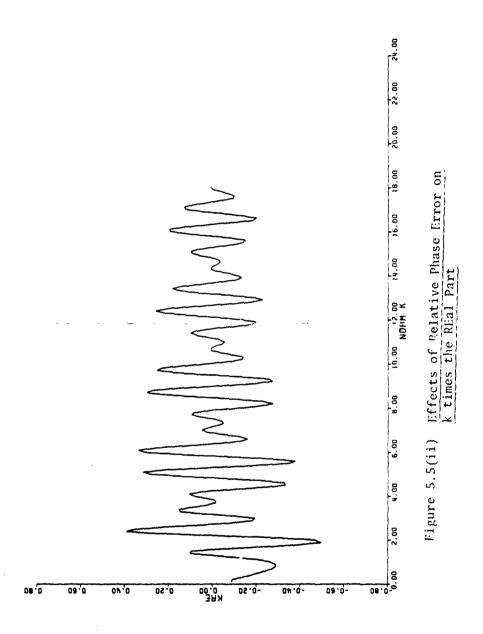
Figure 5.4(b)(i) The Scattering Chart from Experimental Data

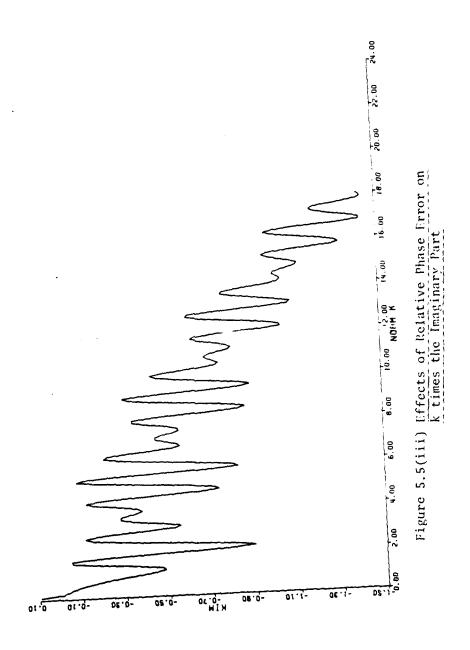






Effects of Relative Phase Error on the Scattering Chart from Theoretical Data Figure 5.5(i)





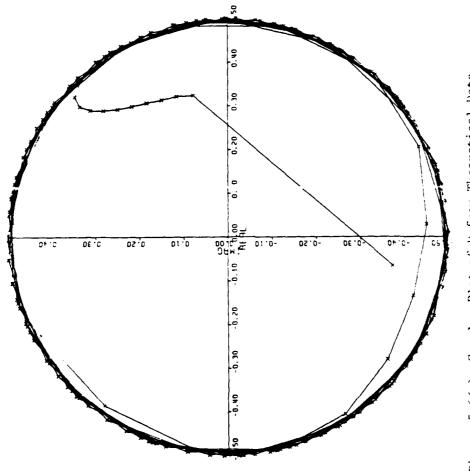
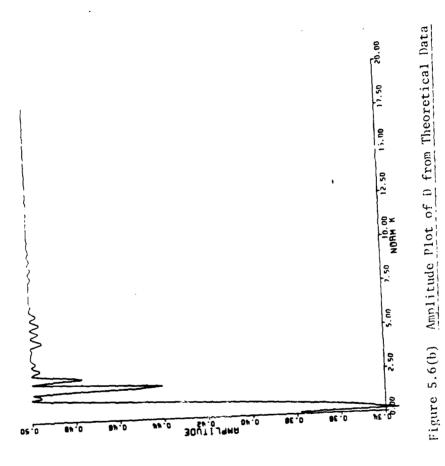
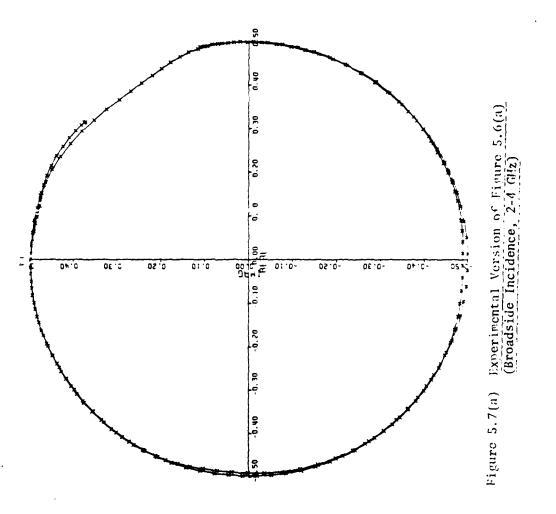
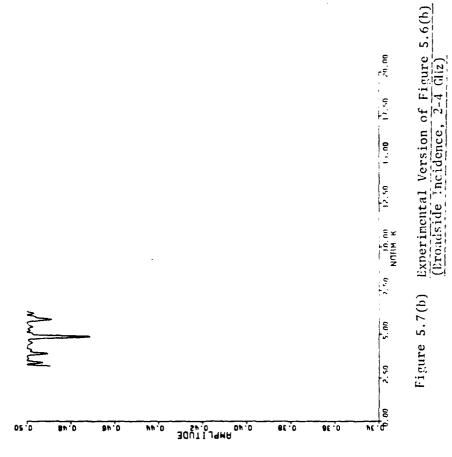


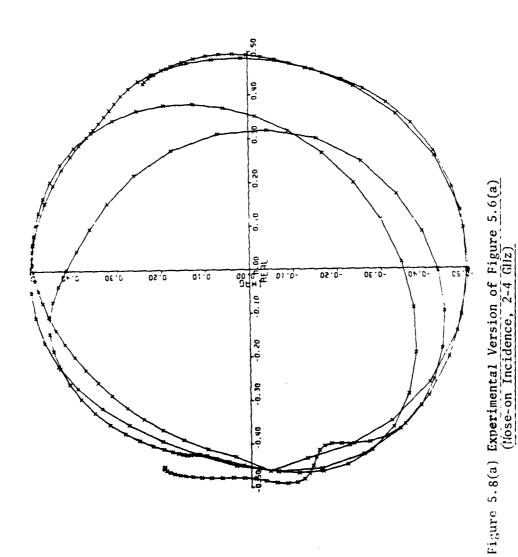
Figure 5.6(a) Complex Plot of D from Theoretical Data

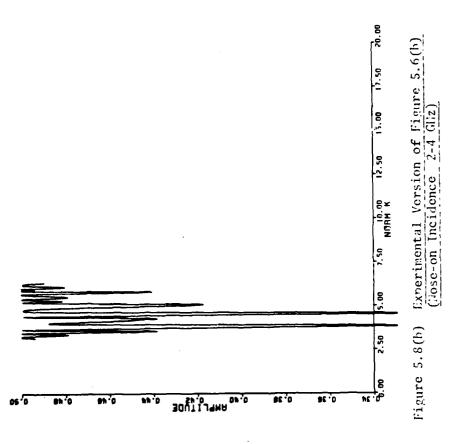


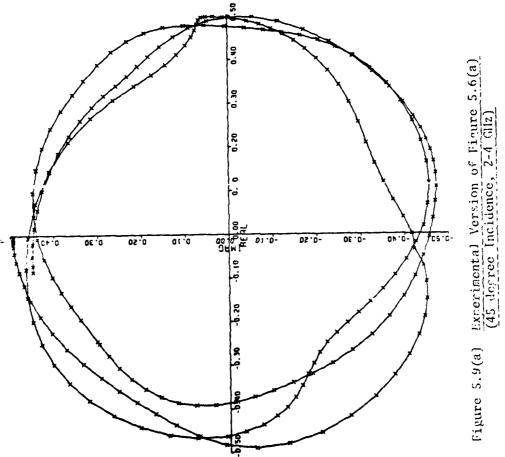
	AD-A	115 60	5	AN	ILLINOIS UNIV AT CHICABO CIRCLE ELECTROMAGNETIC IMAGIETC F/6 20/14 A HIGH FREQUENCY INVERSE SCATTERING MODEL TO RECOVER THE SPECULETC(U) MAY 82 B F00 N00018-80-C-0708											
	UNCLASSIFIED		ED			82-05-	21-01	1-01				NL NL				
		2 or 2														
_	_	<u>.</u>														
H				-												
Ī																
															END PATE FILWED	
															DTIC	

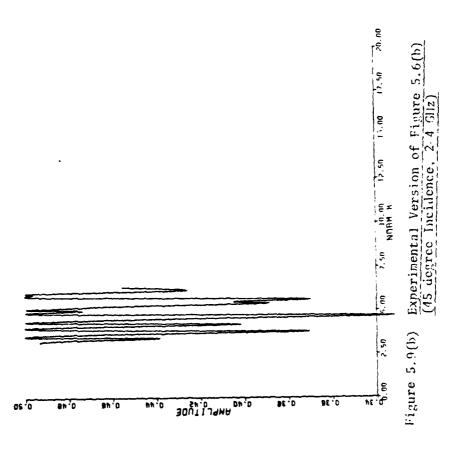


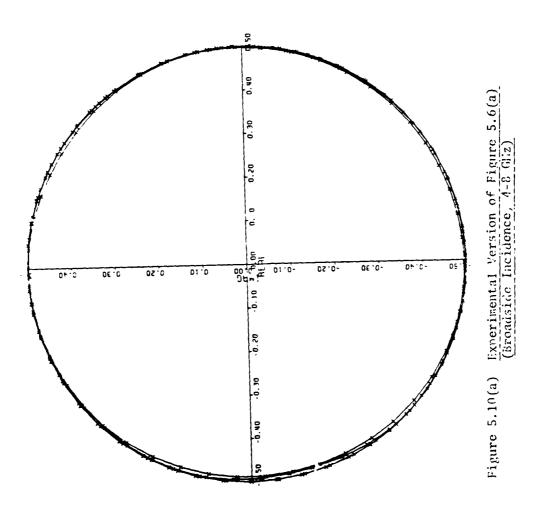


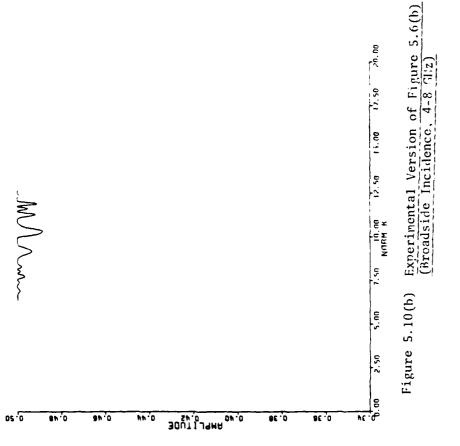












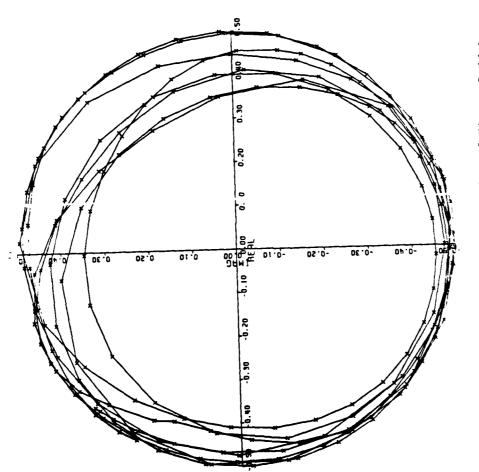
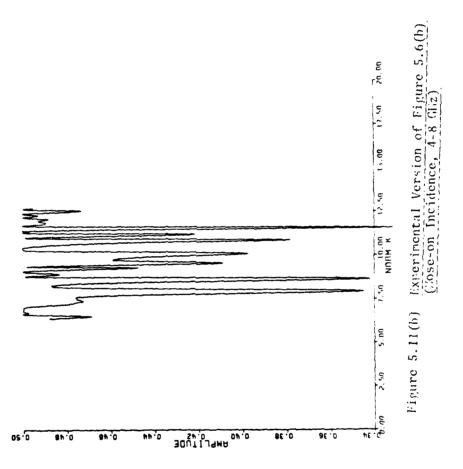
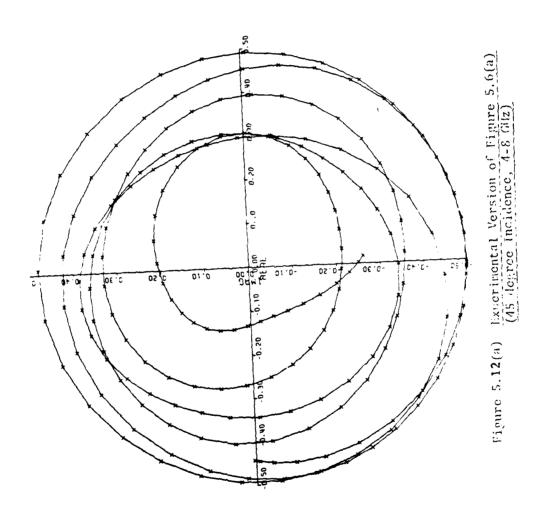
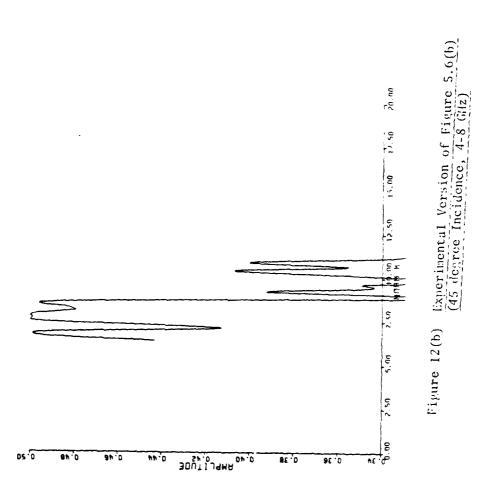


Figure 5.11(a) Experimental Version of Figure 5.6(a) (Soc-on Incidence, 4-8 (412)







CHAPTER VI

CONCLUSION

In the high frequency limits, the phase terms of the polarimetric scattering data have been used to determine the difference between the principal curvatures at the specular point. It is also known that the geometric optics backscattered cross-section is related to the geometric mean of the principal curvatures at the specular point (for smooth, convex objects at least [6]). Therefore, it should be possible to combine the above two concepts and determine the values of the individual principal curvature at the specular point. Further, by judicious use of differential geometry (e.g. Minkowski's formulation or Christoffel-Hurwitz formulation, etc.), these curvature values may yield the actual profile of the scatterer.

It must be pointed out here that the phase-curvature relationship concept does require a smooth, convex, well-behaved surface structure. If there are edge or other singular types of source contributions to the backscattered signal, curvature recovery with the method of this thesis will not be accurate. However, in the most recent development of radar target discrimination [5] the relative phase difference of the like polarized elements has been found to provide one of the most important classifiers for discriminating a smoothly (undulating) curved surface from a discontinuously rough surface with sharp edges.

It is found that there are several probable reasons why the 2-4 GHz experimental data give more accurate results than the 4-8 GHz data.

They are the tangent function in (3.10), the k factor and the relative phase error, which all become more significant at higher frequencies. Thus, the phase measurement needs to be very accurate and more accurate methods of obtaining the complete broadband polarimetric scattering data should be investigated.

It seems that not only is the 2-4 GHz range valid for the first order correction to physical optics for a 6"x12" prolate spheroid, but it is also a compromise range between the high frequency condition required by the curvature recovery model and the drawback to lower frequencies required to prevent critical magnification of measurement errors. At too high a frequency, the product of k and the tangent function will lead to erroneous results from measurement data.

Although the curvature recovery model is restricted to conducting scatterers, it may be possible to extend the model to dielectric scatterers, since the space-time integral equation, on which the model is based, has recently been extended and applied to dielectric bodies by Mieras and Bennett [37].

in concluding, both the amplitude difference and the phase difference of like-polarized elements of the scattering matrix in linear polarization tend to zero at high frequencies, yet the phase difference, however small, does contain curvature information of the scatterer. Regardless of the type of orthogonal polarization bases, the phase sum tends to a constant value which is twice the argument of the Fourier transform of the silhouette area of the target. The phase sum also tends to or equals the argument of the scattering ratio defined in this thesis. The magnitude of the

scattering ratio, whose definition is independent of whether linear, circular or general elliptic polarization is used, approaches 0.5 rapidly as frequency is increased. The magnitude of the ratio is interpreted as the ratio of the maximum radar cross section to the trace of the power scattering matrix. The complex plots of the scattering ratio provide a simple check on the accuracy of high frequency polarimetric measurements. Another curvature recovery equation has been derived in circular polarization basis vector notation. The curvature recovery model is proven to satisfy the image reconstruction identities of invariant transformation. Finally, the values of kb from 3.19 to 6.38 have been found to be most potentially suitable for curvature recovery of the 6"x12" prolate spheroid (and probably targets of similar size and shape), provided that polarimetric measurements can be improved to a better accuracy, and that further correction to physical optics approximation can be obtained.

APPENDIX I

DERIVATIONS OF EXPRESSIONS FOR PRINCIPAL CURVATURES FOR PROLATE SPHEROIDS

Let the semi-major axis of the prolate spheroid be C, and the semi-minor axes be a and b, such that C > a = b.

Let $\vec{r}(u,v)$ be a curvilinear mapping to represent the surface of the prolate spheroid. A set of parametric equations are

 $x = a \sin u \cos v$

y = b sin u cos v

 $z = C \cos u$

In vector form,

$$\vec{r}$$
 = (a sin u cos v) \hat{i} + (b sin u sin v) \hat{j} + (C cos u) k

$$\frac{\partial \vec{r}}{\partial u}$$
 = (a cos u cos v, b cos u sin v, -C sin u)

$$\frac{\partial \vec{r}}{\partial v} = (-a \sin u \sin v, b \sin u \cos v, 0)$$

By definition,

$$E = \frac{\partial \vec{r}}{\partial u} \cdot \frac{\partial \vec{r}}{\partial u}$$

$$= a^2 \cos^2 u + c^2 \sin^2 u$$

$$G = \frac{\partial \vec{r}}{\partial y} \cdot \frac{\partial \vec{r}}{\partial y}$$

$$= a^2 \sin^2 u$$

$$F = \frac{\partial \vec{r}}{\partial u} \cdot \frac{\partial \vec{r}}{\partial v}$$

F = 0 implies that the u, v parametric curves are orthogonal to each other.

Let n be the unit vector normal to the surface.

$$n = \frac{3r}{3u} \times \frac{3r}{3v} / \frac{3r}{3u} \times \frac{3r}{3v}|$$

$$= \frac{(bCsin^{2}u \cos v, aCsin^{2}u \sin v, absin u \cos u)}{(b^{2}C^{2}sin^{4}u \cos^{2}v + a^{2}C^{2}sin^{4}u \sin^{2}v + a^{2}b^{2}sin^{2}u \cos^{2}u)}$$

$$\frac{2r}{3u^{2}} = (-a \sin u \cos v, -b \sin u \sin v, -C \cos u)$$

$$n \cdot \frac{2r}{3u^{2}} = \frac{-aC}{(C^{2}sin^{2}u + a^{2}cos^{2}u)}$$

$$n \cdot \frac{2r}{3v^{2}} = \frac{-aCsin^{2}u}{(C^{2}sin^{2}u + a^{2}cos^{2}u)}$$

$$\frac{3^{2}r}{3v^{2}} = (-a \cos u \sin v, b \cos u \cos v, 0)$$

$$\frac{3^2 r}{3 \sqrt{3} u} = (-a \cos u \sin v, b \cos u \cos v, 0)$$

$$\hat{n} \cdot \frac{3^2 \hat{r}}{3 \sqrt{3} u} = 0$$

By definition,

$$L = \hat{n} \cdot \frac{3^{2} \hat{r}}{\partial u^{2}}$$

$$= \frac{-aC}{\sqrt{a^{2} \cos^{2} u + C^{2} \sin^{2} u}}$$

$$M = \hat{n} \cdot \frac{3^{2} \hat{r}}{\partial v \partial u}$$

$$= 0$$

$$N = \hat{n} \cdot \frac{\partial^2 \hat{r}}{\partial v^2}$$

$$= \frac{-aC\sin^2 u}{\sqrt{a^2\cos^2 u + C^2\sin^2 u}}$$

M = F = 0 implies that the lines of curvatures are the u, v parametric curves chosen. Thus,

$$K_{u} = \frac{L}{E}$$

$$= \frac{-aC}{(a^{2}\cos^{2}u + C^{2}\sin^{2}u)^{3/2}}$$

$$K_{v} = \frac{M}{G}$$

$$= \frac{-C}{a\sqrt{a^{2}\cos^{2}u + C^{2}\sin^{2}u}}$$

and the Gaussian curvature

$$K = K_{u} K_{v}$$

$$= \frac{1}{a^{4}c^{2}\{\frac{x^{2}}{4} + \frac{y^{2}}{4} + \frac{z^{2}}{c^{4}}\}^{2}}$$

In terms of x, y and z,

$$K_{u} = \frac{\frac{-ac}{c^{2}}}{\left[\frac{c^{2}}{a^{2}} \times^{2} + \frac{c^{2}}{a^{2}} y^{2} + \frac{a^{2}}{c^{2}} z^{2}\right]^{3/2}}$$

$$K_{v} = \frac{-c}{a\left[\frac{c^{2}}{2} \times^{2} + \frac{c^{2}}{2} y^{2} + \frac{a^{2}}{c^{2}} z^{2}\right]^{1/2}}$$

Without loss of generality, consider the x-z plane:

The point P(x,z) with an aspect angle ϕ , as shown in Figure A(1), can be expressed in terms of ϕ , by solving

$$x = z \tan \phi$$

and

$$\frac{x^2}{a^2} + \frac{z^2}{c^2} = 1$$

Thus,

$$z = \frac{aC}{\sqrt{\{c^2 \tan^2 \phi + a^2\}}}$$

$$x = \frac{aC \tan \phi}{\sqrt{\{C^2 \tan^2 \phi + a^2\}}}$$

Consequently,

$$K_{u} = \frac{-aC}{\left[\frac{C^{4} \tan^{2} \phi + a^{4}}{C^{2} \tan^{2} \phi + a^{2}}\right]}$$

$$K_{v} = \frac{-C}{a\sqrt{\frac{C^{1}\tan^{2}\phi + a}{C^{2}\tan^{2}\phi + a^{2}}}}$$

For a 2:1 prolate spheroid, C=2a. As an example, suppose $\phi=90$ degrees, hence

$$K_{u} = -\frac{1}{4a}$$

$$K_{V} = -\frac{1}{a}$$

$$\frac{K_u - K_v}{2} = \frac{3}{8a}$$

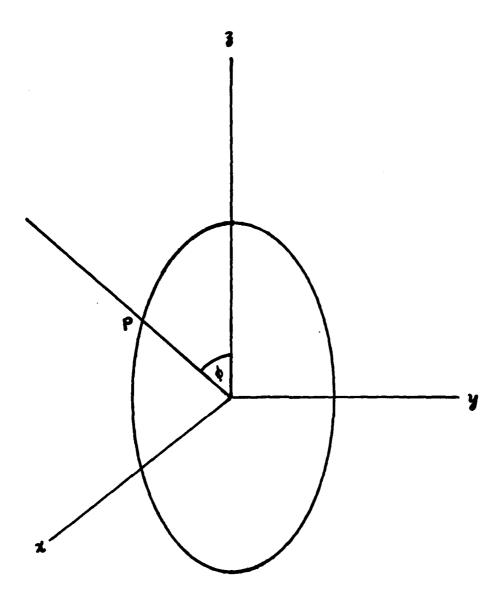


Figure A(1) Geometry of Prolate Spheroid for Curvature Calculation

Since k, the wave number, has been normalized in the phase-curvature relationship, a is set to unity here in conformity.

Thus,

$$\frac{K_u - K_v}{2} = 0.375$$

for broadside incidence.

Other values of curvature difference can be similarly evaluated once the aspect angle $\boldsymbol{\varphi}$ is specified.

APPENDIX II

COMPUTER PROGRAMS

- (a) for computing curvature difference and plotting Figure 5.1(a)
- (b) for plotting the scattering charts with theoretical and experimental data
- (c) for plotting the scattering ratio D

```
RELEASE 2.0
                              MAIN
                                                     DATE = 82134
                                                                                10/45/20
    THEORETICAL VERIFICATION. DATA FROM V90.THEORY.DATA
AND H90.THEORY.DATA
    ASPECT = PHI
          CIMENSION PHIV(180), PHIH(180), RHS(182)
          REAL K(182). KO.KS
          C=1.
          A=C
          4= 2.0 #C
          N=180
          K0=0.1
          KS =0 .1
          PI =4 .0 + AT AN(1 .0)
          THIDEG= 90.
          THI=THIDEG=(PI/180.)
          PHIDEG=90.
          PHI=PHIDEG + (PI/180.)
      READ (8.15) (PHIV(I). I=1.N)
READ (9.15) (PHIM(I). I=1.N)
15 FURMAT(3(10X.FL0.3))
          PRINT 1
         FOFWAT( *0* . 10X . *PHI V* . 36X . *PHIH* . *
                                                                IN RAC!)
       DO 3 I=1.N

3 PRINT 4. PHIV(I). PHIH(I)

4 FORMAT(*0*. 10x. F10.3. 30x. F10.3)

ZT=SIN(THI)*CCS(PHI)
          ET=SIN(THI) #SIN(PHI)
          AK=COS(THI)
          E1 =8+ 3+ SIN(PHI) +SIN(PHI)+ A+ A+COS(PHI) +COS(PHI)
          C1 = (A + A - B + B) + SIN(PHI) + CDS(PHI)
          D1 = A + A + S IN (PHI) + SIN(PHI) + B + B + COS(PHI) + COS(PHI)
          G1 = 4 + 4 + B + B / B1
          HI=C1+C1/(81+81)-01/81
          TAU=ABS(A+A+B+B+B1+SIN(THI)+SIN(THI)/(C1+C1-B1+D1)-C+C+CDS(THI
        1) #COS(THI))
TAU=SQRT(TAU)
          H1=485(H1)
          GG=G1+C/(4+SQRT(H1))
          GG=GG/(TAU++3)
          P= SQR T( 4*4 *Z T*Z T+8 *B *E T*E T+ C* C* 4K* 4K)
          D1 2=(((A*ZT)**2)*(3*B-C*C)+((B*ET)**2)*(4*4-C*C))**2
D1 2=D12+(((B*ET)**2)*(4*4-C*C)+((C*4K)**2)*(4*4-B*B))**2
          DI 2=D12-((3*ET)**4)*((A*A-C*C)**2)-2*((A*ZT)**2)*((C*AK)**2)*(*C*AK)**2)*
         1 日 年 月 ) 年 (日 丰 日 一 C 半 C )
          D12=SQRT(D12)+P/((4+8+C)++2)
          C12=D12*8/2.
PRINT 99. D12
      99 FORMAT( 101 . .
                             EXACT CURVATURE DIFFERENCE BY 2 IS . F10.6)
          PRINT
       5 FORMAT( "U" + 8X + "NORMALIZED K" + 30X + "CURVATURE DIFFERENCE BY 2 FR
         COM PHASE!)
          00 6 I=1.N
          K(1)=K0+(1-1)+KS
          ARG=(PHIH(I) - PHIV(I))/2.
RHS(I)=K(I)+TAN(ARG)
       6 CONTINUE
```

```
DO 7 I=1.N
7 PRINT 8.K(I). RHS(I)
8 FCFMAT(*0*. 10X. F9.2. 30X. F15.6)
PRINT 9. PHIDEG
9 FORMAT(*0*. *THE ASPECT ANGLE HAS BEEN *.F10.6. * DEG*)
PRINT 12. THIDEG
12 FORMAT(*0*. *STARTING NORM K = *.F10.5. *. STEP= *.F10.5)
PRINT 10. KO. KS
10 FORMAT(*0*. *STARTING NORM K = *.F10.5. *. STEP= *.F10.5)
CALL SCALE(*12.0.180.1)
CALL SCALE(*12.0.180.1)
CALL SCALE(*13.8.0.180.1)
CALL AXIS(0.0.0.0.*NORM K*.-6.12.0.0.0.K(181).K(182))
CALL AXIS(0.0.0.0.*NORM K*.-6.12.0.0.0.K(181).K(182))
CALL LINE(*.FH5.180.1)
PRINT 2221.K(181).K(182).RHS(181).RHS(182)
2221 FORMAT(*0*.4F15.5)
CALL ENDPLT
222 FORMAT(*0*.4F15.5)
STOP
END

EFFEC** NOTER M. ID. EBCDIC. SOURCE.NOL IST.NODECK.LOAD.NOMAP.NCTEST
STURCE STATEMENTS = **70.PROGRAM SIZE = **5908**
NC DIAGNOSTICS GENERATED
```

KR7(131)=-2.7 KI7(181)=-2.7 KR7(182)=0.5 KIF(132)=0.5

CALL ENDPLT

CALL LINE(KRP, <1P, 130, 1.0, 4)

CALL LINE(K, K??.130.1.0.4)
CALL ENDELT

CALL HIGHPLT(5, REALPT)

```
CALL BGNPLT(4. [NAG]
CALL SCALE(K:12.9.180.1)
CALL SCALE(K:12.9.180.1)
CALL SCALE(K:12.9.180.1)
CALL AXIS(0.0.).0.*NORM K*.-6.12.0.0.0.K(181).K(192))
CALL AXIS(0.0.).0.*KIM*.3.8.0.90.0.KIP(181).KIP(182))
CALL LINE(K.K:12.180.1.0.4)
CALL ENDPLT
PRINT 553. THIDES
FORMAT(*0*, *ANGLE THI =*.F8.3. * DEGREES*)
PRINT 555. PHIDES
FORMAT(*0*, *ITHE ABDECT ANGLE = *.F8.3.* DEGREES*)
PRINT 556. KO.K3
SEE FORMAT(*0*, *STARTING NORMALIZED K =*.F10.5.* STED=*.F10.5)
PRINT 991
Y41 FORMAT(*0*, *ITHE DRIGINAL VERTICAL DCLARIZATION DATA:*)
PRINT 993
CM3 FORMAT(*0*, *ITHE DRIGINAL HORIZONTAL POLARIZATION DATA:*)
DRINT 993
CM3 FORMAT(*0*, *ITHE DRIGINAL HORIZONTAL POLARIZATION DATA:*)
TOTO
ENO
```

```
RELEASE 2.0
                                                                            DATE = 82134
                                                                                                                   11/02/31
      TIMENSICN AMPV(203), AMPH(203), PHIV(200), PHIH(200)

REAL K(202), RP(202), IP(202), KRP(202), KIP(202)

CJMPLEX CJ. CMPLK. CFXP. RNUM. RDEN, F(200)
              N=2 10
              C11=3.
              C=CIN+2.54/103.
              3=C
A=2.#C
              AIN=2.#CIN
              F0=2.
              S=0.01
CJ=C4PLX(0.0. 1.0)
              PI= . . *A TAN(1.)
THI DEG=90.
              PHIDEG=90.
              READ(3.1)(AMPV(1).7HIV(1).1=1.N)
              READ(9.1)(ANDH(1).PHIH(1).I=1.N)
FCRMAT(10F10.3)
              7C 2 I=1.N
              K(I)=(2.*PI/).7)*(F)+(I-1)*5)*B
PDIFF=PHIH(I)-PHIV(I)
AM=10.**((AMPH(I)-AMPV(I))/10.)
PNJ4=1.-AM*CEXP(CJ*PDIFF)
              TOE N=1.+AM*CEXP(CJ* 7DIFF)
              F(1)=RNUM/RDEN
              RP(I)=REAL(F(I))
              IP(I) = AIMAG(F(I))
              KRP(1)=K(1)*3>(1)
              KIP(1)=K(1)+1P(1)
           S CONTINUE
              PRINT 3
           3 FORMAT(101.1NG-MALIZED K1.19x, TREAL1.19x, TIMAGINARY1.25x, TKPEAL1.
            C19x. 'KIMAG')
      DO 4 I=1.N
4 PRINT 5. K(I). RP(I). IP(I). KRP(I). KIP(I)
5 FORMAT(*0*.F10.5.21X.F14.7.5X.F14.7.13X.F14.7.15X.F14.7)
CALL 3GNPLT(5. CHART)
UNIFIED SCALE FOR ALL IMAGINARY-PEAL CLUSTER PLOTS NOW USED
THER PLOTS HAVE NO UNIFIED SCALE
TO REMOVE UNIFIED SCALE
TO REMOVE UNIFIED SCALE FOR IMAG-REAL CLUSTER PLOTS.
JUST DELETE THE FOLLOWING 4 FORTRAN STATEMENTS.
AND INSERT 2 SUBFOLTINE SCALE. AS FOLLOWS:
CALL SCALE(KRP.8.0.200.1)
CALL SCALE(KRP.8.0.200.1)
      CALL SCALE(KIP.8.3.200.1)
              KRP(201)=-8.0
              KIP(201)=-8.0
              KAP(232)=2.0
              KIP(202)=2.0
              CALL AXIS(0.0.4.0. KPEAL' .-5.8.0.0.0.KRP(201),KPP(202))
CALL AXIS(4.0.1.0. KIMAG' .5.8.0.90.0.KIP(201),KIP(202))
              CALL LINE(KRP. KIP. 200.1.0.4)
              CALL ENDPLT
              CALL BGNPLT(6, REALPT)
              CALL SCALE (K. 12.0.200.1)
              CALL SCALE (KRP.3.0,200.1)
```

```
CALL AXIS(0.0.).0.*NORM K*.-6.12.0.0.0.*K(201).K(202))
CALL AXIS(0.0.).0.*KRE*.3.8.0.90.0.K3P(201).KRP(202))
CALL LINE(K.KR3.200.1.-1.4)
CALL EMDPLT
CALL BOOPLT
CALL BOOPLT
CALL SCALE(K:12.0.200.1)
CALL SCALE(K:12.0.200.1)
CALL SCALE(K:12.0.200.1)
CALL AXIS(0.0.).0.*VIM*.3.9.0.90.0.KIP(201).K(202))
CALL AXIS(0.0.).0.0.*VIM*.3.9.0.90.0.KIP(201).KIP(202))
CALL LINE(K.KIP.200.1.-1.4)
CALL ENDPLT
PRINT 554. CIN.AIN

354 FORMAT(*0*.*PROLATE SPHEROID. 3=C=*.F7.3.* A=*.F7.3.* INS*)
PPINT 553. THI)EG

553 FORMAT(*0*.*PROLATE SPHEROID. 3=C=*.F7.3.* A=*.F7.3.* INS*)
PRINT 555. PHI)EG

555 FORMAT(*0*.*ITHE ASPECT ANGLE = *.F8.3.* DEGREES*)
PRINT 556. FO.3

556 FORMAT(*0*.*ITHE ASPECT ANGLE = *.F8.3.* DEGREES*)
PRINT 991

GG1 FORMAT(*0*.*ITHE DRIGINAL VERTICAL POLARIZATION CATA:*)
PRINT 992. (AMPV(I).PHIV(I).I=1.N)

992 FORMAT(*0*.*ITHE DRIGINAL HORIZONTAL POLARIZATION CATA:*)
PRINT 994. (AMPV(I).PHIH(I).I=1.N)

394 FORMAT(*0*.*ITHE DRIGINAL HORIZONTAL POLARIZATION CATA:*)
PRINT 994. (AMP*(I).PHIH(I).I=1.N)

395 FORMAT(*0*.*ITHE DRIGINAL HORIZONTAL POLARIZATION CATA:*)
PRINT 994. (AMP*(I).PHIH(I).I=1.N)

396 FORMAT(*0*.*ITHE DRIGINAL HORIZONTAL POLARIZATION CATA:*)
PRINT 994. (AMP*(I).PHIH(I).I=1.N)
```

DATE = 82134

11/02/01

```
N - NO. OF FREQUENCIES OR DATA
                   DIMENSION AMPV(19J), AMPH(193), PHIV(190), PHIH(190)
PEAL K(182), KU,KS, RP(192), IP(192), AMP(182), PHASE(192)
COMPLEX CJ. CMPLX, CEXP. RNUM, ROEN, RATIC(190)
                    N=1 10
                    KU= ) . L
                    K 5 = 0 . 1
                   CJ=CMPLX(0.0. 1.0)
P[=4.0ATAN(1.)
                    THIDEG= 10.
                   PHIDEG=90.
                   READ(8.1)(AMPV(1).PHIV(1).I=1.N)
                   READ(9.1)(AMPH(1).7HIH(1).1=1.N)
             1 FORMAT(6F10.6)
                   PRINT 388
       388 FORMAT( 101. INCRMALIZED KI. 3X. IREALI. 15X. IMAGINARYI. 10X. 12HASE! C.14X. AMPLITUSE!)
                    75 2 (=1.N
                    K([]=KO+([-1]#<5
                   PNUM=AMPH(1) *CEXP(CJ*PH[H(1)) *AMPV(1)*CFXP(CJ*PH[V(1)) *CFXP(CJ*PH[V(1)) *CFXP(CJ*
                   FATIC(1)=RNUM/RJEN
                   FP(I)=FEAL(RATIJ(I))
IP(I)=AIMAG(RATID(I))
                    AMM3G=RP([]++2+1P([)++2
                    AMP( [ )=SORT( AVPS ])
                    TAN= [P([]/PP([]
                   PHASE(1)=ATAN(TAN)
                   FRINT 999.K(1).PP(1).IP(1).THASE(1).AMP(1)
       2 CONTINUE
209 FERMAT(*0*,F10.3,4(5X,F14.7))
                    CALL BUNPLT(5.RATIO)
                    PP(131)=-1.0
                    [P(181)=-1.0
                     79(132)=0.25
                    IP(132)=0.25
                    CALL AXIS(0.0...0. REAL .-4.8.0.0.0. RP(181); RP(182))
CALL AXIS(4.3.). 3. "IMAG".4.8.0. JO.O. IM(181). IM(182))
                    CALL LIME (RP. 19.180.1.1.4)
                    CALL ENUPLT
                    CALL BUNPLT ( 3. RAT)
                    AVP(181)=0.0
                    AMP(182)=0.5
                    K(131)=0.0
                    K(182)=2.5
                    CALL AXIS(0.).).0.*NORM K**-5.8.0.0.0.K(181).K(182))
CALL AXIS(0.0.).0.*14.4PLITUDE*.9.2.0.70.0.AMF(181).AMP(182))
CALL LINE(K.AMF.180.1.0.4)
                    CALL ENOPLT
        PRINT 553, THIDEG

PRINT 553, THIDEG

PRINT 553, PHIDEG

PRINT 554, PHIDEG
       JUS FERMAT('0'.'THE ASPECT ANGLE = '.F9.3.' DEGREES')
PRINT 556. KO.KS
DEC FORMAT('0'.'STARTING NORMALIZED K ='.F10.5.' STEP='.F10.5)
```

PRINT 991
391 FORMAT('0'. 'THE URIGINAL VERTICAL PCLARIZATION DATA:')
PRINT 992. (AMPV(I).PHIV(I).I=I.N)
992 FORMAT('0'. 10F10.3) PRINT 993
1993 FORMAT(*0*.*THT DRIGINAL HORIZONTAL POLARIZATION DATA:*)
PRINT 994.(AMP4(I).PHIM(I).I=1.N)
194 FORMAT(*0*. IJFIU.3)
STOP

EFFECT* NOTERM.ID.EDCDIC.SOUPCE.NOLIST.NODECK.LOAD.NOMAP.NOTEST EFFECT* NAME = MAIN . LINECNT = 50 SOURCE STATEMENTS = 63.PROGRAM SIZE = 10518 NO DIAGNOSTICS GENERATED

END

APPENDIX III

THEORETICAL DATA

(obtained from Dr. Sujeet Kumar Chaudhuri, Department of Electrical Engineering, University of Waterloo, Waterloo, Ontario, Canada, 1981)

THE ASPECT ANGLE = 90.000 DEGREES

STARTING NORMALIZED K = 0.10000 STEP = 0.10000

THE DRIGINAL VERTICAL FCLARIZATION DATA:

	0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0
1	0.0822
•	1.105
917 -2.467 043 2.728	2.901 1.043
003 1.875	2.06? 8.003 8
•	950-1
·	0.953
988 -3.100	-1.533 1.034 -2.931 0.988
	0.990
971 0.156	1.354 1.028 0.371 C.971
	1.027
7867 -1.807	-1.604 0.987 -
,	-2.643 0.999 -
013 2.477	2.663 1.013
186	1.661 0.981
020 0.461	0.65 ! 1.020
988 -0.519	-0.31 i 0.988
004 -1.545	-1.346 1.004 -
005 -2.512	-2.317 1.005 -
988	2,9! 3 0,988
* 10	1.451 1.014
989	0.972 0.989

1.007 -1.023 1.002 -1.223 0.996 0.990 -2.019 0.993 -2.242 0.996 1.010 -3.011 1.010 3.054 1.007 0.993 2.234 0.990 2.055 0.990 1.725 -2.504 1.997 -2.457 0.606 1.725 -2.504 1.997 -2.459 1.960 1.505 -2.34 1 1.239 -2.520 1.026 1.049 2.065 1.307 2.483 1.356 1.049 2.065 1.307 2.483 1.356 1.049 2.065 0.907 1.817 0.901 1.114 -1.214 1.113 -1.348 1.066 0.924 -2.176 0.954 -2.436 1.015 1.009 -3.124 1.072 3.002 1.015 1.009 -0.011 1.019 -0.116 1.016 1.000 -0.011 1.019 -0.116 1.016 1.000 -0.011 1.009 -2.009 1.055 0.965 -2.627 0.967 -1.023 0.959 1.039 2.409 1.006 1.223 1.006 0.966 -2.527 0.966 1.223 0.961 1.007 -1.517 1.014 -1.745 0.995 0.966 -2.527 1.000 2.543 0.995 1.016 2.740 1.010 1.000 2.543 0.995 0.996 -2.527 1.000 2.543 0.995 0.996 -2.527 1.000 2.543 0.995	701.0			•	-0.247		-0.443	1.013	-0.635
0.990 -2.014 0.993 -2.242 0.998 1.010 -3.011 1.010 3.054 1.007 0.993 -2.234 0.990 2.055 0.990 POLARIZATION DATA: 0.990 -2.857 0.990 1.725 -2.969 0.490 -2.857 0.906 1.506 -2.344 1.239 -2.837 0.906 1.049 2.065 0.907 1.987 1.986 1.049 2.065 0.907 1.124 0.903 1.049 2.065 0.907 1.124 0.903 1.114 -1.211 1.113 -1.346 1.016 0.924 -2.043 0.956 -2.436 1.016 1.1109 -3.123 1.072 3.002 1.016 1.1093 -3.123 1.013 1.016 1.016 1.1094 -1.211 1.011 1.016 1.016 1.000 -3.123 1.023 1.016 1.001 -2.	-0-830	1.007	-1.025	1.002	-1.223	966*0	-1.424	166*0	-1.627
1.010 -3.011 1.010 3.054 1.007 0.993 2.234 0.990 2.055 0.990 POLARIZATION DATA: 0.238 -2.969 0.490 -2.857 0.8066 1.725 -2.503 1.897 -2.439 1.960 1.066 2.665 1.307 2.483 1.326 1.049 2.065 0.907 1.817 0.843 1.049 2.065 0.907 1.817 0.803 1.049 -2.176 0.954 -2.436 1.016 1.114 -1.214 1.113 -1.348 1.068 0.924 -2.176 0.954 -2.436 1.016 1.009 -3.124 1.072 3.002 1.016 1.009 -3.129 0.997 1.013 0.954 1.000 -0.031 1.019 -0.116 1.048 1.000 -0.031 1.019 -0.116 1.048 1.002 -1.911 1.048 -2.089 1.055 0.965 -2.823 1.046 2.233 1.006 0.967 1.449 0.966 1.223 0.991 1.039 2.409 1.046 2.233 1.006 0.968 -0.577 0.980 -0.795 1.001 1.027 -1.537 1.014 -1.745 0.995 0.946 -2.567 1.002 -2.794 1.019 1.016 2.740 1.002 2.543 0.978 0.998 0.743 1.001 1.001 1.001	-1.832	0.660	-2.019	0.993	-2.242	0.998	-2.444	1.004	-2.643
0.993 2.233 0.990 2.055 0.990 POLARIZATION DATA: 0.490 -2.857 0.806 - 0.238 -2.969 0.490 -2.857 0.806 - 1.725 -2.343 1.239 -2.439 1.960 - 1.505 -2.343 1.239 -2.520 1.026 - 1.026 1.049 2.065 0.907 1.817 0.813 1.356 1.026 1.026 1.049 2.065 0.907 1.817 0.901 1.124 0.901 1.014 0.901 1.014 0.901 1.014 0.901 1.014 0.901 1.014 0.901 1.014 0.901 1.014 0.901 1.014 0.902 1.014 0.902 1.014 0.902 1.014 0.902 1.014 0.902 1.014 0.902 1.014 0.902 1.014 0.902 1.014 0.902 1.002 0.902 1.002 0.902 1.002 0.902 1.002	-2.835	1.010	-3.031	010.1	3.054	1.007	2.859	1.002	2.662
0.238 -2.969 0.490 -2.857 0.806 - 1.725 -2.963 0.490 -2.857 0.806 - 1.505 -2.313 1.897 -2.439 1.960 - 1.506 -2.323 1.307 2.483 1.960 - 1.186 -2.045 1.307 2.483 1.356 - 1.049 2.045 0.907 1.124 0.903 1.026 - 0.903 1.124 0.901 1.124 0.901 1.014 - 1.124 0.906 1.014 - </td <td>2.462</td> <td>0.493</td> <td>2.233</td> <td>0.66.0</td> <td>2.055</td> <td>0.66.0</td> <td>1.850</td> <td>0.993</td> <td>1.646</td>	2.462	0.493	2.233	0.66.0	2.055	0.66.0	1.850	0.993	1.646
0.238 -2.969 0.490 -2.857 0.806 1.725 -2.503 1.897 -2.439 1.960 1.505 -2.343 1.837 -2.520 1.026 1.186 2.605 1.307 2.483 1.356 1.049 2.066 0.907 1.124 0.843 1.084 0.734 1.135 0.607 1.124 0.9880 -0.082 0.906 -0.347 0.901 1.114 -1.213 1.113 -1.348 1.068 0.924 -2.176 0.954 -2.436 1.014 1.109 -3.123 1.072 3.002 1.014 1.043 1.209 0.954 -2.436 1.046 1.000 -0.811 0.967 -1.023 0.954 1.002 -2.039 0.967 -1.023 0.954 1.003 -2.823 1.046 1.048 1.039 2.409 1.046 -2.089 1.066 1.039 2.409 1.046 -2.23 1.036 1.039 2.409	ORI ZONTA		_						
1.725 -2.503 1.897 -2.439 1.960 1.505 -2.343 1.239 -2.520 1.026 1.186 2.605 1.307 2.483 1.356 1.049 2.069 0.907 1.124 0.963 1.080 -0.082 0.956 -2.436 1.014 -1.124 1.109 -1.124 1.113 -1.348 1.016 -1.016 1.109 -2.175 0.954 -2.436 1.014 -1.016 1.109 -3.123 1.072 3.002 1.016 -1.016 1.009 -3.123 1.072 3.002 1.016 -1.016 1.009 -3.123 1.019 -0.116 1.016 -1.016 1.009 -0.911 1.019 -0.116 1.048 -2.089 1.048 1.002 -1.911 1.046 -2.089 1.048 -2.089 1.036 1.022 -1.911 1.046 -2.089 1.036 1.036 1.033 2.409 1.046 -2.233 1.036 1.027 <t< td=""><td>1.316</td><td>0.238</td><td>-2.969</td><td>0.490</td><td>-2.857</td><td>908.0</td><td>-2.762</td><td></td><td>-2.572</td></t<>	1.316	0.238	-2.969	0.490	-2.857	908.0	-2.762		-2.572
1.505 -2.34 i 1.239 -2.520 1.026 1.186 2.605 1.307 2.483 1.356 1.049 2.065 0.907 1.817 0.843 1.084 0.730 1.135 0.607 1.124 0.880 -0.062 0.907 1.817 0.901 1.114 -1.213 1.113 -1.348 1.058 0.924 -2.176 0.954 -2.436 1.014 1.109 -3.123 1.072 3.002 1.016 0.954 2.049 0.954 -2.436 1.016 1.009 -3.123 1.002 1.016 1.016 0.954 2.049 0.954 1.016 1.016 1.002 -2.813 1.016 1.048 -2.089 1.058 1.032 -2.823 1.048 -2.089 1.056 1.016 1.032 -2.823 1.048 -2.039 1.056 1.016 1.033 2.409 1.046 2.233 1.036 0.968 -0.577 1.026 0.296	-2.587	1.725	-2.503	1.897	-2.439	1.960	-2, 383	1.905	-2.347
1.186 2.605 1.307 2.483 1.356 1.049 2.065 0.907 1.617 0.663 1.084 0.730 1.135 0.607 1.124 0.880 -0.082 0.956 -2.436 1.016 1.114 -1.213 1.113 -1.346 1.068 0.924 -2.176 0.954 -2.436 1.014 1.109 -3.123 1.072 3.002 1.014 1.009 -3.123 1.072 3.002 1.014 1.004 1.209 0.996 1.873 1.004 1.002 -0.811 0.967 -1.023 0.954 1.002 -2.823 1.048 -2.089 1.048 1.032 -2.823 1.036 1.036 1.032 1.449 0.966 1.223 0.956 1.032 1.449 0.966 1.223 0.966 1.032 0.433 1.026 0.251 1.006 1.027 -1.537 1.014 -1.745 0.995 0.966 -2.567	-2.342	1.505	~2.38 3	1.239	-2.520	1.026	-2.779	0.955	3.140
1.044 2.065 0.007 1.817 0.843 1.084 0.730 1.135 0.607 1.124 0.880 -0.082 0.856 -0.387 0.901 1.114 -1.213 1.113 -1.348 1.068 1.109 -3.123 1.072 3.002 1.015 0.954 2.099 0.996 1.0873 1.042 1.000 -0.813 0.967 1.013 0.954 1.000 -0.813 0.967 -1.023 0.959 1.003 2.409 1.046 2.233 1.036 0.967 1.449 0.966 1.223 0.991 1.032 2.409 1.026 0.251 1.008 0.968 -0.577 0.980 -0.795 1.001 1.027 -1.537 1.014 -1.745 0.995 0.946 -2.527 1.002 -2.794 1.019 1.016 2.743 1.002 2.543 0.995	2.814	1.186	2.605	1.307	2.483	1.356	2.432	1.318	2.327
1.084 0.733 1.135 0.667 1.124 0.880 -0.082 -0.087 0.901 - 1.114 -1.213 1.113 -1.346 1.068 0.924 -2.176 0.954 -2.436 1.014 1.109 -3.123 1.072 3.002 1.014 0.954 2.083 0.996 1.873 1.042 1.000 -0.031 1.019 -0.116 1.042 1.000 -0.813 0.967 -1.023 0.954 1.022 -1.911 1.046 -2.089 1.068 1.039 -2.823 0.967 -1.023 0.959 1.039 2.403 1.046 2.233 1.068 0.967 1.449 0.966 1.223 0.981 1.032 -0.433 1.026 0.251 1.008 0.968 -0.557 0.986 -0.795 1.001 1.016 -2.567 1.014 -1.745 0.995 0.996 0.996 -0.795 1.019 1.016 2.794	2.227	1.040	2.069	0.907	1.817	0.843	1.479	0.885	1.145
0.880 -0.08? 0.886 -0.387 0.901 -1.114 -1.213 1.113 -1.348 1.068 -0.924 -2.175 0.954 -2.436 1.014 1.015 0.954 2.089 1.072 3.002 1.015 0.954 1.002 1.015 1.0043 1.0042 1.005 0.997 1.013 0.954 0.979 0.031 1.019 -0.116 1.0048 1.055 0.984 0.985 -2.823 0.967 -1.023 0.959 1.055 0.968 -2.823 0.967 -3.043 0.959 1.036 0.967 1.032 0.981 1.032 0.980 1.026 0.251 1.008 0.968 -0.577 0.980 -0.795 1.001 1.027 -1.537 1.004 -1.745 0.995 0.995 0.996 0.996 0.996 0.996 0.996 0.996 0.998 0.742 1.001 1.00	0.897	1.084	0.730	1.135	0.607	1.124	0.493	1.058	0.357
1.114 -1.213 1.113 -1.346 1.066 -2.436 1.014 -1.109 -3.123 1.072 3.002 1.015 1.015 0.954 -2.436 1.014 -1.109 -3.123 1.072 3.002 1.015 0.954 0.954 1.072 3.002 1.015 1.045 1.0043 1.0043 1.0043 1.0097 1.0013 0.954 1.006 1.006 1.0050 1.0	0.173	0.880	-0.082	0.856	-0.387	0.901	-0.679	0.987	-0.910
0.924 -2.176 0.954 -2.436 1.014 -1.109 -3.123 1.072 3.002 1.015 0.954 0.954 0.954 1.072 3.002 1.015 0.954 0.954 0.977 1.013 0.954 1.006 1.0097 0.031 1.019 -0.116 1.008 -1.0097 0.957 1.009 0.957 1.009 0.959 1.022 -1.911 1.0048 -2.089 1.055 0.959 1.039 2.409 1.0046 2.233 1.036 0.959 1.035 0.967 1.032 0.967 1.006 0.967 1.039 0.996 1.005 0.996 1.005 0.996 1.005 0.996 1.001 1.027 -1.537 1.004 -1.745 0.995 1.001 1.016 2.740 1.001 2.543 0.996 0.978 0.998 0.742 0.995 0.998 0.742 0.998 0.742 0.998 0.978 0.998	- 1.081	1.114	-1.213	1.113	-1.348	1.068	-1.495	1.001	-1.679
1,109 -3,123 1,072 3,002 1,015 0,954 2,099 1,873 1,0042 1,043 1,201 0,996 1,873 1,0042 1,000 -0,911 1,019 -0,116 1,0048 1,022 -1,911 1,048 -2,089 1,055 0,967 1,046 -2,233 1,055 1,039 2,403 1,046 2,233 1,036 0,967 1,449 0,966 1,223 0,981 1,032 -0,433 1,026 0,251 1,006 0,968 -0,577 0,986 -0,795 1,001 1,027 -1,537 1,014 -1,745 0,995 1,016 -2,567 1,002 -2,794 1,019 1,016 2,743 1,002 2,543 0,995 0,998 0,742 0,995 0,995 0,978	11601-	0.924	-2.176	0.954	-2.436	1.014	-2.658	1.075	-2.636
0.954 2.099 0.996 1.6873 1.0042 1.0043 1.201 0.997 1.013 0.954 0.979 0.031 1.019 -0.116 1.048 1.000 -0.811 0.967 -1.023 0.950 1.022 -1.911 1.048 -2.089 1.055 0.985 -2.821 0.967 -3.043 0.959 1.039 2.409 1.046 2.233 1.036 0.967 1.449 0.966 1.223 0.961 1.032 0.433 1.026 0.251 1.008 0.968 -0.577 0.980 -0.795 1.001 1.027 -1.537 1.014 -1.745 0.995 0.946 -2.567 1.000 2.543 0.995 1.016 2.743 1.012 1.501 1.021 0.998 0.743 1.001 2.543 0.995	-2.986	1.109	-3.123	1.072	3.002	1.015	2.819	0.963	2.598
1.043 1.20) 6.997 1.013 0.954 0.979 0.031 1.019 -0.116 1.048 1.000 -0.81) 0.967 -1.023 0.950 1.022 -1.911 1.048 -2.089 1.055 0.985 -2.823 0.967 -3.043 0.959 1.039 2.409 1.046 2.233 1.036 0.967 1.449 0.966 1.223 0.981 1.032 0.433 1.026 0.251 1.008 0.968 -0.577 0.980 -0.795 1.001 1.027 -1.537 1.014 -1.745 0.995 0.946 -2.527 1.002 2.543 0.985 0.997 1.693 1.012 1.501 1.021	2.348	0.954	2.099	966 0	1.873	1.042	1.684	1.072	1.519
0.979 0.031 1.019 -0.116 1.048 - 1.000 -0.81) 0.967 -1.023 0.950 - 1.022 -1.911 1.048 -2.089 1.055 - 1.039 2.409 1.046 2.233 1.036 0.959 1.035 - 1.032 0.430 1.026 1.223 0.981 1.008 0.968 -0.577 0.980 -0.795 1.001 - 1.027 -1.537 1.014 -1.745 0.995 - 1.016 2.740 1.002 -2.794 1.019 - 1.016 2.740 1.012 1.501 1.021 0.998 0.74? 0.995 0.995 0.995	1.364	1.043	1.20)	166.0	1.013	0.954	0.795	0.933	0.552
1.000 -0.811 0.967 -1.023 0.950 1.022 -1.911 1.048 -2.089 1.055 0.985 -2.087 -3.043 0.959 1.039 2.409 1.046 2.233 1.036 0.967 1.449 0.966 1.223 0.981 1.032 0.433 1.026 0.251 1.008 0.968 -0.577 0.980 -0.795 1.001 1.027 -1.537 1.014 -1.745 0.995 0.946 -2.587 1.002 -2.794 1.019 1.016 2.743 1.012 -2.543 0.985 0.998 0.742 0.985 0.978	0.306	0.979	160.0	1.019	-0-116	1.048	-0.289	1.053	-0.453
1,022 -1,911 1,048 -2,089 1,055 -6,985 0,985 0,985 1,039 2,409 1,046 2,233 1,036 0,989 1,035 0,989 1,032 0,981 1,032 0,981 1,032 0,981 1,032 0,981 1,032 0,981 1,008 0,986 -2,587 1,014 -1,745 0,995 0,985 0,998 0,74? 0,985 0,985 0,998 0,74? 0,985 0,998 0,74? 0,985 0,536 0,978	-0.623	1.000	-0.81)	0.967	-1.023	0.950	-1.255	0.958	-1.491
0.985 -2.823 0.967 -3.043 0.959 1.039 2.409 1.046 2.233 1.036 0.967 1.449 0.966 1.223 0.981 1.032 0.433 1.026 0.251 1.008 0.968 -0.577 0.980 -0.795 1.001 1.027 -1.537 1.014 -1.745 0.995 0.946 -2.567 1.002 -2.794 1.019 1.016 2.740 1.000 2.543 0.985 0.997 1.693 1.012 1.501 1.021	-1.713	1.022	-1.911	1.048	-2.089	1.055	-2.257	1.042	-2.423
1.039 2.409 1.046 2.233 1.036 0.967 1.449 0.966 1.223 0.981 1.032 0.433 1.026 0.251 1.008 0.968 -0.577 0.980 -0.795 1.001 1.027 -1.537 1.014 -1.745 0.995 0.946 -2.527 1.002 -2.794 1.019 1.016 2.743 1.000 2.543 0.985 0.997 1.693 1.012 1.501 1.021	-2.614	0.985	-2.823	0.967	-3.043	0.959	3.011	0.00	2.192
0.967 1.449 0.966 1.223 0.981 1.032 0.433 1.026 0.251 1.008 0.968 -0.577 0.980 -0.795 1.001 1.027 -1.517 1.014 -1.745 0.995 0.946 -2.527 1.002 -2.794 1.019 1.016 2.743 1.000 2.543 0.985 0.997 1.691 1.012 1.501 1.021 0.998 0.74? 0.985 0.536 0.978	2.591	1.039	2.409	1.046	2.233	1.036	2.058	1.012	1.972
1.032 0.433 1.026 0.251 1.008 0.968 -0.577 0.980 -0.795 1.001 - 1.027 -1.537 1.014 -1.745 0.995 - 0.946 -2.527 1.000 -2.794 1.019 - 1.016 2.743 1.000 2.543 0.995 0.997 1.693 1.012 1.501 1.021 0.998 0.74? 0.995 0.536 0.978	1.665	0.967	1.449	0.966	1.223	0.981	1.004	1.004	0.799
0.998	019.0	1.032	0.433	1.026	0.251	1.008	0.063	0.985	-0-139
1.027 -1.537 1.014 -1.745 0.995 0.946 -2.527 1.002 -2.794 1.019 1.016 2.743 1.000 2.543 0.985 0.997 1.693 1.012 1.501 1.021 0.998 0.74? 0.985 0.536 0.978	-0.354	0.968	-0.577	0.980	-0.795	1.001	100-1-	1.020	-1.194
0.946 -2.587 1.002 -2.794 1.019 -1.016 2.743 0.995 0.995 0.997 1.693 1.012 1.501 1.021 0.998 0.74? 0.998 0.536 0.978	-1.376	1.027	-1.537	1.014	-1.745	0.995	-1.943	196.0	-2.150
1.016 2.743 1.000 2.543 0.985 0.997 1.691 1.012 1.501 1.021 0.998 0.74? 0.985 0.536 0.978	-2.372	986.0	-2.587	1.002	-2.794	1.019	-2.988	1.029	3.110
0.997 1.691 1.012 1.501 1.021 0.998 0.74? 0.985 0.536 0.978	2.927	910.1	2.740	1.000	2.543	0.985	2. 336	0.979	2.121
0,998 0,74? 0,985 0,536 0,978	1.906	766.0	1.691	1.012	1.501	1.021	1.312	1.021	1.127
	0.538	866.0	0.74 ?	0.985	0.536	0.978	0.322	0.982	0.108

-0.969	-1.499	-2.863	2.393	1.420	0.193	-0.588	-1.500	-2.595	2.693	1.685
1.013	966.0	1.006	1.904	0.995	1.008	0.989	1.013	0.66.0	1.013	0.992
-0.679	-1.690	-2.669	2 • 591	1.621	0.592	-0-383	-1.406	-2.387	2.885	1.892
1.019	0.985	1.016	0.994	1.005	1.000	966.0	1.007	0.993	1.012	066.0
164.0-	-1.478	-2.480	2.800	1.815	0.795	-0.183	-1.209	-2.183	3.079	5.099
1.016	0.983	1.020	186.0	1.014	0.991	1.005	666*0	0.999	1.007	966.0
-0.363	-1.257	-2.2.2	3.013	2.005	1.003	110.0	-1.005	-1.981	-3.003	2.301
1.006	0.949	1.016	0.987	1.011	986.0	1.001	166.0	1.007	0.999	10001
-0.100	-1.063	5.099	-3.065	2.195	1.213	0.203	-0.797	162-1-	-2.802	5.499
564.0	1.001	1.007	986.0	£10.1	0.987	1.013	0.987	1.013	0.993	1.008

APPENDIX IV

EXPERIMENTAL DATA

(obtained from Dr. Jonathan D. Young, Electro-Science Laboratory, Department of Electrical Engineering, Ohio State University, Columbus, Ohio, 1982)

	0.569 28.253 0.567 24.935 0.774 29.350 0.915 29.186 1.068 29.122 1.202 29.487 1.351 28.572 1.572 24.974 1.616 27.986
	28.185 29.782 29.421 29.402 29.194 29.194 29.147 28.946
	0.529 2 0.647 2 0.709 0.752 0.884 1.039 1.319 1.319
	28.143 28.627 29.333 29.447 29.216 29.168 29.000 28.267
0.01C0 GHZ	
0.016	0.625 0.625 0.625 0.735 0.735 1.003 1.1471
, 11	28.119 28.119 28.218 29.218 29.253 29.253 29.263
oo step	ב איז
THE ASPECT ANGLE # 90.000 JEST-	

	27.893	1.882	27.900	1.903	27.909	1.930	27.915	1.951	27.916	1.971
6	27.910	1.991	106.72	2.013	27.895	2.038	27.893	2.068	27.897	2.103
	27.908	2.134	27.926	2-170	27.944	2.206	27.964	2.242	27.990	2.279
ť	24.022	2.315	28.063	2 • 35 2	28.110	2.388	28.168	2.422	28.213	2.455
	28.298	2.488	28.361	2.513	28.422	5.549	28.479	2.578	28,529	2.605
•	28.573	2.634	28.621	2,661	28.669	2.687	28.712	2.714	28.747	2.741
	28.779	2.769	28.806	2.754	28.828	2.829	28.849	2.860	28.871	2.691
•	28.891	2,921	28.910	2.953	28.925	2.983	28.935	3,013	28.937	3.043
	28.931	3.073	28.918	3.102	28.897	3.130	28.875	-3.127	28.853	-3.102
•	28.835	-3.080	28.839	-3.062	28.882	-3.040	28.956	-3.021	29.052	-3.001
	29.160	-2.982	29.269	-2.963	29,365	-2.938	29.436	-2.913	29.478	-2.885
:	29.489	-2.856	59.469	-2.825	29.418	-2.791	29.343	-2, 756	29.250	-2.719
	29.144	-2.681	29.032	-2.64 1	28.920	-2.605	28.822	-2.568	28.742	-2.533
•	28.680	-2.500	28.637	-2.47)	28.598	-2.442	28.549	-2.413	28.487	-2.384
	20.413	-2.354	28.336	-2.324	28.256	-2,293	28.172	-2.262	28.093	-2.23
,	28.028	-2-194	27.980	-2.151	27.947	-2.120	27.930	-2.081	27.933	-2.042
	27.953	-2.002	27.976	-1.961	27,995	-1,928	28.009	-1.893	28.009	-1.859
,	27.994	-1.825	27.969	-1.791	27.942	-1.756	27.918	-1.720	27.896	-1.683
	27.883	-1.646	27.886	-1.663	27.903	125.1-	27.934	-1.533	27.982	-1.196
	28.047	-1.460	28.118	-1.423	28.188	-1.392	28.260	-1 - 359	29.326	-1.329
	26.378	- 1 - 30 2	28.412	-1.277	28.431	-1.253	28.437	-1.230	28.435	-1.207
,	28.427	-1.184	28.418	191-1-	28.415	-1.136	28.425	-1.109	28.447	-1.081
	28.483	-1.051	28.535	-1.023	28.600	-0.987	28.672	-0-954	28.749	-0.921
,	28.826	-0.888	28.898	-0.857	28.959	-0.826	29.002	-0. 797	29.021	-0.770
	29.029	-0.745	29.031	-0.723	29.030	-0.698	29.023	-0.676	29.013	-0.655
,	24.998	-0.635	28.974	-0.614	28.941	-0.592	28.899	-0.570	28.845	-0.544
	28.778	-0.516	28.701	-0.483	28.621	-0.453	28.543	-0.418	28.473	-0.383
,	28.411	-0.347	28.359	-0.310	28.319	-0.274	28.290	-0.238	28.271	-0.204
	28+253	-0.171	28.234	-0-137	28.214	-0-102	28.188	-0.066	28.160	-0.030
,	28.135	0 • 00 5	28.119	0.035	28.112	0.063	28.108	0.084	24.103	0.097

THE URIGINAL HORIZGNTAL POLARIZATION DATA:

	29.022	0.354	29.006	0.355	28.964	0.364	28.901	0.378	28.819	101.0
•	28.730	0.421	28.649	0.441	28.592	0.478	2A.566	0.508	29.559	0.537
	28.560	0.563	28.576	0.585	28.611	0.609	28.655	0.631	28.732	0.654
•	28.798	0.678	28.855	0.703	28.904	0.736	28.950	0.170	28.999	0.808
	23.049	0.846	29.105	0.842	29.171	916.0	29.246	0.948	29.317	0.977
(•	29.373	1.000	29.408	1.013	29.418	1.032	29.399	1.045	29.348	1.058
	23.278	1.073	29.199	1.092	29.114	1.116	29.026	:::	28.943	1.175
. •	28.874	1.208	28.825	1.243	28.799	1.271	28.796	1.297	24.912	1.317
	28.839	1.331	28.873	1.341	28.904	1.349	28.920	1.357	28.906	1 - 369
	28.861	1.386	28.790	1.410	28.700	1.441	28.601	1.482	29.503	1.510
	28.414	1.585	28.340	1.645	28.287	1.709	28.255	1.773	28.242	1.635
	28.245	1.892	28.260	1.94	28.280	1.988	28.299	2.024	28.310	2.053
	28.307	2.079	28,285	20102	28.246	2.124	28.196	2.149	28,139	2.175
	28.084	2.20€	28.040	2.238	28.022	2.276	28.028	2.317	28.057	2.359
	28.107	2.402	28.172	2.44.3	28,245	2.477	28.313	2 • 506	28.370	2.528
	28.409	2.543	28.436	2.653	28.453	2.560	28.45B	2+556	28.454	2.575
	28.444	2.589	28.432	2.663	28.424	2.635	28.422	2.667	28.429	2.704
	28.452	2.747	28.490	2.794	28.537	2.842	28.583	2.890	28.621	2.938
	28.649	2.983	28.667	3.026	28.673	3.066	28.663	3.105	28.641	3.141
	28.610	-3.109	28.570	-3.080	28.514	-3.052	28.445	-3.026	28.171	-3.000
	28,293	-2.975	28.213	-2.943	28.135	-2-919	28.066	-2-989	28.012	-2.855
	27.974	-2.819	27.958	-2.781	27.965	-2.738	27.995	-2.694	28.045	-2.652
	28.111	-2.611	28.180	-2.574	28.236	-2.540	28.270	-2.509	28.274	-2.4A0
	28.249	-2.452	28.206	-2.421	28.156	-2.394	28-107	-2.363	28.065	-2,330
	28.039	-2.294	28.033	-2.251	28.047	-2.220	28.078	-2.195	28.120	-2.152
	28.166	-2,122	28.210	-2.096	28.247	-2.072	28.270	-2.049	28.275	-2.026
	28.264	-2.000	28.244	-1.973	28.217	-1.935	20.188	-1.898	28.167	-1.458
	28.154	-1.818	28.180	-1.791	28.208	-1.744	28.242	-1.711	28.280	-1.681
	28,317	-1.652	28,345	-1.623	28.361	-1.598	28.368	-1.570	28.374	-1.541
	29.392	-1.505	28.422	-1.474	28.461	-1.438	28.507	-1.402	24.552	-1.369
	28.587	-1.335	28.608	-1-304	28.609	-1.275	28.593	-1.249	24.556	-1.222

-1.074	-0.914	-0.170	-0.640	-0.508	-0.315	-0-124	0.035	0.125
28.253	28.067	28.094	28.395	28.295	28.398	2 A. 764	28.484	26.093
-1-108	-0.943	-0.800				-0.158	900 0	0.120
28.312	28.087				28.318	28.733	28.581	28.134
-1-140	-0.974	-0.828	-0.689	-0.566	000-0-	+61-0-		901.0
28.377	28.116	28.043	28.290	28.387	28.267	28.674	28.666	28.200
-1.163	-1.006	-0.857	-0.714	-0.5.32	-0.433	-0.232	-0.053	0.087
28.443	28.154	28.046	28.221		28.246		28.730	28.287
-1.196	-1.040	-0.865	-0.741			-0.273	-0.091	0.063
28.504	28.201						26.764	28.384

.

THE ASPECT ANGLE =	VGLE H	0.0 DEGREES	ES					
TING FRE	0 = 2.0	STARTING FREQ = 2.00000 STEP=	0.010CC GHZ	SHZ				
DRIGINAL	VERT ICAL	THE URIGINAL VERTICAL PCLARIZATION DATA:	N DATA:					
15.676	1.310	15.746	1.313	15.949	1.324	16.258	1.339	16.630
17.026	1.368	17.421	1.373	17.823	1.369	18.209	1.400	18.541
119.61	130	19.023	1.452	19.207	1.478	19.381	1.508	19.563
19.762	1.575	19.976	1.613	20.214	1.643	20.468	1.674	20.721
20.956	1.725	21.152	1.745	21.269	1.760	21.267	1.770	21.138
20.885	1.783	20.505	1.793	19.996	1.600	19.371	1.816	18.655
17.866	1.871	17.091	116.1	16.125	1.962	15.184	2.025	14.207
13.210	2.198	12.204	2 • 30 3	11.253	2.437	10.445	2.586	9.844
9.503	2.928	9.522	3.116	9.930	-2.983	10.674	-2.824	11.589
12,553	-2.615	13.522	-2.554	14.457	-2.513	15.324	-2.488	16.106

1.354 1.613 1.541 1.702 1.778 1.839 2.104 2.702 -2.702

16.787	-2.451	17.334	-2.421	17.744	-2.386	18.034	-2, 337	18.221	-2.279
16.317	-2.212	18.364	-2-135	18.401	-2.055	18.454	-1.974	18.533	ACP - 1 -
18.638	-1.832	18.744	-1.77.	18.842	-1.739	16.913	-1.713	18.924	-1.702
18.855	-1.702	18.694	-1.711	18.440	-1.723	18.134	-1.733	17.915	-1.733
17.502	-1.719	17.191	-1.631	16.899	-1.644	16.623	-1.587	16.374	-1.519
16.182	-1.440	16.050	-1.353	15.950	-1.267	15.875	-1.177	13.913	-1.082
15.748	-0.984	15.685	-0.873	15.649	-0.764	15.652	-0.641	15.731	-0.510
15.903	-0.378	16.176	-0.253	16.537	-0-141	16.944	-0.045	17.336	0.039
17.674	0.117	17.960	0.193	18.199	0.274	18.396	0.355	18.563	0.437
18.720	0.520	18.875	0.633	19.048	0.680	19.247	0.745	19.457	0.796
13.671	0.831	19.881	0.853	20.070	0.852	20,233	0.842	20.364	0.822
20.490	0.799	20.622	0.773	20.750	0.762	20.857	0.757	20.932	0.767
20.968	0.792	20.958	0.821	20.889	0.874	20.752	0.925	20.545	0.979
20.269	1.034	19.942	1.087	19.574	1.135	19.156	1.177	19.582	1.217
18.160	1.254	17.597	1.293	17.006	1.337	16.407	1.390	15.825	1.157
15.306	1.541	14.836	1.642	14.409	1.763	14.044	1.903	13.777	2.063
13.644	2.237	13.688	2.414	13,843	2.589	14.155	2.753	14.570	2.903
15.036	3.036	15.520	-3-133	16.017	-3.035	16.523	-2.956	17.025	-2.894
17.508	-2.848	17.961	-2.814	18.375	-2.790	18.745	-2.774	10.048	-2.764
19.274	-2.759	19.426	-2.756	19.523	-2,753	19.563	-2.751	19.542	-2.745
19.465	-2.737	19,338	-2.74.2	19.177	-2.701	18.988	-2.674	18.778	-2.643
18.550	-2.609	18,309	-2.572	18.045	-2,532	17.745	-2.487	17,396	-2.437
16.971	-2,380	16.470	-2.314	15.917	-2,236	15.348	-2.143	14.799	-2.043
14.330	116.1-	14.016	-1.779	13.892	-1.644	13.923	-1.512	14.064	-1.390
14.262	-1.275	14.475	-1-131	14.679	-1.093	14.851	-1.012	14.986	-0.935
15.106	-0.858	15.226	-0.731	15.357	-0.701	15.500	-0.620	15.562	-0.541
15.862	-0.467	16.088	-0.402	16.327	-0.343	16.583	-0.289	16.853	412.0-
17.127	-0.187	17.399	-0.142	17.656	860.0-	17.897	-0.056	18.120	-0-015
18.313	0.021	18.458	0.052	18.545	0.078	18.586	101.0	14.595	121.0
18,575	0.138	18.526	0.151	18.460	091.0	18.392	0.168	19.145	0.172

THE ORIGINAL HORIZONTAL POLARIZATION DATA:

-1.813 -1.789 -1.639 -1.328 0.685 3.058 -2.457 -2.085 -1.869 -1.492 -0.597 0.115 0.411 0.502 0.958 1.137 1.423 1.973 2.630 3.070 -3.010 1.755 2,393 1.621 1.596 1.984 -2.548 6.159 20.195 20.936 960.0 17.995 11.319 3.988 17.237 9.874 19.615 8.269 0.789 0.205 8.858 3.956 15.347 18.037 9.204 14.764 13.211 15.989 3.841 13,541 7.463 5.779 6.123 -1-900 -1.802 -1.678 -1.512 -0.982 0.483 0.638 1.730 1.924 2.293 2.896 -2.582 -2.142 -1.821 -1.379 0.002 0.378 0.908 1.104 1.304 1.843 2.515 3.003 3.043 1.601 1.678 -2.882 0.679 014.0 14.240 17.628 19.026 19.963 20.896 966.03 18.484 15.527 11.589 13.087 6.668 18.938 9.825 19.746 8.638 16.489 14.718 9.107 6.626 4.734 13.337 15.296 17.223 -1.825 -1.809 -1.714 -1.418 -1.040 -0.138 .2.728 -1.537 0.595 2.750 -2.204 -1.938 0.334 194.0 1.074 2,935 3.079 1.578 2.203 0.854 1.277 1.724 2.382 -2.915 1.665 1.712 1.872 19.383 17.147 18.842 19.744 20.785 20.632 18.941 16.220 12.303 19.739 8.954 14.576 14.780 14.251 12.331 16.053 16.831 16.977 18.608 13.283 6.928 -1.933 -1.813 -1.745 -1.566 -1.447 -1.163 -0.30 3 0.277 0.558 1.826 2.623 2.891 -2.273 -1.982 0.451 151.0 1.0.1 1.223 1.614 2.243 2.847 3.11) .2.945 2.773 1.551 1.651 1.763 2.122 20.619 9.675 17.411 3.965 16.600 18.629 615.61 20.800 16.853 13.116 11.670 15.385 18.213 019.61 19.878 19.220 15.020 20.192 18.452 15.174 17.436 13.364 19:361 14.291 -1.848 -1.770 -1.259 5.049 -2.355 -1.601 -0.499 0.205 0.740 1.692 3.05e -1.471 0.434 1.513 2.107 2.749 2.502 -2.032 -1.816 0.527 1.003 3.120 1.636 1.787 1.174 2.978 17.440 17.699 15.993 20.417 20.903 13.948 11.317 19.435 19.890 9.438 17.863 15.495 19.435 20.792 19.954 18.642 18.367 17.751 6.206 13.875 19.376 B\$4.61 109.01 15.671 3.869 13.561

,

17.436	-2.328	17.839	-2.261		-2.193	17.695	-2.126	17.599	-2.059
17.500	-1.995	17.387	-1.934	17.246	-1.876	17.067	-1.819	16.837	-1.762
16.558	-1-704	16.243	-1.641	15.897	-1.573	15.525	-1.497	15.158	-11-
14.825	-1.319	14.559	-1.213	14.375	-1.099	14.280	-0.978	14.295	-0.452
14.438	-0.724	14.707	635*0-		-0.483	15.529	-0.380	15.024	-0.295
15.522	-0.223	16.998	-0.151	17.433	+11.0-	17.805	-0-074	18.100	-0.040
18.325	-0.013	18.491	0.012		0.035			18.715	0.085
18.732	0.111	18-737	0.143	18.745	0.172	18.777	0.205	18.835	0.237
19.914	0.267	19.006	0.233	19.100	0.315	19.182	0.329	19.237	0.337

		5 23.429	3 23.410	9 23.220	6 21.943	0 20.722	\$ 20.594	7 19.972	6 20.378	8 21.617	
		-2.105	-2.013	-1.769	-1.556	-1.200	-0.955	-0.777	-0.356	-0.018	:
		23,479	23.415	23,293	22,327	20.789	20.656	20.148	20.046	21.471	
		-2.107	-2.049	-1.820	-1.602	-1.278	-0.985	-0.825	-0.456	-0.061	600
GHZ		23.533	23.411	23,339	22.662	20.948	20.693	20.301	19.846	21.286	100.10
0.010¢ 0 GHZ	DN DATA:	-2-111	-2.075	-1.872	-1.642	-1.36)	-1.023	-0.864	-0.554	-0-114	0.063
2.00000 STEP=	ORIGINAL VEHTICAL FCLARIZATION DATA:	23.563	23.412	23.370	22.917	21.211	20.705	20.426	19.767	21.039	21.822
0 = 2.00	VEHT ICAL	-2.107	-2.091	-1.922	-1.680	-1.436	-1.071	-0.897	-0.643	-0.181	0.045
STARTING FRED =	THE ORIGINAL	23.572	23.410	23.193	23.099	21.558	20.705	20.522	19.824	20.723	21.734

-1.970 -1.970 -1.501 -1.130 -0.927 -0.717 -0.263

21.999	541.0	22.057	0.231	22.136	0.221	22.253	0.227	22.428	0.722
22.665	0.208	22,927	0.185	23.171	0.162	23.356	0.1.0	23.503	621.0
23.578	0.113	23.586	0.11.0	23,511	0.128	23,353	0.160	23.140	0.211
22.897	0.281	25.652	0.364	22.436	0.467	22.260	0.571	22.112	0.573
21.963	0.764	21.787	0.8.6	21.578	0.920	21+357	0.66	21.128	1.057
20.887	1.123	20.637	1.133	20.391	1.263	20.159	1.335	19.985	1.405
19.849	1.466	19.748	1.542	19.687	1.574	19,652	1.616	19.639	1.649
19.595	1.680	19.516	1.712	19.414	1.752	19.299	1.798	19.190	1.852
19.110	1.912	19.089	1.933	19.160	2.053	19.320	2.128	19.537	2.203
19.788	2.277	20.059	2 . 34 3	20.337	2.418	20.610	2.483	20.854	2.543
21.056	2.589	21.220	2.653	21.356	2.700	21.464	2.747	21.542	2.192
21.614	2.838	21.712	2.887	21.846	2.936	21.996	2.983	22.151	3.027
22.301	3.065	22.434	3.093	22.538	3.130	22.605	-3, 127	22.619	-3.104
22.571	-3.082	22.460	-3.061	22.286	-3.038	22.047	-3.015	21.749	-2.989
21.414	-2.960	21.079	-2.927	20,773	-2.889	20.513	-2.846	20.319	-2.799
20.213	-2.749	20-171	-2.69 \$	20.154	-2.646	20-136	-2,593	20.123	-2.540
20.117	-2.484	20.117	-2.423	20.120	-2.367	20.122	-2,308	50.119	-2.248
20.101	-2.189	20.063	-2-130	20.012	-2.072	19.954	-2.014	19.900	-1.957
19.848	-1.902	19.788	-1.851	19.717	-1.798	19.644	-1.744	19.555	-1.591
19.445	-1.636	19.306	-1.573	19.158	-1.513	19.020	-1.444	19.901	-1.172
18.808	-1.298	18.750	-1.224	18.733	-1.150	18.765	-1.079	19.944	-1.010
19.962	-0.945	19.111	-0.883	19.280	-0.830	19.461	-0.782	19.635	-0.737
19.810	+69*0-	19.980	-0.653	20.150	-0.611	20.308	-0.571	20.444	-0.531
20.568	-0.493	20.685	-0.457	20.787	-0.424	20.860	-0.393	20.901	-0.363
20.925	-0.333	20.933	-0.303	20.922	-0.272	20.884	-0.241	118.02	-0.209
20-711	-0.173	20.585	-0-135	20.442	960.0-	20.283	-0.055	20.113	-0.011
19.043	0.036	19.779	0.095	19.627	0.138	19.487	0.192	19.365	0.248
19.276	0.305	19.225	0.363	19.199	614.0	19.176	0.473	19.148	0.529
19.132	0.585	19.136	0.543	19.164	0.700	19.208	0.757	10.268	0.811
19.341	0 - 86 7	19.421	0.911	19.492	0.950	19.548	0.977	19.586	0.993
THE DRIGINAL	HORIZONTAL	POLARIZA	POLARIZATION DATA:						

THE DRIGINAL HURIZONTAL POLARIZATION DAFA:

22.671 -2.525	21.904 -2.333	20.943 -2.035	20.786 -1.660	21.110 -1.285	22.015 -0.963	23.004 -0.758	21,535 -0,597	23.447 -0.413											7 7 0 0 0 1 1 0 1 1	, , , , , , , , , , , , , , , , , , ,									
-2.554	-2,379	-2.103	-1.737	-1.360	-1.019	-0.793	-0.631	-0.451	-0.270	-0.129	0.042	0.404	0.839	1.108	1,373	1.698	1.849	1.934	2,085	2.314	2.589	2.863	-3.063	-2, 785	-2.557	-2.275		-2.019	-2.019
22.749	22.107	21.069	20.736	21.016	21.796	25,852	23.459	23,530	22.969	22.293	21.540	20.370	20.169	20.853	21.164	21.844	22.878	23.454	23.481	22.689	21.863	21.301	21.100	21.443	21.698	22.650		23.170	23.170
-2.576	-2.422	-2.168	-1.816	-1.434	-1.080	-0.830	-0.662	-0.488	-0.302	-0.156	-0.00	0.318	0.760	1.067	1.307	1.645	1.831	116.1	2.049	2.260	2,535	2.803	-3.135	-2.829	-2.608	-2.326	900	69199	-1.940
22.799	22.271	21.244	20.724	20.951	21.595	22.863	23, 382	23.581	23.098	22.419	21.723	20.574	20.097	20.737	21.127	21.641	22.693	23,370	23,550	22.873	22.017	21.408	21.078	21.389	21.615	22.414	24.1.79		22.723
165.5-	-2.461	-2.221	-1.893	-1.509	-1-144	-0.851	-0.693	-0.525	-0.334	-0.183	-0.04 3	0.236	0.615	1.024	1.243	1.583	1.303	1.632	2.015	2.210	2.481	2.749	3.073	-2.873	-2.65.3	-2.383	-2.153	,	-1.383
22.837	22.416	21.452	20.765	20.896	21.404	22.483	23.276	23.597	23,222	22.552	21.887	20.817	20.073	20.597	21.084	21.458	25.492	23.277	23.580	23.061	22.178	21.505	21.083	21.312	21.557	22.184	23.129		75.866
965-7-	-2.495	- 2.283	-1.966	-1.584	-1.212	- 0.913	-0.724	195.0-	-0.375	-0-211	-0.075	0.162	0.585	0.973	1.196	1.516	1.779	1.877	1.986	2.164	2.425	2.69€	5.995	-2.933	-2.700	-2.442	-2.193		-2.022
759.77	22.550	21.670	20.444	20.841	21.243	22.253	23.147	23.579	23.340	22.693	22.035	21.081	20.110	20.439	21.022	21.314	22.279	23.172	23.566	23.233	22.344	21.606	21.123	21.226	21.519	21.980	23.023		86.72

-1.013 21.769	21.389 -0.777 21.251	20.863 -0.676 20.767 -0.641	20.349 -0.369 20.309	20-771 0.050 20.973	21.611 0.227 21.592	21.691 0.342 21.655	21.513 0.582 21.497	21.490 0.750 21.489
-1.086	-0.806	-0.701	-0.455	-0.018	0.208	0.307	0.531	0.729
21.881	21.517	20.947	20.434	20.579	21.499	21.721	21.540	21.485
		-0.723	-0.533	1 60 .0-	0.183	0.283	0.47)	0.702
21.944	21.622	21.031	20.542	20.425	21,354	21.742	21.574	21.481
-1.245	-0.891	-0.737	-0.593	-0.184	0.150	0.259	0.429	0.668
22.019	21.705	21 - 130	20.658	20.332	21 - 179	21 - 736	21.614	21.486

. . . .

•	STARTING FREG =		4.00000 STEP=	0.0200 GHZ	2H5				
	THE ORIGINAL	AL VERTICA	VERTICAL FCLARIZATION DATA:	DATA:					
•	28.369	0.262	28.348	0.263	28.290	0.285	28.213	0.314	28.130
	26.057	004.0	28.004	0.453	27.980	915.0	27.985	0.581	28.021
•	28,063	0.718	28-104	0.787	28-141	0.856	28.183	0.929	28.229
	28.277	1.075	28.320	1.143	28.349	1.223	28.369	1.298	28,384
•	78.396	1.440	28.413	1.505	28.436	1.566	28.461	1.623	28.482
	28.502	1.728	28.519	1.773	28,525	1.828	28.512	1.877	29.476
•	29.419	1.978	28.344	2.032	28.260	2.088	28.171	2.147	24.0AS
	28.010	2.27	27.956	2.343	27.926	2.414	27.917	2.483	27,930
•	27.964	2.613	28.013	2.673	28.071	2.728	28.127	2.790	28.174
	28.211	2.876	28.243	2.933	28.235	2.971	28.213	3.022	28.172

THE ASPECT ANGLE = 90.000 DEGREES

0.352 0.649 1.002 1.370 1.676 1.625 2.203 2.550 2.550

	28.131	3.139	28.108	-3.073	28.107	-3.010	28.142	-2,938	28.215	-2.865
	28.319	-2.793	28.444	-2.72?	28.576	-2.653	28.697	-2.587	28.790	-2.522
	28.850	-2.459	28.884	-2. 195	28.887	-2.334	28.861	-2.270	28.815	-2.207
•	28.761	-2.143	28.701	-2.073	28.636	-2.013	28.564	-1.948	29.484	-1.883
	28.393	-1.818	28.235	-1.753	28.202	-1.687	28.124	-1.621	28.063	-1.555
	28.023	-1.489	28.005	-1.42 2	28.013	-1,355	28.039	-1.288	28.084	-1.221
	28.144	-1.155	28.213	-1.031	28.287	-1.027	28.361	-0.965	28.428	-0.906
	28.489	-0.849	28.542	-0.793	28.583	-0.738	28.607	-0.683	28.614	-0.628
	28.608	-0.575	28.533	-0.525	28.540	-0-477	28.483	-0.430	28.417	-0.383
	28.351	-0.333	28.288	-0.233	28.234	-0.222	28.192	-0.160	28.162	-0.093
	28.151	-0.027	28.158	10.0	28.177	0.111	28.206	0.180	28.245	0.245
	28.286	0.311	28.321	0.373	28.347	0.433	28.361	0.489	28.157	0.543
	28.349	965.0	28.344	0.643	28,343	0.704	28.343	0.761	28,350	0.821
,	28.364	0.833	28.376	6+6-0	28.377	1.017	28.369	1.067	28,363	1.158
	29.360	1.229	28.362	1.297	28,371	1.365	28.381	1.429	28,388	1.492
	26.391	1.553	28.390	1.614	28.379	1.674	28.356	1.734	28.353	1.795
	28.328	1.856	28.311	1.923	28.309	1.985	28.318	2.052	28,336	2.118
	28.355	2.183	28.372	2.247	28.382	2.310	28.379	2.371	28.374	2.430
	28.372	2.486	28.370	2.541	28.376	2.594	28.380	2.647	28, 382	2.698
	28.371	2.751	28.328	2.801	28.254	2.860	28.150	2.919	28.018	2.982
	27.867	3.051	27.713	3.121	27.584	-3.084	27.489	-3.006	27.439	-2.932
	27.429	-2.065	27.469	-2.801	27.555	-2.748	27.685	-2.698	27,837	-2.654
	27.999	-2.611	28.169	-2.563	28.335	-2.524	28.487	-2.477	28.610	-2.427
	28.693	-2,373	28.729	-2.315	28.704	-2,257	28.619	-2.193	28.478	-2.126
	28.291	-2.055	28.085	-1.973	27.886	-1.900	27.719	-1.820	27.599	-1.742
	27.564	-1.688	27.608	-1.603	27.700	-1.539	27.812	-1.483	27.922	-1.428
	28.017	-1.371	28.082	-1.313	28.105	-1.252	28.087	-1.186	28.039	-1.116
	21.912	-1.042	27.897	-0. 16 1	27.831	-0.889	27.778	-0.813	27.745	-0.739
	27.723	-0.063	27.711	165.0-	27.700	-0.519	27.710	-0.447	27.760	-0.378
	27.848	-0.314	27.966	-0.233	28.101	-0.212	28.227	-0-179	28.320	-0.159
-	THE URIGINAL	HURIZENTAL		POLARIZATION DATA:						

THE URIGINAL HURIZENTAL POLARIZATION DATA:

-2,554 0.299 2.005 2.312 -0.791 0.548 1.190 1.509 2.917 2.696 3.123 -1.075 1.843 2.207 2.585 -2.765 1.125 1.597 -2.815 -2.278 -1.961 -1.645 -1.361 -0.487 -0.119 0.921 .3.064 1.106 -2.454 27.963 27.679 27.910 27.758 2A.088 29.512 28.479 28.014 17.799 27.919 28.160 27.547 28.196 28.404 28.633 18.347 28.534 28.176 27.572 27.454 27.925 28.397 27.831 29.424 28.372 2.245 -2.025 -1.129 -0.848 -0.552 -0.199 1.133 2.614 3.040 -2.876 -2.606 -1.397 0.218 0.584 0.870 1.442 1.776 2.512 2,855 -3, 123 -2.826 1.067 1.350 1.634 1.947 -2.337 -1 - 709 2.131 -2.519 27.841 28.153 28.489 28.341 28.616 28.446 27.861 27.634 27.979 28.473 28.522 28.451 28.088 27.816 27.904 27.858 27.867 28.107 28.286 27.633 27.457 27.502 27.849 28.180 28,391 28,351 -2.940 -0.276 -0.905 -0.615 0.135 0.518 0.819 1.077 2.534 -2.657 -2.394 -2.090 181.1. 1.377 1.710 2.065 3.100 1.575 1.887 2.182 2.954 -1.772 -1.457 2.437 2.792 2.578 28.170 27.473 28.523 27.979 28.503 27.817 28.049 27.832 27.752 28.024 28.030 28.575 27.613 27.879 28.505 28.490 27.851 27.897 27.884 28, 368 194.72 28.272 28.404 28.067 28, 342 27.777 28.109 28.317 28.167 1 96 .0 -0.05) 6 4 4 * 0 1.239 1.513 2.122 2.457 2.867 -3.003 -2.449 -1.234 -0.673 0.765 1.312 1.640 1.982 2.363 2.727 3.033 1.824 -2.709 -1.835 -1.511 -0.343 1.024 -2.154 610.82 8.045 8.206 27.628 28.486 27.898 27.458 28.365 28.521 18.581 28.106 27.795 28.312 28.521 28.517 28.252 27.802 27.890 27.904 27.815 28.027 28.281 28.417 27.909 27.476 27.677 28.055 28.240 28,182 2.063 2.303 0.376 1.250 194.1 2.781 -2.502 -2.216 -1.898 -1.582 -1.286 -1.020 -0.734 -0.419 0.708 1.577 2.658 -3.081 -0.035 0.972 1.911 -2.761 2.284 2.978 28.204 27.797 080.8 28.010 8.04A 28.620 28.232 27.728 28.527 28.478 27.951 27.878 27.911 27.815 27.977 28.220 28.048 27.510 28.375 28.148 11.677 26.532 28.331 28.434 27.452 28.464 27.604 27.992 28.239 28 - 22 1) 7

•

	28.
191 -2.062	28.191 -2.06?
	28.113 -1.592
	28.410 -1.374
332 -1-103	26.332 -1.103
	28.023 -0.753
	28.090 -0.450
	27.912 -0.152
925 0.203	27,925 0,20

	1.674	2.425	-3.067	-2.088	0.012	1.208
	14.870	14.706	16.123	13.041	13.521	12.990
	1.559	2.360	3,088	-2.174	-0.639	1.019
	4 6 7	14.410	15.687	16.351	12,960	13.157
	•	2.182	2,958	-2.255	-0.828	0.226
ZHS		15.773	15.626	16.703	14.174	15,505
ES 0.02CC GHZ	N DATA:	1.436	2.82?	-2.862	-1.833	0.153
. 0.0 DEGREES	POLARIZATIO	16.062	15.338	16.600	14.785	14.682 15.715 13.908
11.7	VERTICAL	1.421	1.824	-2,943	-1,242	0.080
THE ASPECT ANGLE	STARTING FREG # 4.00000 012.	15.159	14.462	15.364	15.366	14,114
-						•

DEGREES

	112-21	1.407	12.666	1.603	12.769	1.812	13,051	2.008	13.489	2.187
	14.040	2,343	14.637	2.477	15, 229	2.586	15.744	2.674	15.152	2.744
	16.429	2.800	16.549	2.841	16.508	2.892	16.294	2.941	15.905	2.999
	15.354	3.071	14.650	-3.12)	13.825	-2.998	12.933	-2.846	12.053	-2.661
	11.293	-2.445	10.746	-2.202	10.529	-1.945	10.677	-1.688	11.143	-1.447
•	11.869	-1.235	12,763	-1.058	13.732	916.0-	14.689	-0.806	15.552	-0.725
	16.283	-0.664	16.864	619.0-	17.275	-0.577	17.513	-0.538	17.560	161.0-
	11.417	-0.442	17.088	-0.373	16.581	-0.297	15.915	-0.192	15.145	-0.055
	14.363	611.0	13,737	0.3 15	13.428	0.579	13.535	0.828	13.974	1,059
•	14.580	1.261	15.234	1.4 13	15.853	1.586	16.398	1.719	16.855	1.839
	17.216	1.350	17.490	2.057	17.699	2,163	17.843	2.269	17.929	2,376
•	17.954	2.483	17.921	2.543	17.821	2.693	17.639	2,793	17.361	2.892
	17.009	2.993	16.583	3.103	16.108	-3.065	15.606	-2.936	15.125	-2.793
	14.716	-2.635	14.433	-2.06 6	14.302	-2.294	14.339	-2.127	14.493	-1.969
	14.719	-1.822	14.946	-1.681	15.155	-1.549	15.364	-1.412	15.603	-1.272
•	15.850	-1.134	16.112	-1-003	16.373	-0.875	16.598	-0.761	16.761	-0.526
	16.830	-0.558	16.793	-0.463	16.633	-0.367	16.344	-0.264	15.943	-0.147
7	15.443	110.0-	14.925	6+1.0	14.525	0.333	14,369	0.525	10.474	0.707
	10.763	0.867	15.108	1 66.0	15,393	1.103	15.567	1.190	15.600	1.266
•	15.493	1.339	15.238	1.413	14.843	1.492	14.347	1.582	13.806	1.582
	13,325	1.800	12.978	1.933	12.792	2.072	12,713	2.210	12.648	2.340
•	12.507	2.465	12.200	2.583	11.679	2.723	106.01	2.881	9.934	3.079
	9.946	-2.948	8.293	-2.621	8.451	-2.262	4.417	-1.942	10.779	-1.534
	12.166	-1.509	13.405	-1.372	14.414	-1.264	15.175	-1.174	15.731	-1.594
	15.085	-1.020	16.271	145-0-	16.291	-0.875	16.172	-0.796	15.930	-0.705
	15.596	965*0-	15.219	-0.457	14.894	-0.320	14.669	-0.159	14.494	0.005
	14.356	0.167	14.267	0.313	14.197	0.462	14.165	0.592	14.213	0.709
•	14.337	0.806	14.482	0.493	14.654	0.962	14.832	1.018	15.006	1.067
	15.144	1.116	15.237	1.154	15.239	1.211	15.166	1.260	15.017	1.314
•	14.825	1.372	14.559	1.433	14.262	1.508	14.021	1.572	13.865	1.620
Ξ	THE URIGINAL	HORIZGNTAL		POLARIZATIUN DATA:						

THE URIGINAL HORIZGNTAL POLARIZATIUN DATA:

1.637 2.113 1.089 0.510 0.077 1.191 1.954 2.363 2.806 .2.667 -1.025 0.558 0.093 196.0 1.569 1.952 2.550 2.686 .2.037 -1.562 -1.230 0.352 0.577 1.342 2.664 3.125 2.743 0.182 166.0 1.134 1.931 15.363 13.200 14.490 17.095 15.946 15.713 14.420 4.938 17.137 16.829 5.902 18.707 14.002 6.059 12.334 11.963 14.767 7.511 5.532 17.161 17.651 14.353 15.359 16.378 16:201 11.247 13.458 0.810 1.482 .0.599 .0.162 1.002 2.289 2.698 .2.870 1.814 -1-122 099-0-0.060 1.872 2.392 .2.894 2.125 -1.736 1.330 0.453 1.652 3.045 0.272 1.841 1.22.1 2.531 2.829 .2.287 1.271 .0.591 7.470 16.789 17.740 614.9 13.593 3.900 6.768 7.020 14.826 3.728 15.719 13.000 17.136 15.576 15.673 6.386 5.198 14.933 5.611 17.170 6.924 14.925 4.427 6.896 11.471 14.811 1.574 14.306 3.522 -2.010 -1.228 0.625 2.377 2.915 -2.432 -1.477 .0.695 0.250 0.174 0.820 1.707 2.215 2.603 -0.755 1.383 1.796 2.257 -2.229 -1.809 -0.201 -3-119 1.123 -3.052 0.313 -1.421 .0.801 3.619 4.060 7.365 6.370 4.125 3.411 6.305 7.229 15.329 5.969 3.675 5.028 5.193 7.026 7.076 5.479 6.548 7.428 15.781 7.748 6.864 15.294 1.261 4.134 6.291 5.684 2.229 2.685 -0.330 1.263 1.723 2.142 2.999 -2.561 -1.655 -0.805 0.655 1.55.2 2.517 3.073 2.223 -1.347 .0.646 0.493 -1.883 2.203 2.875 -0.377 0.0e7 2.137 2.941 0.123 0.986 -2.354 -1.503 1.343 -0.974 3.388 6.798 7.184 666.5 3.929 6.072 7.384 901.9 7.649 7.224 14.724 3.093 5.744 5.834 3.699 14.842 960*9 4.323 160.91 6.037 3.033 189.51 4.887 17.491 7.321 11.371 11.874 5.971 -0.935 2.929 -1.482 -1.958 2.018 3.079 0.509 1.378 2.439 .2.448 544.0-0.259 1.135 1.648 2.024 -2.506 -1.586 -2.655 -1.913 -0.423 2.051 -0.934 2.734 0.105 2.777 0.00 -1.115 0.974 0.976 1.227 60000 14.743 17.241 17.533 15.123 15.088 14.896 13.853 1.359 10.474 15.507 3.016 15.329 7.280 6.302 3.958 11.753 12.637 15.783 6.263 5.940 604.9 4.063 5.673 17.481 4.387 7.294 9.414 17.451 , 3 ł

•

			!	•	,	1 40 01	21.13	11111	2.357
25.5	1.492	11.741	1.677	101:1	1.690		1		
	9. 6.	12.167	2.715	12.666	2.852	13.111	2.972	13.502	3.081
00011	1,70		* U.J. F.	14.540	-2.916	14.827	-2.826	15.069	-2.735
700.7	501 · 5			15.384	-2.462	15.288	-2.367	18.097	-2.265
15.261	-2.043	125.61		101741	-1-882	13.741	-1.712	13.529	-1.520
14.820	-2.104	200	081.1	14.085	866-0-	14.481	118.0-	14.830	-0.695
13.522	-1.322	13.731		0 4 C 4 T	-0.375	15.157	-0.276	14.939	-0.171
15.081	4/G *0-	190 41	5,000	13.947	0.220	13.699	0.376	13.587	0.533
14.628	0.682	13.824	0.80	14.086	0.908	14,327	0.978	14.486	1.019

• •

-

		2.108	2.189	2.875	-3+038	-2.485	-1.926	-1.262		•	0.000	
		18.383	19.525	20.400	19.950	18.575	14.545	20.246	21.064	20.529	27.625	
		1.928	2, 398	2,805	-3, 121	-2.612	-1.961	-1.353	-0.982	-0.586	-0.099	
		18.269	19.214	20,328	20.010	18.769	18,395	19,871	21.070	20.719	20.537	
		1.865	2,300	2.733	3.086	-2.733	-2.095	-1.455	-1.045	189.0-	-0.187	
	GH2	8 2 - 6	200	20-196	20.237	4100	19:014	10.486	20.997	20.823	20.530	
n L	0.02CL 0 GHZ	N DATA:	1.840	502.2		20.5	-2.843	-2.2.2.	10001-	11111-		
5.000 DEGRE	4.00000 STEP=	ERTICAL POLARIZATION DATA:	18.136	18.727	2 10.02	20.351	19.298	18.351	19.118	20,835	20.924	*66.02
GLE = A	00.4	VERT 1CAL	1.812	2.101	2.576	2+6+3	-2.947	-2.358	-1.694	-1.182	10.845	-0.386
THE ASPECT ANGLE = 45.000 DEGREES	STARTING FRED =	THE DRIGINAL VE	19-122	18.534	19.790	20.406	19.584	18.433	18,795	20,580	21.008	20,560
	•		•		•		•		•		•	

L										
	20.460	901.0	20.327	0.207	20.135	0.313	19.918	0.426	19.702	0.546
•	19.517	0.670	19.392	951.0	19.330	0.921	19.322	1.000	19.357	1.164
	19.419	1.280	19.511	1.395	19.627	1.509	19.741	1.617	19.847	1.719
•	19.925	1.811	196.61	1.895	19.944	1.974	19.870	2.052	19.729	2.132
	19.523	2.215	19.235	2.302	18.881	2.393	18.467	2.486	19.014	2.587
•	17.553	2.695	17.119	2.811	16.765	2.943	16.521	3.079	16.389	-3.061
	16.379	-2.920	16.478	-2.783	16.687	-2.653	16.982	-2.535	17.314	-2.431
•	17.637	-2.339	17.921	-2.256	18.158	-2.182	18.339	-2.114	18.450	-2.049
	18.495	-1.981	18.476	116.1-	18.392	-1.834	18.265	-1.749	19.105	-1.657
•	17.934	-1.557	17.767	-1.450	17.608	-1.338	17.472	-1.221	17.365	-1.101
	17.290	-0.981	17.258	-0.862	17,256	-0.748	17.268	-0-639	17.276	-0.536
•	17.272	-0.438	17.258	-0.345	17.236	-0.257	17.228	-0.170	17.239	-0.083
	17.273	900.0	17.312	f 50°0	17.358	0.199	17.409	0.304	17.484	0.413
	17.588	0.522	17.725	0.630	17.886	0.737	18.047	0.843	18.214	0.949
	18.372	1.052	18.512	1.153	18.624	1.252	18.706	1.351	18.745	1.4.8
	18.734	1.543	18.662	1.638	18.531	1.735	18.345	1.837	18.130	1.946
	17.920	2.062	17.743	2.193	17.618	2.300	17.561	2.420	17.582	2.535
	17.679	2.647	17.846	2.753	18.063	2.846	18.298	2.936	18.516	3.020
	16.713	3.101	18.857	-3.101	18.938	-3.018	18.964	-2.930	18.941	-2.837
•	118.877	-2.738	18.780	-2.635	18.649	-2.528	18.490	-2.419	18.320	-2.309
	18.154	-2.201	1 7.997	-2.09	17.841	-1.994	17.684	-1-898	17.518	-1.807
•	17.338	-1.720	17.138	-1.633	16.920	-1.557	16.701	-1.475	16.483	-1.368
	16.275	-1.294	16.106	-1.189	15.973	-1.079	15.880	-0.966	15.833	-0.845
•	15.864	-0.722	15.978	-0.593	16.196	-0.479	16.478	-0.369	16.791	-0.271
	17.066	-0.183	17.333	-0-103	17,561	-0.026	17.731	0.047	17.833	0.122
•	17.858	661.0	17.773	0.283	17.602	0.373	17.361	0.475	17.095	0.591
	16.821	0.717	16.587	0.852	16.380	066*0	16.233	1.129	16.134	1.262
•	10.013	1.385	15.837	1.453	15.563	1.603	15.196	1.709	14.743	1.824
	14.275	1.956	13.898	2.103	13.747	2.270	13.903	2.446	14.402	2.611
•	15,113	2.749	15.871	2.835	16.507	2.934	16.978	2.985	17.258	3.013
•										

THE OHIGINAL HOMIZENTAL POLARIZATION DATA:

2.273 1.158 1.688 2.558 -2.929 3.026 2.790 -2.212 -1 - 753 -0.398 0.624 -2.155 -2.01? -1.581 -1.043 -0.554 0.507 1.513 2.511 -1-382 .0.991 0.061 -0.031 0.995 2.033 2.503 2.961 21.043 11.599 20.816 181.187 21.335 20.930 21.034 10.883 20.555 20.572 20.427 20.542 21.594 21.866 21.355 21.970 21.988 199.03 21.034 124.0 106.12 21.954 21.343 22.152 162.15 21.244 21.682 22.354 2.748 1.857 2, 399 2.931 -2.889 -2.329 -1.829 -1.461 1.001 -0.486 -0.040 0.507 1.058 1.576 2.154 -3.021 -2.554 -2.094 -1.677 -1.153 -0.648 -0.142 0.402 0.901 1.403 1.937 2.408 2.874 0.028 0.483 866.0 20.542 869.03 20.516 20.903 21.528 11.577 21.475 21.194 21.875 11.986 21.975 21.337 11.659 21.542 0.670 \$16.03 21.852 1.884 216.15 21.605 20.622 196 0 21.007 20.441 22.052 22.398 21.933 -1.536 -1.107 -0.577 -0-135 0.389 1.467 2.037 2.634 -3.116 -2.178 -1.263 0.295 0.806 1.295 1.836 2.287 2.833 -2.984 -2.448 -1.912 0.957 -2.651 -1.767 -0.741 -0.251 2.783 1.754 2.314 3.076 20.483 0.565 20.485 20.773 21.429 21.582 21.800 21.600 21-160 22.003 21.965 11.948 21.418 21.285 854.03 11.475 21.713 10.750 978.0 21.054 20.174 1.084 \$69.03 21.761 21.537 21.955 22.390 20.581 22.043 -3.015 -2.586 -1.205 -0.675 .0.223 0.276 1.362 1.923 5.516 -2.743 -1.852 -1.372 .0.355 0.173 0.7C3 1.732 2.177 2.733 -2.003 -1.609 0.851 3.06 -2.267 -0.837 1-132 2.693 3.127 1.655 2.221 11.258 21.819 20.905 984.0 20.665 20.655 80.658 609-13 21.616 22.007 21.074 969.0 20.634 21.126 20.456 21.316 21.728 21.122 20.551 21.181 21.960 196-12 21.544 1+2-12 21.479 22.336 -1.479 1.563 3.116 2.683 2.103 1.680 1.298 0.312 0.166 0.740 1. 259 1.803 2.394 2.965 2.838 -2.360 -1.933 .0.937 0.457 6.000 609.0 1.623 3.045 2.069 0.781 1.092 2.128 2.597 2.824 2.622 21.819 20.529 19.931 21.844 21.109 20-711 21.072 21.106 20.771 20.600 20.617 20.436 20.576 21.187 21.653 21.251 21.994 21.222 21.428 21.030 20.826 21.620 21.452 21.962 11.993 191.15 22.253

					, 1 , 2 , 3 7	0.00	22.570	1.563	22.43H
10.0	22.125	1.738	22.680	1.695	32. K23	7. 7		•	006.77
•			265 6 2 3 2	1 • 300	22.686	1.219	22.801	72171	100
101.1	22.542	1.394	23 500						23.195
•			23.077	0.698	23.144	0.822	041.140		
1.055	22.007	440					010.77	0.307	22.665
		• 00	23.073	0.498	22,956	604.0	6	,	
0.565	23.154	409					22.40	-0.213	25.492
•	346 . 33	9.101	22.466	-0.005	22.434				,
0.005	23 643						22.131	-0.670	22.707
	5 6 5 7 7	-0.402	22.610	-0.497	22.672	4 8 4	1		
					610.77	-1.023	22,502	-1.126	47 F. C.C.
-0.755	22.761	-0.841	22.707	120.02				000	21.871
			461933	104.1-	22.060	-1.567	490.16	707	
-1.231	22.252	011		•			211.12	-2.294	21.757
-1.500	21.790	-1.928	21.731	-2.051	21.705	101	,		

r

_

- [1] A. J. Poelman, "A Study of Controllable Polarization Applied to Radar", Proceedings Military Microwaves 80 Conference, The Cunard International Hotel, London, England, pp. 389-404, Oct. 1980. [Microwave Exhibitions and Publishers Ltd., Temple House, 36 High Street, Sevenoaks, Kent, England, TN13 lJG.]
- [2] S. K. Chaudhuri and W.-M. Boerner, "Polarization utilization in profile inversion of a perfectly conducting prolate spheroid", IEEE Trans. Antennas Propagation, Vol. AP-25, pp. 505-511, July 1977.
- [3] W.-M. Boerner, M. B. El-Arini, C. Y. Chan, and P. M. Mastoris, "Polarization dependence in electromagnetic inverse problem", IEEE Trans. Antennas Propagation (Special issue on inverse methods in electromagnetics), Vol. AP-29, pp. 262-269, March, 1981.
- [4] J. R. Huynen, "Phenomenological theory of radar targets", Ph.D. dissertation, Technical University, Delft, The Netherlands, 1970.
- [5] W.-M. Boerner, M. B. El-Arini, S. Saatchi, M. Davidovitz, J. Nespor, W.-S. Ip, "Polarization utilization in radar target reconstruction", Final Report on Contract No. NAV-AIR-NO0019-80-C-0620, Sept. 15, 1981.
- W.-M. Boerner, "Polarization utilization in electromagnetic inverse scattering", Chapter 7 in <u>Inverse Scattering Problems in Optics</u>,
 H. P. Baltes, Ed., Topics in Current Physics, Vol. 20, New York: Springer, 1980, pp. 237-290.
- [7] D. B. Kanareykin, N. F. Pavlov, and U. A. Potekhin, <u>The Polarization of Radar Signals</u>, Sovyetskoye Radio, Moscow, 1966. English translation of Chapters 10-12: Radar Polarization Effects, New York: Macmillan
- [8] C. L. Bennett, A. M. Auckenthaler, R. S. Smith, J. D. DeLorenzo, "Space-time integral equation approach to the large body scattering problems", Sperry Research Center, Sudbury, MA, Final Report on Contract No. F30602-71-C-0162, AD 763/94, May, 1973.
- [9] C. L. Bennett, "Time domain inverse scattering", IEEE Trans. Antennas Propagations (Special issue on inverse methods in electromagnetics), Vol. AP-29, pp. 213-219, March, 1981.
- [10] E. M. Kennaugh and R. L. Cosgriff, "The use of impulse response in electromagnetic scattering problems", IRE National Convention Record, Part I, pp. 72-77, 1958.
- [11] Yu N. Barabenenkov, "Scattering of electromagnetic delta pulses by ideally conducting bodies of finite dimensions", Radiotekhonika i Electronika (USSR), Vol. 8, pp. 1069-1071, June 1963. C. C. Translation.
- [12] E. M. Kennaugh and D. L. Moffatt, "Transient and impulse response approximation", IEEE Proc. Vol. 53 (8), 1965, pp. 893-901.

- [13] D.J. Struik, "Lectures on classical differential geometry", Addison Wesley Publishing Company, Inc., Reading MA, U.S.A. and London, England, 1961.
- [14] M.M. Lipschitz, "Theory and problem of differential geometry", McGraw-Hill Book Co., Inc., N.Y., 1969.
- [15] S.K. Chaudhuri, "A time domain synthesis of electromagnetic backscattering by conducting ellipsoids", IEEE Trans. Antennas Propagations, Vol. AP-28, July 1980, pp. 523-530.
- [16] C.L. Bennett, "A technique for computing approximate electromagnetic impulse responses of conducting bodies", Ph D. dissertation, Purdue University, Lafayette, Indiana, August, 1981.
- [17] S.K. Chaudhuri and W-M. Boerner, "A monostatic inverse scattering model based on polarization utilization", Applied Physics, Vol. 11, No. 4, pp. 337-350, December 1976.
- [18] E.M. Kennaugh, "Polarization properties of radar reflections", M.S. thesis, Dept. Elec. Eng., O.S.U., Columbus, 1952 (O.S.U. Antenna Lab., Rep. 389-12, March 1, 1952)
- [19] J.D. Kraus, Radio Astronomy, New York: McGraw-Hill, 1966.
- [20] E.M. Kennaugh, "Effects of type of polarization on echo characteristics", 0.S.U., Antenna Lab., Columbus, OH, Reps. 389-1 to 389-24, 1949-1954.
- [21] W-M. Boerner, "State of the Art Review--- Polarization Utilization in Electromagnetic Inverse Scattering", Communications Lab. Rept, 78-3, Oct. 1978.
- [22] J.D. Young, "Target Imaging from Multiple-frequency radar returns", O.S.U., ESL Rept, 2768-6, June 1971.
- [23] C.L. Bennett, R. Hieronymus and H. Mieras, "Impulse Response Target Study", Sperry Research Center, Final Rept, 1977.
- [24] S.K. Chaudhuri, "Utilization of Polarization-Depolarization Characteristics in Electromagnetic Inverse Scattering", PhD thesis, Univ. of Manitoba, Winnipeg, Canada, 1977.
- [25] W-M. Boerner, C-M. Ho, B.Y. Foo, "Use of Radon's Projection Theory in Electromagnetic Inverse Scattering", IEEE Trans Antennas Prop., Special Issue on Inverse Methods in Electromagnetics, Vol. AP-29, pp. 262-269, March 1981.
- [26] C-M. Ho, "The Extension of Physical Optics Far-Field Inverse Scattering using Radon's Transform Theories and Polarization Utilization", M.S. thesis, Dept. of Information Eng., Univ. of Illinois at Chicago Circle, Chicago, 1980.

- [27] M.W. Long, "Radar Reflectivity of Land and Sea", D.C. Heath & Company, 1975.
- [28] E.A. Walton, "Analysis of Accuracy of a Radar Backscatter Measurement System using Phasor Compensation", OSU-ESL, Oct. 1981.
- [29] G.C. McCormick and A.Hendry, "Techniques for the Determination of the Polarization Properties of Precipitation", Radio Science, Vol. 14, No. 6, pp. 1027-1040, Nov-Dec 1979.
- [30] A.L. Maffett, "Scattering Matrices", in Methods of Radar Cross-Section Analysis, Ed. by J.W. Crispin, Jr. and K.M Siegal, Academic Press, NY, 1968.
- [31] M.W. Long, "Backscattering for Circular Polarization", Electronics Letters, Vol. 2, pp. 351, Sept. 1966.
- [32] L.E. Allan and G.C. McCormick, 'Measurements of the Backscatter Matrix of Dielectric Spheroids", IEEE trans. Antennas & Prop., Vol. AP-26, No. 4, July 1978.
- [33] L.E. Allan and G.C. McCormick, 'Measurements of Backscatter Matrix of Dielectric Bodies', IEEE Trans. Antennas & Prop., Vol. AP-28, No. 2, March 1980.
- [34] G.C. McCormick and A. Hendry, "Principles for the Radar Determination of the Polarization Properties of Precipitation", Radio Science, Vol. 10, pp. 421-434, April 1975.
- [35] W-M. Boerner, M.B. El-Arini, C-Y. Chan, S. Saatchi, W-S. Ip, P.M. Mastoris, B-Y. Foo, "Polarization Utilization in Radar Target Reconstruction", Technical Rept., UICC CL-EMID-NANRAR-81-01, Jan. 1981.
- [36] L.A. Morgan, S. Weisbrod, "RCS Matrix Studies of Sea Clutter", Teledyne Micronetics, Final Rept., San Diego, California, Mar. 1982.
- [37] H. Mieras and C.L. Bennett, "Space-Time Integral Equation Approach to Dielectric Targets", IEEE Trans. Antennas & Prop., Vol. AP-30, No. 1, January 1982.
- [38] S.H. Bickel, "Some Invariant Properties of the Polarization Scattering Matrix", Proc. IEEE, Vol. 53, pp. 1070-1072, Aug. 1965.
- [39] C.D. Graves, "Radar Polarization Power Scattering Matrix", Proc. IRE, Vol. 44, pp. 248-252, Feb. 1956.
- [40] J.S. Hollis, T.G. Hickman and T.J. Lyon, "Polarization Theory", in Microwave Antenna Measurements, Ed., by J.S. Hollis, T.J. Lyon and L. Clayton, Scientific-Atlanta, Inc., Atlanta, Georgia, July 1970.

- [41] G. Sinclair, "Modification of the Radar Equation for Arbitrary Targets and Arbitrary Polarization", Rept. #302-19, Antenna Lab., OSU, Columbus, Ohio, 1948.
- [42] J.R. Huynen, "Phenomenological Theory of Radar Targets", Ph.D Dissertation, Drukkerij Bronder-Offset, N.V., Rotterdom, 1970.

RESUME

: Thomas Bing-Yuen Foo Address: 838 S. Miller St.,

Chicago, Illinois, 60607 : (312) 829-7725 <Home>

: (312) 996-5140 <Office>

Date of Birth : January 25, 1956

Marital Status : Single Status : visa student

Place of Birth/Citizenship: Hong Kong

EDUCATION:

Phone

June 80 - present

University of Illinois at Chicago

Degree: M. Sc. in information Engineering Major : Electromagnetics, remote sensing Cumulative Grade Point: 5.00 out of 5.00

Expected Graduation Date: around end of Mar. 82

Sept 79 - June 80

Ohio State University, Columbus, Ohio Enrolled in Graduate School,

Department of Electrical Engineering

Major : Communications, Applied Electromagnetics Cumulative Graduate Grade Point: 4.00 out of 4.00

Sept 74 - May 78

University of Manitoba, Winnipeg, Canada

Degree : B. Sc. in Civil Engineering

Major : Structural Analysis

Cumulative Grade Point: 3.74 out of 4.00

EXPERIENCE:

June 80 - present

- Research Assistant in ^position

> Electromagnetic Imaging Group Communications Laboratory, UICC

Research/Thesis Adviser- Dr.Wolfgang-Martin Boerner

Research Area - Electromagnetic Inverse Scattering, Polarization Utilization in radar

target identification and shape

reconstruction

July 81 Temporarily involved in the balloon

> lauching project of NRCC in the investigation of powerline low frequency radiation to the earth

s atmosphere

Mar 81 - June 81

Position

- Teaching Assistant in electromagnetic theories

ACADAMIC HONORS:

Dean's Honor List in Civil Engineering University of Manitoba

PUBLICATIONS :

- (1) W-M Boerner, Chuk-Min Ho and Bing-yuen Foo, "Use of Radon's Projection Theory in electromagnetic Inverse Scattering", special issue, IEEE, vol. AP-29, Mar. 81
- (2) W-M Boerner, M. B. El-Arini, C. Y. Chan, S. S. Saatchi, Bing-Yuen Foo "Polarization Utilization in Radar Classification and Imaging" Electromagnetic Imaging Division, Communications Laboratory, Information Engineering Dept., U.I.C.C., 1981
- (3) W-M Boerner, M. B. El-Arini, C. Y. Chan, S. S. Saatchi, W. S. Ip, P. M. Mastoris, Bing-Yuen Foo "Polarization Utilization in Radar Target Reconstruction" Technical Report, UICC, CL-EMID-NANRAR-81-01 Jan. 81
- (4) S. K. Chaudhuri, B-Y. Foo, W-M Boerner,
 "A High Frequency Inverse Scattering Model to
 Recover the Specular Point Curvature from
 Polarimetric Scattering Matrix Data",
 University of Waterloo, Canada /University of
 Illinois at Chicago, Dec. 1981, (In print)

PROFESSIONAL ACTIVITIES :

Dec 80 - present

Student member of IEEE

Jan 82 - present

Member of American Association for the Advancement of Science

HOBBIES :

Table Tennis, Tennis, Badminton



COLLEGE OF ENGINEERING

NIVERSITY OF ILLINOIS AT CHICAGO CIRCLE

M.S. THESIS DEFENSE

DATE:

Friday, May 21, 1982

TIME:

1:00 p.m.

PLACE:

1127 SB0

SPEAKER:

TITIE:

Bing-Yuen (Thomas) Foo

ABSTRACT:

"A High Frequency Inverse Scattering Model to Recover the Specular Point Curvature from Polarimetric Scattering Data"

Based on the time-domain first order correction to the physical optics current approximation, a relationship between the phase factors of the polarimetric scattering matrix elements and the principal curvatures at the specular point of a scatterer is established.

The above phase-curvature relationship is tested by applying it to theoretical as well as experimental backscattering data obtained for a prolate spheroidal scatterer. The results of these tests not only determine the acceptability of the phase-curvature relationship, they also point out the range of kb values over which the first order correction to the physical optics currents is valid, and which serves as a compromise range between the high frequency condition required by the curvature recovery model and the drawback to lower frequencies required to prevent critical magnification of measurement errors.

Another curvature recovery equation is derived by transforming the linear polarization basis to the circular polarization basis. The curvature recovery model is proven to satisfy the image reconstruction identities of invariant transformation. A scattering ratio is defined and its behavior is investigated at high frequencies. Its plots on the complex plane provides a simple way to help judge the accuracy of polarimetric scattering measurement, regardless of whether a linear or a circular polarization basis is used.

